

[0001] **SYSTEM AND METHOD FOR RECEIVED
CHANNEL POWER INDICATOR (RCPI) MEASUREMENT**

[0002] **FIELD OF INVENTION**

[0003] The present invention relates to wireless communications systems, and more particularly, to a system and method for measuring a received channel power indicator (RCPI).

[0004] **BACKGROUND**

[0005] A received signal strength indicator (RSSI) is defined at the antenna input connector, but it is not fully specified, because there are no unit definitions and no performance requirements, such as accuracy or testability. It is not possible to extract meaning from a comparison of RSSIs from different stations and from different channels/physical layers (PHYs) within the same station. RSSI may have limited use for evaluating access point (AP) options within a station, such as a wireless local area network (WLAN) station, and within a given PHY, but is not useful for evaluations between PHYs. RSSI is rescaled between direct sequence spread spectrum (DSSS) and orthogonal frequency division multiplex (OFDM) PHYs. RSSI is clearly not useable by network management for handoff or load balancing. RSSI from one station does not relate to RSSI from any other station. In high interference environments, RSSI is not an adequate indicator of desired signal quality, since it indicates the sum of:

[0006] desired signal + noise + interference powers.

[0007] **SUMMARY**

[0008] A receiver analyses a signal in order to obtain a received channel power indicator (RCPI) value. The RCPI value is a measure of the received radio frequency (RF) power in the selected channel, measured at the antenna connector. This parameter is a measure by the PHY sublayer of the received RF power in the channel measured over the physical layer convergence protocol (PLCP) preamble and over the entire received frame. RCPI is a

monotonically increasing, logarithmic function of the received power level defined in dBm.

[0009] BRIEF DESCRIPTION OF THE DRAWINGS

[0010] Figure 1 is a block diagram of a system employing RCPI and PSNI according to the present invention.

[0011] Figures 2A and 2B are diagrams showing examples of RCPI and PSNI measurements.

[0012] Figure 3 shows RCPI and PSNI in relation to SNR in the demodulator.

[0013] Figure 4 is a graph showing PSNI value versus BER.

[0014] Figure 5 is a diagram of a first PLCP.

[0015] Figure 6 is a diagram of a second PLCP.

[0016] Figure 7 is a diagram of a third PLCP.

DETAILED DESCRIPTION OF THE PREFERRED EMBODIMENTS

[0017] A received channel power indicator (RCPI) disclosed herein provides a quantized, objective input power measure ($S+N+I$), wherein S is the signal power, N is the noise power, and I is the interference power. A perceived signal to noise indicator (PSNI) provides a value which is a measure of the perceived, post-processing signal-to-noise-plus-interference ($S/(N+I)$) ratio in the demodulator. The PSNI provides a quantized, comparative measure of received signal quality [observed $S/(N+I)$] for all channels/rates and among all PHYs and between all stations.

[0018] Figure 1 is a block diagram of a system 11 employing RCPI and PSNI according to the present invention. The system 11 includes a front end 12, an analog-digital converter 13, and a PHY specific demodulator and tracking loops module 14. The front end 12, an analog-digital converter 13, and a PHY specific demodulator and tracking loops module 14 form an AGC section 15. Also shown is an optional FEC decoder 17 and a frame check module 18. The system 11 resolves an incoming signal in a plurality of stages,

shown as A through E. The RCPI according to the present invention measures total radio frequency (RF) power A at an antenna input 21. This is the total RF power, measured as $RF\ S/(N+I)$ for each AP. The output of the front end 12, represented at B, is $BB\ S/(N+I)$ for each AP, where BB is a power constant used by the AGC 15. The output of the demodulator and tracking loops 14, which also corresponds to the output of the FEC decoder 17, is the bit error rate (BER) for each data rate from each AP. The output of the frame check module 18 is the frame error rate (FER) for each data rate from each AP. The PSNI measures observed $S/(N+I)$ within a demodulator but normalizes measurement for the FER at output E.

[0019] Figures 2A and 2B are diagrams showing examples of RCPI and PSNI measurements at different stations 31-33. The measurements at Figure 2A correspond to measured RCPI power at A (Figure 1). The signal is measured at the same objective SNR for the different stations 31-33. The measurements at Figure 2B correspond to a measured PSNI quantity, used to determine FER at E (Figure 1). The signal is measured at the same subjective SNR for the different stations 31-33. Although an accurate $S/(N+I)$ measurement at A is somewhat useful, RF demodulation implementations vary widely from station to station, and the $S/(N+I)$ measurement cannot be used comparatively between stations 31-33 to evaluate delivered signal quality. Accurate FER measurement at output E is an ideal quality measure, but cannot be measured frame by frame. FER can only be accurately measured over a large number of frames, typically several hundred to several thousand frames. Also, FERs are comparable only at the same frame size and data rate.

[0020] Figure 3 shows RCPI and PSNI in relation to SNR in the demodulator. Starting with Boltzman's constant, a thermal input noise level and interference power at input appear at the antenna connector input (21, Figure 1). The received channel power level is that signal, minus channel impairments, receive amp noise, demodulator loss, FEC decoder loss, and a combination of a theoretical SNR for required reception and operating margin.

The level at the antenna connector input is either the input SNR ratio or the input analog SNIR, depending on whether interference power at input is included. The signal after channel impairments, receive amp noise, demodulator loss and FEC decoder loss is the observed digital SNIR, which is the PSNI in the demodulator. The PSNI in the demodulator, when combined with the theoretical SNR for required reception, or the SNR for required reception plus the operating margin, is the received channel power level desired.

