

Description

The present invention relates generally to recording and/or reproducing apparatus, and more particularly to apparatus in which audio signal is modulated upon a carrier whose frequency is switched from one frequency to another in response to a transducer head being switched from one track to the next and recorded on a common track with a video signal and upon playback a tracking control signal is derived from frequency-modulated crosstalk components detected from side tracks.

In helical scan type magnetic tape recording and reproducing apparatus, a pair of transducer heads are mounted at equal angular positions on the circumference of a rotary drum and a magnetic tape is looped around the drum over an arc of 180-degree and transported at a speed much lower than the circumferential speed of the rotary drum, and audio signal is recorded through stationary heads along one edge of the tape.

In video tape recording, the recent effort is toward recording as much video information as possible in as small a tape length as possible and lowering the tape transport speed. One disadvantage of the stationary head recording is that reduction in tape transport speed tends to degrade sound quality much more than it adversely affects on picture quality.

To overcome this problem, shared use of common tracks by video and sound signals has been developed. According to this method, baseband audio signal is modulated on a single carrier having a frequency differing from the frequency ranges of modulated video components. Because of the high rotational speed of the rotary drum, high quality sound recording is made possible even when the tape transport speed is lowered.

On the other hand, recent advances in tape recording material and in transducer heads have made possible high density video recording, which has enabled reduction of tape speed to such an extent that a single cassette tape for home video tape recorders can hold six-hour recordings. In such high density recording, signals are recorded on narrow, closely spaced apart tracks with no guard bands therebetween. A high degree of precision is thus required for controlling the transducer head to keep it in registry with desired track. This is particularly important for home video tape recorders where the tape drive mechanism is required to be as simple as possible to meet the objectives of low cost and ease of servicing. Difficulty has been encountered to ensure on-track registration when reproducing a material recorded on a different apparatus. The same applies to recorders of the type wherein special features are provided which include slow-motion, fast-motion and frame freeze reproduction. In such recorders use is made of a head moving mechanism which moves the head in a direction

transverse to the length of track in response to an error signal representing the deviation of the head from the intended track.

To overcome this problem, one current approach (GB-A-2095436) involves the use of a pilot signal generator for generating pilot signals having different frequencies and recording the pilot signals on different tracks, and an envelope detector for detecting the envelope of the pilot signals to derive a tracking control signal therefrom. However, the pilot signal generator adds to the complexity and cost of the equipment and tends to interfere with desired signals.

Another approach is to record a frequency-modulated video signal and to oscillate the video head transversely at a frequency 480 Hz during playback to cause variations in the FM signal to extract a tracking control signal from such variations. However, this approach still suffers from a drawback in that the FM signal variations appear at the output of a frequency demodulator as a noise.

In a recording apparatus of the type having a transducer head which scans a recording medium along each of a plurality of tracks, an information signal is frequency-modulated upon carriers of different frequencies to generate a series of frequency-modulated different frequency carriers each of which occurs in synchronism with the shifting of the transducer head from one track to the next. The modulated carriers are serially applied to the head to record each of the different frequency carriers on a different track of the recording medium, so that there is a difference in frequency between carriers recorded on adjacent tracks. During playback, the frequency-modulated carriers are sequentially reproduced and separated into different frequency carrier components according to different tracks. The carrier components are fed to frequency demodulators to be demodulated respectively, and sequentially combined to regenerate the original information signal. The invention allows tracking control to be effected on the transducer head in response to at least one of the individual carriers which are derived from tracks on either side of the track being scanned.

Preferably, the tracking control signal is derived from those carriers which are in turn derived from opposite sides of the intended track by detecting the difference in amplitude therebetween.