[0021] Figure 4 is a graph showing PSNI value versus BER.

[0022] The present invention measures input power RCPI, which is preferably specified in a manner similar to RSSI, for example by an eight (8) bit unsigned value, monotonically increasing with increasing signal strength. In one preferred embodiment, RCPI is logarithmically scaled in dB to measure the total received power within the defined channel bandwidth at the antenna connector A (see Figure 1), and includes power from the desired signal, noise and interference. RCPI values are quantified in dBm and accuracy is specified across the entire RCPI range. The 221 RCPI levels range from -110 dBm to 0 dBm with 2 units per dB. RCPI accuracy is +/- 5dB across the defined range, equivalent to accuracy specified by TGH for the RPI histogram. RCPI may be used wherever RSSI is specified. The above makes the RCPI value highly desirable and broadly applicable as a frame quality measurement useful for all PHYs and STAs. It is noted that the use of the RCPI provides ease of implementation with modern manufacturers and paves the way for advanced antenna designs.

[0023] The RCPI value is a measure of the received RF power in the selected channel, measured at the antenna connector A. This parameter is a measure by the PHY sublayer of the received RF power in the channel measured over the PLCP preamble and over the entire received frame. RCPI is a monotonically increasing, logarithmic function of the received power level defined in dBm. The allowed values for the RCPI parameter are an eight (8)

bit value in the range from 0 through 220 (decimal), with indicated values rounded to the nearest 0.5 dB as follows:

- 0: Power < -110 dBm
- 1: Power = -109.5 dBm
- 2: Power = -109.0 dBm

:
:
:

220: Power > -0 dBm

221-255: reserved

[0024] Accuracy for each measurement is +/- 5dB. The measurement assumes a receiver noise equivalent bandwidth of 22 MHz.

[0025] The perceived signal to noise plus interference (PSNI) indicator is a demodulator specific, post-processing estimation of observed S/(N+I) and BER/FER. All digital demodulators use tracking loops and complex post-processing to demodulate received symbols. Many internal demodulator metrics are proportional to perceived S/(N+I). Some examples include:

phase shift keying (PSK): baseband phase jitter and received error vector magnitude (EVM)

DSSS: spreading code correlation quality

OFDM: frequency tracking and channel tracking stability

[0026] Demodulator internal metrics are available on a frame-by-frame basis. Demodulator metrics proportional to S/(N+I) are available at all data rates. Demodulator internal metrics may be calibrated with respect to actual FER performance to accurately indicate perceived or observed S/(N+I) in a controlled environment with additive white Gaussian noise (AWGN). Such demodulator internal metrics provide fast estimations of S/(N+I) in both interference environments and interference-free (noise only) environments. In one potential implementation of IEEE 802.11, called Task Group K of 802.11 (TGK), it is not necessary to specify which demodulator metrics to use, but it is only necessary to specify how the quantized PSNI indicator relates to

S/(N+I) and FER. For example, TKG may specify that for STAs which comply with 802.11k, STAs may report RCPI (or PSNI) in any data field calling for RSSI. A new station management entity (SME) control variable could specify the use of the RSSI field: a) for RCPI, b) for PSNI (perceived signal to noise indication), or c) for RSSI.

[0027] One preferred embodiment of the present invention measures output signal quality to obtain PSNI. PSNI is specified like RSSI: an eight-bit unsigned value, monotonically increasing with increasing S/(N+I). PSNI is logarithmically scaled to perceived S/(N+I) which relates directly to FER performance. A PSNI output value is specified for each data rate using FER points: a first point to an "anchor" indicator, and additional points to quantize and scale the indicator slope and the range of values. Accuracy of the PSNI in AWGN is specified to be +/- 1dB for all FER points. PSNI range spans the lower 32 dB portion of the operating range of S/(N+I) to cover high FERs at data rates from 1 to 54 Mbps.

[0028] The PSNI indicator is a measure of the perceived, post-processing signal-to-noise-plus-interference (S/(N+I)) ratio in the demodulator. The allowed values for the PSNI parameter are an eight (8) bit value in the range from 0 through 255 (decimal). This parameter is a measure by the PHY sublayer of the perceived signal quality observed after RF downconversion to baseband, and is derived from internal digital signal processing metrics of the demodulator used to receive the current frame. PSNI is measured over the PLCP preamble and over the entire received frame. PSNI is intended to be used in a relative manner, and it is a monotonically increasing, logarithmic function of the observed S/(N+I). PSNI accuracy and range are specified in additive white Gaussian noise (AWGN) at given FERs for each data rate as depicted in Table 1.

[0029] Theoretical FEC coding gain assumed in FER calculations is given by:

R = 1/2, 5.4dB gain

R = 2/3, 4.7dB gain

$R = 3/4$, 4.4dB gain

[0030] The PSNI special value of zero indicates an inability to measure PSNI. When PSNI exceeds the high end of the measurable range for a given data rate, the maximum PSNI for that rate is reported.

[0031] The data rate/modulation (DRM) adjustments used to offset BER curves are as depicted in Table 2.

[0032] Table 3 shows the values obtained for E_b/N_o , BER, and FER for a PSNI = 101. Bit error rates (BERs) vary based on the forward error correction (FEC) coding used at each data rate. FERs vary based on BER and PPDU length. Note that for any station, PSNI will vary only as a result of changing channel conditions or changing received desired signal power level. Note that for efficiency, all stations preferably operate at the highest data rate possible while maintaining an acceptable FER (QoS).

[0033] Analyzing PSNI in relation to observed E_b/N_o , a PSNI value of zero is selected for a post-processing, observed E_b/N_o (O_{EbNo}) equal to 4.4dB, for binary phase shift keying (BPSK) at 1Mbit/s data rate. Eight units (steps) per dB is selected to provide a 32 dB range in an eight-bit PSNI value. For 1 Mbit/s BPSK operation, $PSNI = 8*[O_{EbNo} - 4.4dB]$. In general, for all other data rates and modulations:

[0034] $PSNI = 8*[O_{EbNo} - 4.4dB + DRM_x - CF_y]$. Equation (1)

[0035] where DRM_x is an S/N adjustment unique for each data rate/demodulation combination. DRM values are calculated in Table 1. CF_y is a hardware-specific factor used to account for implementation variances in each FEC decoder in the station. $CF_y = CG_{theo} - CG_{act} =$ actual FEC decoder loss, for each decoder at each specified FER point. CG_{theo} represents theoretical coding gain and CG_{act} represents actual coding gain. CG_{theo} values are listed in Table 1. When no FEC decoder is used, $CF_y = 0$. Equation 1 is the foundation of the PSNI measurement.

[0036] Analyzing PSNI in relation to input signal to noise plus interference ratio (ISNIR) as follows:

[0037] $SNR = C / N$, where $E_b = C * T_b$, $N = N_o * NBW$ (noise BW) and $DR=1/T_b$.

[0038] Thus,

[0039] $SNR = \frac{E_b / T_b}{N_o * NBW} = \frac{E_b * DR}{N_o * NBW} = E_b N_o * DR / NBW$ Equation (2)

[0040] In db: $SNR = E_b N_o + DR - NBW$ Equation (3a)

[0041] For $DR = 1$ Mbit/s and $NBW = 22$ MHz,

[0042] $SNR = E_b N_o + 60\text{dB} - 73.4\text{dB} = E_b N_o - 13.4\text{dB}$ Equation (3b)

[0043] In general,

[0044] $SNR = E_b N_o - 13.4\text{dB} \text{ DRM}_x$, with DRM_x from Table 1. Equation (4)

[0045] Therefore,

[0046] $E_b N_o = SNR + 13.4\text{dB} - \text{DRM}_x$ Equation (5a)

[0047] and

[0048] $O E_b N_o = \text{OSNIR} + 13.4\text{dB} - \text{DRM}_x$ Equation (5b)

[0049] Employing Equation 1, i.e., $PSNI = 8*[O E_b N_o - 4.4\text{dB} + \text{DRM}_x - CF_y]$, and substituting for $O E_b N_o$, $PSNI = 8*[(\text{OSNIR} + 13.4\text{dB} - \text{DRM}_x) - 4.4\text{dB} + \text{DRM}_x - CF_y]$, and $PSNI = 8*[\text{OSNIR} + 9.0\text{dB} - CF_y]$. Since

[0050] $\text{ISNIR} = \text{OSNIR} + \text{TML} + \text{CI}$, Equation (6)

[0051] where TML is the modem implementation loss and CI is the sum of all channel impairments, we have

[0052] $PSNI = 8*[(\text{ISNIR} - \text{TML} - \text{CI}) + 9.0\text{dB} - CF_y]$ Equation (7)

[0053] Analyzing $PSNI$ in relation to BER/FER , $PSNI$ is a direct measure of observed $SNIR$ considering all channel impairments and implementation losses measured at the demodulator. $PSNI$ is specified with respect to output FER , which considers all implementation losses including any FEC decoder implementation loss. Each station measures $PSNI$ using a correction factor CF_y to account for the actual coding gain (CG_{act}) of each FEC decoder. Any station measuring $PSNI$ on a frame using FEC will use CF_y so that the reported $PSNI$ from all stations is normalized and assumes a theoretical coding gain.

[0054] $CF_y = CG_{theo} - CG_{act} = \text{actual FEC decoder loss.}$ Equation (8)

[0055] Reported PSNI value may be used to estimate OEBNo and BER/FER (QOS) for the reporting station for each data rate.

[0056] $OEBNo = (PSNI/8) + 4.4dB - DRM_x$ Equation (9)

[0057] For data rates without an FEC decoder, OEBNo is used with the theoretical PSK EbNo curve to estimate BER. For data rates with an FEC decoder, OEBNo is used with the theoretical FEC EbNo curve for PSK to estimate BER. Note that the PSNI relationship to BER is specified only for AWGN. Since the net effect of channel impairments is to degrade OEBNo in the same way as AWGN, PSNI should be an adequate estimator of BER in all channel conditions.

[0058] PHY-RXSTART.indication (physical layer-receive start indication)

[0059] This is an indication by the PHY to the local medium access control (MAC) entity that the PLCP has received a valid start frame delimiter (SFD) and PLCP Header. The primitive provides the following parameter: PHY-RXSTART.indication (RXVECTOR):

[0060] The RXVECTOR represents a list of parameters that the PHY provides the local MAC entity upon receipt of a valid PLCP Header or upon receipt of the last PSDU data bit in the received frame. This vector contains both MAC and MAC management parameters. This primitive is generated by the local PHY entity to the MAC sublayer when the PHY has successfully validated the PLCP header error check (HEC) cyclic redundancy code (CRC) at the start of a new PLCP protocol data unit (PPDU). The effect of receipt of this primitive by the MAC is unspecified.

[0061] PHY-RXEND.indication

[0062] This primitive is an indication by the PHY to the local MAC entity that the MPDU currently being received is complete. The primitive provides the following parameters: PHY-RXEND.indication (RXERROR,

RXVECTOR). The RXERROR parameter can convey one or more of the following values: NoError, FormatViolation, CarrierLost, or UnsupportedRate. A number of error conditions may occur after the PLCP's receive state machine has detected what appears to be a valid preamble and SFD. The following describes the parameter returned for each of those error conditions.