The present invention will be described in further detail with reference to the accompanying drawings, in which:

Fig. 1 is a plan view illustrating a recording head assembly of a magnetic tape recording apparatus;

Fig. 2 is a side elevational view of the recording head assembly of Fig. 1;

Fig. 3 is a block diagram of the recording circuit of an embodiment of the invention;

Fig. 4 is an illustration of the energy distribution of the video and audio signal components

over the video frequency spectrum;

Fig. 5 is an illustration of a track pattern according to an embodiment of the invention;

Fig. 6 is a block diagram of a sound reproduction circuitry;

Fig. 7 is a block diagram of a video reproduction circuitry;

Figs. 8A and 8B are illustrations of a first example of the right-channel recording and reproduction circuits, respectively;

Figs. 9A and 9B are illustrations of a second example of the right-channel recording and reproduction circuits, respectively;

Figs. 10 and 10a are illustrations of another embodiment of the invention for providing time division multiplexing of stereophonic signals;

Fig. 11 is an illustration of a track pattern recorded according to the circuitry of Fig. 10; and

Fig. 12 is a block diagram from demultiplexing the time-division multiplexed signals, and Fig. 12A is a timing diagram for describing the operation of this circuitry.

Referring now to Figs. 1 and 2, there is shown a helical scan magnetic tape recording head assembly as schematically illustrated. The assembly includes a pair of video transducer heads H_1 and H_2 mounted in diametrically opposite positions of the circumference of an upper rotary drum 1 which is rotatably and coaxially mounted on a lower stationary drum 2. Also mounted on the circumference of the rotary drum 1 is a pair of audio transducer heads HA_1 and HA_2 in diametrically opposite positions advanced by angle θ from the video heads H_1 and H_2 , respectively. A magnetic tape 3 is looped around an arc somewhat greater than a 180° arc by means of a pair of guide posts 4a and 4b in such an orientation that each of the video and audio heads makes a helical pass over a section of the tape 3 which is transported in the direction X as the drum 1 is rotated counterclockwise to record video (field) and sound signals along a common track skewed relative to the length of the tape. Video heads H_1 and H_2 are located a distance d higher than audio heads HA_1 and HA_2 as shown in Fig. 2 so that video and audio heads H_1 and HA_1 form a pair that makes contact with the tape at substantially the same time and heads H_2 and HA_2 form another pair that makes contact with it substantially the same time during each helical scan. Typically, with a rotary drum having a diameter of 62 mm, the angle of advance θ and the vertical spacing d are 55° and 18 micrometers, respectively.

Preferably, video heads H_1 and H_2 are oriented so that their azimuths, or vertical settings of head gaps have $+6^\circ$ and -6° inclinations, respectively, and audio heads HA_1 and HA_2 are oriented so as to provide a greater angular difference therebetween than is provided between video heads H_1 and H_2 . Suitable azimuths for audio heads HA_1 and HA_2 are -30° and $+30^\circ$, respectively. These angular relationships minimize mutual interference between audio signals and interference between audio and video signals. The video and

audio heads are mounted respectively on a head moving mechanism which is preferably comprises a bimorph element responsive to an electrical signal applied thereto to contract or expand to minimize lateral displacement of the head from the right track.

In Fig. 3, the video and audio recording circuitry of an embodiment of the present invention is shown. The recording circuit generally comprises a left-channel sound recording section 67L, a right-channel sound recording circuit 67R, and a conventional video recording section including a luminance recording circuit 15 and chrominance recording circuit 16. Left-and right-channel audio signals are applied to terminals 5 and 6, respectively, of the left-and right-channel audio recording circuits and a video signal is applied to terminal 7. The video input to the terminal 7 is a standard color television signal which is separated in frequency between luminance and chrominance components in respective recording circuits 15 and 16. In the luminance recording circuit 15 the input signal is modulated in frequency upon a carrier with such a frequency shift that the sync tip level corresponds to 3.4 MHz and the white peak level corresponds to 4.4 MHz. The luminance input is modulated upon a 629 kHz color subcarrier and phase-shifted for purposes of minimizing crosstalk in a manner known in the art. The frequency-division multiplexed luminance and chrominance signals are fed to video heads H_1 and H_2 by way of rotary transformers 18 and 19.