[0063] *NoError*. This value is used to indicate that no error occurred during the receive process in the PLCP.

[0064] *FormatViolation*. This value is used to indicate that the format of the received PPDU was in error.

[0065] *CarrierLost*. This value is used to indicate that during the reception of the incoming protocol data unit (MPDU), the carrier was lost and no further processing of the MPDU can be accomplished.

[0066] *UnsupportedRate*. This value is used to indicate that during the reception of the incoming PPDU, a nonsupported data rate was detected.

[0067] The RXVECTOR represents a list of parameters that the PHY provides the local MAC entity upon receipt of a valid PLCP Header or upon receipt of the last PSDU data bit in the received frame. This vector contains both MAC and MAC management parameters. This primitive is generated by the PHY for the local MAC entity to indicate that the receive state machine has completed a reception with or without errors. The effect of receipt of this primitive by the MAC is unspecified.

[0068] Receive PLCP

[0069] A first embodiment of a receive PLCP is shown in Figure 5. In order to receive data, PHY-TXSTART.request is disabled so that the PHY entity is in the receive state. Further, through station management via the Physical Layer Management Entity (PLME), the PHY is set to the appropriate channel and the Clear Channel Assessment (CCA) method is chosen. Other receive parameters such as RSSI, RCPI, signal quality (SQ), and indicated

data rate are accessed via the PHY-service access point (PHY-SAP), shown in Table 4.

[0070] Upon receiving the transmitted energy, according to the selected CCA mode, the physical layer medium - energy detection (PMD_ED) is enabled when the RSSI level crosses the ED_THRESHOLD and/or PMD_Carrier_Sense (PMD_CS) is enabled after code lock is established. These conditions are used to indicate activity to the MAC via PHY-CCA.indicate. PHY-CCA.indicate (BUSY) is issued for energy detection (ED) and/or code lock prior to correct reception of the PLCP frame. The PMD primitives PMD - signal quality (PMD_SQ) and PMD_RSSI are issued to update the RSSI and SQ parameters reported to the MAC.

[0071] After PHY-CCA.indicate is issued, the PHY entity begins searching for the Start of Frame Delimiter SFD field. Once the SFD field is detected, CRC-16 processing is initiated and the PLCP IEEE 802.11 SIGNAL, IEEE 802.11 SERVICE and LENGTH fields are received. The CRC-16 frame check sequence (FCS) is processed, and if the CRC-16 FCS check fails, the PHY receiver returns to the RX IDLE state. Should the status of CCA return to the IDLE state during reception prior to completion of the full PLCP processing, the PHY receiver returns to the RX IDLE state.

[0072] If the PLCP Header reception is successful (and the SIGNAL field is completely recognizable and supported), a PHY-RXSTART.indicate (RXVECTOR) is issued. The RXVECTOR associated with this primitive includes the SIGNAL field, the SERVICE field, the -- MAC Protocol Data Units (MPDU) length in bytes (calculated from the LENGTH field in microseconds), the antenna used for receive (RX_ANTENNA), RSSI, PSNI, and SQ.

[0073] PMD_RCPL.indicate (RCPI).

[0074] This primitive is the measure of the channel power received by the DSSS PHY measured at the antenna connector, and is generated by the PMD sublayer, provides the received channel power indicator to the PLCP and

MAC. The primitive includes the following parameter: PMD_RCPI.indicate (RCPI). The RCPI is a measure of the channel power received by the direct sequence spread spectrum PHY (DSSS PHY) as measured at the antenna connector. RCPI value indications of eight bits (221 levels) are supported. This primitive is generated by the PMD when the DSSS PHY is in the receive state. It is continuously available to the PLCP, which, in turn, provides the parameter to the MAC entity. This parameter is provided to the PLCP layer for information only. The RCPI may be used in conjunction with PSNI to measure input signal quality.

[0075] RXVECTOR RCPI

[0076] The allowed values for the RCPI parameter are in the range from 0 through 220 and is a measure by the PHY of the received channel power at the antenna connector. RCPI value indications of eight bits (221 levels) are supported. PSNI is measured during the reception of the PLCP preamble and over the entire received frame. The RCPI is measured in the same manner as described above.

[0077] A second embodiment of a receive PLCP is shown in Figure 6. In order to receive data, PHY-TXSTART.request is disabled so that the PHY entity is in the receive state. Further, through station management (via the PLME) the PHY is set to the appropriate frequency. Other receive parameters, such as RSSI, RCPI, and indicated data rate, are accessed via the PHY-SAP as previously described.

[0078] PMD_RCPI.indicate

[0079] This primitive, generated by the PMD sublayer, provides the received channel power indicator to the PLCP and MAC entity and provides the following parameter: PMD_RCPI.indicate (RCPI).

[0080] The RCPI is a measure of the channel power received by the DSSS PHY as measured at the antenna connector. RCPI value indications of eight bits (221 levels) are supported. This primitive is generated by the PMD

when the DSSS PHY is in the receive state. It is continuously available to the PLCP, which, in turn, provides the parameter to the MAC entity. This parameter is provided to the PLCP layer for information only. The RCPI may be used in conjunction with PSNI to measure input signal quality.

[0081] Receive PLCP

[0082] The receive procedures for receivers configured to receive the mandatory and optional PLCPs, rates, and modulations are described herein. A receiver that supports this High Rate extension of the standard is capable of receiving 5.5 Mbit/s and 11 Mbit/s, in addition to 1 Mbit/s and 2 Mbit/s. If the PHY implements the Short Preamble option, it detects both Short and Long Preamble formats and indicates which type of preamble was received in the RXVECTOR. If the PHY implements the packet binary convolution code (PBCC) Modulation option, it detects either complementary code keying (CCK) or PBCC Modulations, as indicated in the SIGNAL field, and reports the type of modulation used in the RXVECTOR.

[0083] The receiver implements the CCA procedure, in which, upon receiving a PPDU, the receiver distinguishes between a long and short header format by the value of the SFD. The receiver demodulates a long PLCP header using BPSK at 1 Mbit/s. The receiver demodulates a short PLCP header using QPSK at 2 Mbit/s. The receiver uses the SIGNAL and SERVICE fields of the PLCP header to determine the data rate and modulation of the PSDU.

[0084] A third embodiment of a receive PLCP is shown in Figure 7. In order to receive data, the PHY-TXSTART.request is disabled so that the PHY entity is in the receive state. Further, through station management via the PLME, the PHY is set to the appropriate channel and the CCA method chosen. Others receive parameters, such as RSSI, RCPI, SQ, and indicated data rate, are accessed via the PHY-SAP as described previously.

[0085] Upon receiving the transmitted energy, according to the selected CCA mode, the PMD_ED is enabled as the RSSI reaches the

ED_THRESHOLD, and/or PMD_CS is enabled after code lock is established. These conditions are used to indicate activity to the MAC via PHY-CCA.indicate. PHY-CCA.indicate (BUSY) is issued for ED and/or code lock prior to correct reception of the PLCP header. The PMD primitives, PMD_SQ, PMD_RSSI, PMD_RCPI, are issued to update the RSSI, RCPI, and SQ parameters reported to the MAC.

[0086] After PHY-CCA.indicate is issued, the PHY entity begins searching for the SFD field. Once the SFD field is detected, CRC-16 processing is initiated and the PLCP SIGNAL, SERVICE, and LENGTH fields are received. The CRC-16 FCS is processed. If the CRC-16 FCS check fails, the PHY receiver returns to the RX IDLE state. If the status of CCA returns to the IDLE state during reception prior to completion of the full PLCP processing, the PHY receiver returns to the RX IDLE state.

[0087] If the PLCP header reception is successful (and the SIGNAL field is completely recognizable and supported), a PHY-RXSTART.indicate (RXVECTOR) is issued. The RXVECTOR associated with this primitive includes: the SIGNAL field; the SERVICE field; the PSDU length in octets (calculated from the LENGTH field in microseconds and the data rate in Mbit/s); RXPREAMBLE_TYPE (which is an enumerated type taking on values SHORTPREAMBLE or LONGPREAMBLE); and the antenna used for receive (RX_ANTENNA), RSSI, RCPI, and SQ.

[0088] PMD_PSNI.indicate

[0089] following parameter:

Parameter	Associated primitive	Value	Description
RCPI	PMD_RCPI.indicate	8 bits of RCPI	The RCPI is a measure of the antenna connector channel power received by the High Rate PHY, as defined above in paragraph [0023].

This primitive is generated by the PMD when the High Rate PHY is in

the receive state. It is continuously available to the PLCP, which, in turn, provides the parameter to the MAC entity. This parameter is provided to the PLCP layer for information only. The RCPI may be used in conjunction with PSNI to measure input signal quality. The RCPI is measured as described above.

Table 1. PSNI Accuracy and Range

Data Rate (Mbit/s)	Mode	FEC Rate	PPDU Length (bytes)	FER 0.2%	PSNI
1	DSSS	no FEC	125	10% ± 0.2%	32 ± 8
2	DSSS	no FEC	1250	10% ± 0.2%	66 ± 8
5.5	HRDSSS	no FEC	1250	10% ± 0.2%	101 ± 8
5.5	HRDSSS	1/2	1250	10% ± 0.2%	58 ± 8
6	OFDM	1/2	1250	10% ± 0.2%	61 ± 8
9	OFDM	3/4	1250	10% ± 0.2%	83 ± 8
11	HRDSSS	no FEC	1250	10% ± 0.2%	125 ± 8
11	HRDSSS	1/2	1250	10% ± 0.2%	82 ± 8
12	OFDM	1/2	1250	10% ± 0.2%	85 ± 8
18	OFDM	3/4	1250	10% ± 0.2%	107 ± 8
24	OFDM	1/2	1250	10% ± 0.2%	141 ± 8
36	OFDM	3/4	1250	10% ± 0.2%	163 ± 8
48	OFDM	2/3	1250	10% ±	206 ± 8

				0.2%	
54	OFDM	3/4	1250	10% ± 0.2%	213 ± 8

Table 2: Data Rate/Modulation (DRM) Adjustments

Data rate (Mbps)	Mode	Modulation	FEC code rate	Coded rate (Mbps)	Data rate adj. (db)	Modulation adj. (db)	DRM: Total S/N adj.(db)
1	DSSS	BPSK	none (SF=11)	1	0	0	0
2	DSSS	QPSK	none (SF=11)	2	3	0	3
5.5	DSSS 11b	QPSK	none (SF=16)	5.5	7.4	0	7.4
5.5	DSSS 11b	QPSK	1/2 (SF=2)	11	7.4	0	7.4
6	OFDM	BPSK	1/2	12	7.78	0	7.78
9	OFDM	BPSK	3/4	12	9.54	0	9.54
11	DSSS 11b	QPSK	1/2 (SF=1)	11	10.41	0	10.41
11	DSSS 11b	QPSK	none (SF=8)	11	10.41	0	10.41
12	OFDM	QPSK	1/2	24	10.79	0	10.79
18	OFDM	QPSK	3/4	24	12.55	0	12.55
22	DSSS 11g	8PSK	2/3 (SF=1)	33	13.42	0	13.42
24	OFDM	16-QAM	1/2	48	13.8	3.98	17.78
33	DSSS 11g	8PSK	2/3 (SF=1)	49.5	15.19	0	15.19