The left-channel sound recording circuit 67L comprises a set of four frequency modulators 8, 9, 10 and 11, and an electronic rotary switch or analog multiplexer 12. The left-channel audio signal is modulated in frequency with a frequency shift of ± 100 kHz by frequency modulators 8, 9, 10 and 11 upon carriers having frequencies f_1 , f_3 , f_4 and f_2 , respectively, to generate a set of four frequency-modulated outputs A_1 , A_2 , A_3 and A_4 . In a preferred embodiment, frequencies f_1 , f_2 , f_3 and f_4 are 1.1 MHz, 1.35 MHz, 1.6 MHz and 1.85 MHz, respectively. A drum pulse generator 17 generates a pulse in response to each revolution of the rotary drum 1 and feeds the pulse to rotary switch 12 to sequentially couple the FM audio signals A_1 , A_2 , A_3 , A_4 to a recording amplifier 68 in response to each transducer head being shifted from one track to the next.

As shown in Fig. 4, the left-and right-channel FM audio signals span a range of frequencies lying between the frequency-modulated luminance signal FM-Y on the higher frequency end of the spectrum and the chrominance signal C on the lower frequency end.

Amplifier 68 strengthens the input signals to a level sufficient to magnetize the recording medium to saturation. Similarly, the frequency-modulated luminance signal is amplified to a level sufficient to cause magnetic saturation. The chrominance signal is, however, amplified to a level insufficient to cause saturation.

As will be described later, FM audio signals are

recorded prior to the recording of video signals. Because of the high amplification and long wavelengths relative to the luminance signal, the audio signals are recorded in a region as deep as 1 micrometer from the surface. Due to the short wavelengths in comparison with the FM audio signals, the luminance signal magnetizes a shallower region of the tape which is typically 0.2 to 0.3 micrometers from the surface. Although the chrominance signal has shorter wavelengths than the FM audio signals, the below-the-saturation level amplification does not permit it to reach the deeper area. Thus, video components are recorded in a shallower region of the magnetic tape and audio signals are buried underneath. The audio and video heads are arranged so that audio and video signals are almost simultaneously but respectively recorded on adjacent tracks with the video track being displaced by one track pitch in the direction of tape motion from the audio track. In the illustrated embodiment, the track pitch equals the track width.

Audio signal is recorded as follows. For convenience it is assumed that audio recording begins with signal A_1 through head HA_1 . Audio head HA_1 scans a track t_1 , while video head H_1 scans a track t_0 displaced one track from t_1 with no guard band in the direction of tape motion, Fig. 5. Signal A_2 is recorded on track t_2 through head HA_2 and video signal is recorded through head H_2 on the upper area of track t_1 below which audio signal A_1 is recorded.

Likewise, signal A_3 is recorded on track t_3 through head HA_1 and video signal is recorded on the upper area of track t_2 overlapping signal A_2 , and signal A_4 is recorded on track t_4 through head HA_2 while video signal is recorded on the upper area of track t_3 overlapping signal A_3 . In this way audio signals A_1 to A_4 are recorded on tracks t_1 to t_4 , respectively, while video signal is recorded on the upper regions of tracks t_0 to t_3 . This track pattern recyclically repeats during subsequent recording. It will be noted that there is a frequency difference of 0.5 MHz on one side of each track and a difference of 0.25 MHz on the other side. The track-to-track frequency variation is utilized advantageously for both audio and video tracking control purposes during playback operation which will be described hereinbelow with reference to Fig. 6. In Fig. 5, illustrated at TA and TC are portions of the tape which are used for recording audio signals with stationary heads in conventional manner and recording control signals, respectively.

In Fig. 6, an audio reproduction circuit is illustrated. Because of the difference in azimuth between heads HA_1 and HA_2 , audio signals A_1 and A_3 are exclusively detected by head HA_1 and signals A_2 and A_4 are exclusively detected by head HA_2 . The detected audio signals are fed through rotary transformers 13, 14 and playback amplifiers 23, 24 to an electronic switch 35. Switch 35 alternately couples signals A_1 to A_4 to an output lead 35a in response to drum pulses. This switching eliminates the components which

briefly overlap between adjacent tracks due to the longer tape length that is wound on the circumference of rotary drum 1 than the period of one video field. The output of switch 35 is connected to a left-channel reproduction circuit 69L and to a right-channel reproduction circuit 69R.