36	OFDM	16-QAM	$\frac{3}{4}$	48	15.56	3.98	19.54
48	OFDM	64-QAM	$\frac{2}{3}$	72	16.81	8.45	25.26
54	OFDM	64-QAM	$\frac{3}{4}$	72	17.82	8.45	25.77

Table 3.

Data Rate (Mbps)/FEC	Eb/No (dB)	BER	PPDU Length	FER
5.5 (none)	9.6	1.00E-05	1250	1.00E-01
5.5 (R=1/2)	9.6	1.25E-13	1250	1.00E-09
6 (R=1/2)	9.2	1.25E-12	1250	1.00E-08
9 (R=3/4)	7.5	6.00E-09	1250	4.80E-05
11 (none)	6.6	1.25E-03	10	1.00E-01
11 (R=1/2)	6.6	2.00E-09	12	2.00E-05
12 (R=1/2)	6.2	2.00E-08	1250	2.00E-04
18 (R=3/4)	4.4	1.40E-04	10	1.10E-02

Table 4. PHY-SAP service primitive parameters

Parameter	Associated primitive	Value
DATA	PHY-DATA.request PHY-DATA.indication	Octet value X'00'-X'FF'
TXVECTOR	PHY-TXSTART.request	A set of parameters
STATUS	PHY-CCA.indication	BUSY, IDLE
RXVECTOR	PHY-RXSTART.indication PHY-RXEND.indication	A set of parameters
RXERROR	PHY-RXEND.indication	NoError, FormatViolation, Carrier-Lost, UnsupportedRate

Table 5. PMD_SAP peer-to-peer service primitive parameters

Parameter	Associated primitive	Value
LENGTH	RXVECTOR, TXVECTOR	0 to 2E13-1
DATARATE	RXVECTOR, TXVECTOR	1,2 Mbit/s
SERVICE	RXVECTOR, TXVECTOR	1,2 Mbit/s
TXPWR_LEVEL	TXVECTOR	1,2 Mbit/s
TX_ANTENNA	TXVECTOR	1,2 Mbit/s
RSSI	RXVECTOR	1,2 Mbit/s
RCPI	RXVECTOR	1,2 Mbit/s
SQ	RXVECTOR	1,2 Mbit/s
RX_ANTENNA	RXVECTOR	1,2 Mbit/s

Table 6. PMD_SAP sublayer-to-sublayer service primitives

Primitive	Request	Indicate	Confirm	Response
PMD_TXSTART	X		--	--
PMD_TXEND	X	X	--	--
PMD_ANTSEL	X		--	--
PMD_TXPWRLEVL	X		--	--
PMD_RATE	X	X	--	--
PMD_RSSI		X	--	--
PMD_RCPI		X	--	--
PMD_SQ		X	--	--
PMD_CS		X	--	--
PMD_ED	X	X	--	--

Table 7. PMD_SAP service primitive parameters

Parameter	Associated Primitive	Value
DATA	PHY-DATA.request PHY-DATA.indicate	Octet value: X'00'-X'FF'
TXVECTOR	PHY-DATA.request	A set of parameters
RXVECTOR	PHY-DATA.indicate	A set of parameters
TX_UNIT	PMD-DATA.request	1,0: DBPSK dibit combinations 00,01,11,10: DQPSK
RX_UNIT	PMD-DATA.indicate	1,0: DBPSK dibit combinations 00,01,11,10: DQPSK
RF_STATE	PMD_TXE.request	Receive, Transmit
ANT_STATE	PMD- ANTSEL.indicate PMD-ANTSEL.request	1 to 256
TXPWR_LEVEL	PHY-TXSTART	0,1,2,3 (max of 4 levels)
RATE	PMD-RATE.indicate PMD-RATE.request	X'0A' for 1 Mbit/s DBPSK X'14' for 2 Mbit/s DQPSK
RSSI	PMD-RSSI.indicate	0-8 bits of RSSI
RCPI	PMD-RCPI.indicate	8 bits of RCPI
SQ	PMD-SQ.indicate	0-8 bits of SQ

Table 8. RXVECTOR parameters

Parameter	Associated primitive	Value
LENGTH	PHY-RXSTART.indicate	0 to 4095
RSSI	PHY-RXSTART.indicate (RXVECTOR)	0 to RSSI maximum
RCPII	PHY-RXSTART.indicate (RXVECTOR) PHY-RXEND.indicate (RXVECTOR)	0 to 255
DATARATE	PHY-RXSTART.request (RXVECTOR)	6, 9, 12, 18, 24, 36, 48, 54
SERVICE	PHY-RXSTART.request (RXVECTOR)	Null

Table 9. PMD_SAP sublayer-to-sublayer service primitives

Primitive	Request	Indicate	Confirm	Response
PMD_TXSTART	X	--	--	--
PMD_TXEND	X	--	--	--
PMD_TXPWRLEVL	X	--	--	--
PMD_RATE	X	--	--	--
PMD_RSSI	--	X	--	--
PMD_RCPI	--	X	--	--