The left-channel circuit 69L comprises a set of four band-pass filters 25, 26, 27 and 28 having passbands of frequencies f_1 , f_3 , f_4 and f_2 , respectively. These band-pass filters separate the sequential FM audio carriers into individual carrier components A_1 , A_2 , A_3 and A_4 which are demodulated by frequency demodulators 29, 30, 31, 32, respectively. The demodulated carrier components are sequentially combined by means of an electronic rotary switch 33 in response to drum pulses to recover the original audio signal which is fed to an output terminal 34.

The outputs of the band-pass filters 25, 26, 27, 28 are further applied to a tracking control circuit 100. The control circuit 100 includes an electronic switch 53 having four input terminals A, B, C, D and two output terminals E, F which are simultaneously respectively coupled to input terminals A and C and switched to terminals B and D in response to a drum pulse. The output terminals E and F of switch 53 are connected to level detectors 45 and 46, respectively. The output of detector 45 is connected to input terminals A and D of an electronic switch 48, the output of detector 46 being connected to input terminals B and C of switch 48. Switch 48 has output terminals E and F which are respectively coupled to terminals A and C and switched to terminals B and D in response to a switching pulse supplied from a T flip-flop 47 which takes its input from the drum pulse generator 17, whereby switch 48 remains switched to the A and B positions when tracks t_4 and t_1 are successively scanned and remains switched to the C and D positions when tracks t_2 and t_3 are successively scanned. The output terminals E and F of switch 48 are connected to the positive and negative inputs of a differential amplifier 51, respectively, the differential output of which is coupled to the audio head moving mechanism 52 on which the audio heads HA_1 and HA_2 are located.

The operation of the tracking control circuit 100 is as follows. Assume that head HA_2 is scanning track t_2 to provide signal A_2 which is detected by band-pass filter 26, recovered by frequency demodulator 30 and fed to output terminal 34. Due to crosstalk, head HA_2 picks up carrier components A_1 and A_3 from adjacent tracks t_1 and t_3 in addition to carrier component A_2 . Since these crosstalk components are undesirable for normal sound reproduction, only the desired component A_2 is allowed to pass through band-pass filter 26. However, the crosstalk components A_1 and A_3 are utilized for tracking control purposes as follows.

Crosstalk components A_1 and A_3 are coupled by switch 53 to level detectors 45 and 46, respectively, where their amplitudes are detected and fed to the input terminals of switch 48. The output terminals E and F of switch 48 are now in

connection to its A and C input terminals to apply signals A_1 and A_3 to the positive and negative inputs of differential amplifier 51 respectively to provide an error voltage to the moving mechanism 52.

The operation of the tracking control circuit 100 is as follows. It is assumed that audio heads tend to deviate in the direction of tape motion and head HA_2 is following track t_2 with a deviation toward track t_1 . In this instance, carriers A_1 and A_3 are the crosstalk components which are to be compared with each other and switches 53 and 48 are both in the A and C positions.

With the deviation toward track t_1 , carrier A_1 is stronger than carrier A_3 at the inputs to differential amplifier 51 and a positive error voltage is generated to cause the head moving mechanism 52 to move the head HA_2 away from track t_1 .

When audio head HA_1 takes turn to scan the next track t_3 , carriers A_2 and A_4 are the crosstalk components to be compared. Switch 53 is changed to positions B and D, while switch 48 stays in the A and C positions, so that carrier components A_2 and A_4 are applied to the positive and negative inputs of differential amplifier 51 respectively. With a head displacement toward track t_2 , signal A_2 is stronger than signal A_4 and a positive error voltage will be generated for correcting the path of head HA_1 in a direction away from track t_2 toward track t_3 .

When audio head HA_2 subsequently scans track t_4 , carrier components A_3 and A_1 become the crosstalk components and switch 53 returns to the A and B positions. Switch 48 is now changed to the B and D positions, whereby carrier components A_3 and A_1 are applied to the positive and negative inputs of differential amplifiers 51, respectively. With head HA_1 still tending to deviate in the direction of tape motion, i.e. toward track t_3 , signal A_3 is stronger than signal A_1 and a positive error signal will be developed.