Table 10. List of parameters for the PMD primitives

Parameter	Associated primitive	Value
TXD_UNIT	PMD-DATA.request	One(1), Zero(0): one OFDM symbol value
RXD_UNIT	PMD-DATA.indicate	One(1), Zero(0): one OFDM symbol value
TXPWR_LEVEL	PHY-TXSTART	1-8 (max of 8 levels)
RATE	PMD-RATE.request	12 Mbit/s (for BPSK) 24 Mbit/s (for QPSK) 48 Mbit/s (for 16-QAM) 72 Mbit/s (for 64-QAM)
RSSI	PMD-RSSI.indicate	0-8 bits of RSSI
RCPI	PMD-RCPI.indicate	8 bits of RCPI

Table 11. Parameter vectors

Parameter	Associated vector	Value
DATARATE	RXVECTOR, TXVECTOR	The rate used to transmit the PSDU in Mbit/s.
LENGTH	RXVECTOR, TXVECTOR	The length of the PSDU in octets.
PREAMBLE_TYPE	RXVECTOR, TXVECTOR	The preamble used for the transmission of this PPDU. This is an enumerated type that can take the value SHORTPREAMBLE or LONGPREAMBLE.
MODULATION	RXVECTOR, TXVECTOR	The modulation used for the transmission of this PSDU. This is an integer where 0 means CCK and 1 means PBCC.
RX_ANTENNA	RXVECTOR	The antenna used for receive.
RSSI	RXVECTOR	0-8 bits of RSSI
RCPI	RXVECTOR	8 bits of RCPI
SQ	RXVECTOR	0-8 bits of SQ

Table 12. PMD_SAP sublayer-to-sublayer service primitives

Primitive	Request	Indicate	Confirm	Response
PMD_TXSTART	X	--	--	--
PMD_TXEND	X	--	--	--
PMD_PSNI	--	X	--	--
PMD_TXPWRLVL	X	--	--	--
PMD_MODULATION	X	X	--	--
PMD_PREAMBLE	X	X	--	--
PMD_RATE	X	X	--	--
PMD_RSSI	--	X	--	--
PMD_RCPI	--	X	--	--
PMD_SQ		X	--	--
PMD_CS		X	--	--
PMD_ED	X	X	--	--

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What is claimed is:

1. A method for determining a received channel power indicator (RCPI), comprising the steps of:
 - a) measuring radio frequency (RF) power of a signal received at an antenna;
 - b) converting the measurement to a given parameter; and
 - c) scaling the parameter to obtain the RCPI value.
2. The method of claim 1 further comprising:
presenting the RCPI value in an eight (8) bit digital format corresponding to a decibel (dB) value.
3. The method of claim 2 further comprising presenting the RCPI value in a range from 0 through 221 corresponding to a decibel (dB) value.
4. The method of claims 1, 2 or 3 further comprising measuring RCPI during reception of a physical layer convergence protocol (PLCP) preamble and over reception of an entire frame.
5. The method of claims 1, 2 or 3 further comprising measuring RCPI at a physical layer (PHY).
6. The method of claims 1, 2 or 3 further comprising measuring RCPI at a direct sequence spread spectrum (DSSS).
7. The method of claim 5 further comprising providing the RCPI to a medium access control (MAC) layer.

8. The method of any one of the preceding claims wherein step (a) comprises measuring a signal which includes signal, noise and interference.

9. A method for determining a perceived signal to noise indicator for a given wireless station, which is a quantized, comparative measure expressed in decibels (dB) of received signal quality represented by signal to noise plus interference ratio (S/N+I) regardless of channel used for radio transmission and transmission rate comprising:

determining a relationship between PSNI and signal to noise ratio (Eb/No) by obtaining an observed Eb/No represented by O EbNo, when PSNI=O employing binary phase shift keying (BDSK) at a give data rate;

employing predetermined number of units per dB to obtain a predetermined range in for a PSNI value represented by a corresponding number of binary bits; and

modifying the PSNI relationship by adding a data rate/modulation (DRMX) value to accommodate a given data rate and modulation and a given hardware factor (CFy) to account for variances in each forward error correction (FEC) decoder in the station.

10. The method of claim 9 further comprising:

employing a predetermined number of units per dB to obtain a predetermined dB range for a PSNI value represented by a predetermined number of binary bits, whereby for operation at said given date rate employing

BPSK, $PSNI = XdB * [OEbNo - 4.4dB + DRMX - Cfy]$, where XdB represents a predetermined dB value; and

modifying the PSNI relationship by adding a data rate/modulation (DRMX) value to accommodate a given data rate and modulation and a given hardware factor (CFy) to account for variances in each forward error correction (FEC) decoder in the station and multiplied by the predetermined number of units per dB (XdB) whereby

$$PSNI = XdB * [OEBNo - 4.4dB + DRMx - CFy].$$

11. The method of claim 10 further comprising substituting $OSNIR + 13.4dB - DRMx$ for $OEBNo$, where $OSNIR$ is an observe signal to noise plus interference ration, and $PSNI$ is represented by

$$PSNI = XdB * [OSNIR + 9.0dB - CFy].$$

12. The method of claim 11 further comprising substituting $ISNIR - TML - CI$ for $DSNIR$, where TML represents modem implementation loss and CI represents a sum of all channel impairments

$$PSNI = XdB * [(ISNIR - TM - CI + 9.0dB - CFy)]$$

yielding a $PSNI$ which is a direct measure of observed $SNIR$.

13. The method of claim 9 wherein the correction factor CFy is utilized to account for an actual coding gain ($CGact$) of an FEC decoder.