When head HA_1 next scans track t_1' , carrier components A_4 and A_2 are the crosstalk components and switch 53 changes to positions B and D. Switch 48 stays in positions B and D to connect carrier components A_4 and A_2 to the positive and negative inputs of differential amplifier 51. With a displacement toward track t_4 , signal A_4 is greater than signal A_2 and a positive error signal is generated.

A negative error voltage will thus be generated during each track scan if the head deviates in a direction opposite to the direction of tape motion to reduce the displacement substantially to zero.

Fig. 7 is an illustration of a video reproduction circuit according to the present invention. This circuit comprises an electronic switch 40 having two inputs coupled respectively to video heads H_1 and H_2 through playback amplifiers 38 and 39 to alternately couple the video signals to an output terminal 40a. The output terminal 40a is connected to luminance and chrominance reproduction circuits 41 and 42 of a well known design for converting the frequency-divided luminance and chrominance signals to the original signals. The

output terminal 40a is also connected to band-pass filters 54, 55, 56 and 57 which are tuned to frequencies f_1, f_3, f_4, f_2 , respectively, for separating FM audio carrier components, which are detected by video heads together with the video signals, into different frequency components A_1, A_2, A_3, A_4 . The circuitry includes a video tracking control circuit 200 identical to the audio tracking control circuit 100 used in the circuitry of Fig. 6. The separated audio signals are fed to the video tracking control circuit 200 to effect tracking control on the video heads by deriving an error voltage in the same manner as in circuit 100 and applying it to a video head moving mechanism 66 on which the video heads are mounted.

Fig. 8A illustrates one example of the right-channel sound recording circuit 67R. In this example, the circuit 67R comprises a single frequency modulator 71 as opposed to the multiple frequency modulator configuration of the left-channel sound recording circuit 67L. The right-channel signal is modulated upon a carrier having a frequency f_5 which is typically 2.1 MHz. The frequency-modulated right-channel signal, designated A_5 , is combined in the amplifier 68 with the left-channel signal and fed through rotary transformers 13, 14 to audio heads HA_1 and HA_2 .

Fig. 8B shows one example of the right-channel reproduction circuit 69R which is used in conjunction with the circuit of Fig. 8A. In Fig. 8B, the circuit 69R comprises a band-pass filter 72 and a frequency demodulator 73 connected in a series circuit. The band-pass filter 72 is tuned to frequency f_5 which is 2.1 MHz to pass the signal A_5 which has been derived from the frequency modulator 71, Fig. 8A.

Fig. 9A is an illustration of another example of the right-channel sound recording circuit 6 of Fig. 3. This circuit is generally similar to the left-channel sound recording circuit 67L in that it comprises frequency modulators 74, 75, 76 and 77 which frequency-modulate carriers having frequencies f_5 (=2.1 MHz), f_7 (=2.6 MHz), f_8 (=2.85 MHz) and f_6 (=2.35 MHz), respectively, with the right-channel audio signal to generate frequency-modulated audio signals A_5, A_6, A_7 and A_8 . These frequency-modulated audio signals are fed to respective input terminals of an electronic rotary switch 78 to multiplex the input signals in sequence in response to drum pulses applied thereto. Thus, a first set of frequency-modulated audio signals having carrier frequencies f_1 to f_4 and a second set of frequency-modulated audio signals having carrier frequencies f_5 to f_8 are generated in the circuits 67L and 67R, Fig. 3, and combined in the recording amplifier 68 and recorded as a frequency-division multiplexed signal on magnetic recording medium. The frequency-modulated audio signals A_5 — A_8 appear in the frequency spectrum as indicated by dotted lines in Fig. 5.