14. The method of claim 9 further comprising normalizing $PSNI$ for all stations by employing the value CFy .

15. The method of claim 9 further comprising normalizing $PSNI$ for all stations by employing the value CFy so that $PSNI$ assumes a theoretical coding gain.

16. In a wireless slotted communications system which includes at least one base station and at least one WTRU, receiver for use in one of the base station or WTRU, capable of determining a received channel power indicator (RCPI), the receiver comprising:

a radio front end capable of measuring radio frequency (RF) power of a signal received at an antenna;

an AGC circuit for converting the measurement to a given parameter; and

a circuit for scaling the given parameter in decibels (dB), to obtain an RCPI value.

17. The receiver of claim 16 further wherein the circuit for scaling presents the RCPI value in an eight (8) bit digital format at a range from 0 through 221.

18. The receiver of claims 16 or 17 wherein the receiver measures RCPI during reception of a physical layer convergence protocol (PLCP) preamble and over reception of an entire frame.

19. The receiver of claims 16 or 17 wherein the receiver measures RCPI at a physical layer (PHY).

20. The receiver of claims 16 or 17 wherein the receiver measures RCPI at a direct sequence spread spectrum (DSSS).

21. The receiver of claim 16 wherein the receiver provides the RCPI to a medium access control (MAC) layer.

22. The receiver of any one of claims 16 to 21 wherein the receiver measures a signal which includes signal, noise and interference.

23. The receiver of any one of claims 16 to 22 further comprising:
a circuit for determining a relationship between PSNI and signal to noise ratio (E_b/N_o) by obtaining an observed E_b/N_o represented by O_{EbNo} , when $PSNI=0$ employing binary phase shift keying (BPSK) at a given data rate, employing a predetermined units per dB to obtain a predetermined dB range in for a PSNI value represented a corresponding number of binary bits; and

a circuit for ranging the PSNI value and modifying the PSNI relationship by adding a data rate/modulation (DRMX) value to accommodate a given data rate and modulation and a given hardware factor (CFy) to account for variances in each forward error correction (FEC) decoder in the station.

24. The receiver of claim 23 further comprising:
the circuit for ranging the PSNI value and modifying the PSNI relationship substituting $OSNIR + 13.4\text{dB-DRM}_x$ for O_{EbNo} , where $OSNIR$ is an observe signal to noise plus interference ratio; PSNI is represented by $PSNI = X_{dB} * [OSNIR + 9.0\text{dB-CF}_y]$, where X_{dB} represents a predetermined dB value, substituting $ISNIR - TML - CI$ for $DSNIR$, where TML represents modem implementation loss and CI represents a sum of all channel impairments, yielding a PSNI which is a direct measure of observed SNIR.

25. In a multiuser slotted wireless communications which includes at least one base station and at least one WTRU, a method for scaling received signals, the method comprising:

establishing a perceived signal to noise indicator (PSNI) value, the PSNI value providing a measure of perceived, post-processing signal-to-noise-plus-interference ratio ($S/(N+I)$) ratio in a demodulator so as to measure in a current received frame at a PHY sublayer a perceived signal quality observed after an RF downconversion, the measurement derived from predetermined internal digital signal processing metrics of the demodulator

26. The method of claim 25, further characterized by measuring the PSNI value over measured over the physical layer convergence protocol (PLCP) preamble and over the received frame.

27. The method of claims 25 or 26, further characterized by utilizing the PSNI value as a monotonically increasing, logarithmic function of the observed $S/(N+I)$.

28. The method of claims 25 or 26, further characterized by utilizing the PSNI value as a monotonically increasing, logarithmic function of the observed $S/(N+I)$, the PSNI providing an indication of additive white Gaussian noise (AWGN) at given FERs for each data rate.

29. The method of any one of claims 25 to 28, further characterized by utilizing values for the PSNI parameter as an 8 bit value in the range from 0 through 255.

30. In a multiuser wireless communications system, a method for scaling received signals, the method comprising:

receiving a signal and demodulating the signal to at least a physical sublayer; and

establishing a perceived signal to noise indicator (PSNI) value, the PSNI value providing a measure of perceived, post-processing signal-to-noise-plus-interference ratio ($S/(N+I)$) ratio for the physical sublayer of the demodulated signal, the measurement derived from predetermined internal digital signal processing metrics of the demodulator.

31. The method of claim 30, wherein the PSNI value provides a measure in a current received frame at the physical sublayer of perceived, post-processing signal-to-noise-plus-interference ($S/(N+I)$) ratio in the demodulator.

32. The method of claims 30 or 31, further characterized by measuring the PSNI value over measured over the physical layer convergence protocol (PLCP) preamble and over the received frame, the PSNI value utilized as a monotonically increasing, logarithmic function of the observed $S/(N+I)$.

33. The method of claims 30 or 31, further characterized by measuring the PSNI value over measured over the physical layer convergence protocol (PLCP) preamble and over the received frame, the PSNI value utilized as a monotonically increasing, logarithmic function of the observed $S/(N+I)$ and providing an indication of additive white Gaussian noise (AWGN) at given FERs for each data rate.

34. The method of any one of claims 30 to 33, further characterized by utilizing values for the PSNI parameter as an 8 bit value in the range from 0 through 255.

35. A multiuser communications receiver comprising:

an RF stage receiving an antenna input signal;
a demodulator for demodulating the signal to at least a physical sublayer;
a circuit for determining a perceived signal to noise indicator (PSNI) value, as described in claim 30; and
a circuit for providing a signal adjustment value in response to the PSNI.