The right-channel sound reproduction circuit 69R shown in Fig. 9B is used in the circuitry of Fig. 6 in conjunction with the recording circuit of Fig. 9A. The F4 audio signal with sequentially varying

frequencies is separated into individual components A_5 to A_8 by band-pass filters 79, 80, 81 and 82 tuned to frequencies f_5 , f_7 , f_8 , f_6 respectively, fed to frequency demodulators 83, 84, 85, 86 respectively and applied to a rotary switch 87 to be multiplexed into a sequential signal for application to the audio heads. Preferably, the outputs of band-pass filters 79—82 are fed to a second audio tracking control circuit 300 identical to the first audio tracking circuit 100. The arrangements of Figs. 9A and 9B are preferred when a higher degree of precision control and a higher degree of crosstalk separation are desired.

In Fig. 10 is shown another embodiment of the present invention in which parts corresponding to those in Fig. 3 are indicated by the same numerals as used in Fig. 3. In this embodiment, left- and right-channel audio signals L and R are recorded in a time-division multiplexed format along each track by time-compressing the both channel signals into one-half the original and recording the time-compressed right-channel signal CR onto a first half section of each track and recording the time-compressed left-channel signal CL onto a second half section of the track as schematically shown in Fig. 11.

To this end, the sound recording circuit includes a left-channel time compressor 90 which typically comprises a charge-coupled device and a control circuit that determines the clock rate of the CCD. The time-compressed left-channel signal CL is delayed by one-half the field interval in a delay circuit 91, the delayed signal DCL being coupled to an input of a switch 92 through which it is applied to frequency modulators 8 to 11. The right-channel signal R is fed to a right-channel time compressor 93 identical in construction to time compressor 90, the compressed right-channel signal CR being coupled to a second input of switch 92 and through which it is fed to the frequency modulators 8—11. The switch 92 alternately couples the signals DCL and CR alternately in response to a control signal applied thereto from a frequency doubler 94. Frequency doubler 94 comprises a 1/2 field delay circuit 95, a NOR gate 96 and an AND gate 97. The 1/2 field delay 103 introduces a delay time of one-half the field interval to drum pulses, the delayed drum pulses being fed to NOR gate 96 and AND gate 97 to each of which nondelayed drum pulse is also applied. The outputs of the gates 96 and 97 are combined to produce a pulse train signal having a repetition frequency twice as high as the drum pulse. The output of switch 92 is a sequence of compressed right- and left-channel signals CR and CL which are frequency modulated upon respective carriers in a manner as described in Fig. 3 and fed to switch 12 which sequentially couples the frequency-modulated signals to amplifier 68 and thence to audio heads HA_1 , HA_2 in response to each head being shifted from one track to the next. Therefore, left-channel signal components L_1 and L_2 are compressed respectively into components CL_1 and CL_2 , as shown at Fig. 10A, and delayed for 1/2 field interval in

succession so that delayed components CL_1 and CL_2 appear respectively in the second half periods of the field intervals occupied by the original left-channel components L_2 and L_3 and the right-channel components R_1 and R_2 appear in the first half period of the same field intervals.

Since the frequency modulators 8—11 are commonly shared by left- and right-channel signals, the frequency modulated components fall in the same frequency range occupied by the frequency modulated signals A_1 to A_4 of the embodiment of Fig. 3. Thus, the time-division multiplexing relaxes the bandwidth requirement for audio signals.

Fig. 12 is an illustration of the circuitry for reproducing the time-compressed audio signals, in which parts corresponding to those in Fig. 6 are marked with the same reference numerals as used therein and the description thereof being omitted for brevity. Additionally included in this circuitry is a demultiplexing switch 101 which is connected to the output of switch 33. The sequentially combined time-compressed signals CR plus CL are separated into different frequency components by band-pass filters 25—28 and applied respectively to frequency demodulators 29—32 on the one hand and to tracking control circuit 100 on the other. The frequency demodulated signals right- and left-channel compressed signals CR, CL derived from each track are selected by switch 33 and fed to switch 101. This switch is operated in response to an output signal from a frequency doubler 102 identical to that shown in Fig. 10. The compressed signals CL and CR on each track are separated from each other by switch 101 and respectively applied to a left-channel time expander 103 and to a 1/2 field delay 104 which delays the compressed right-channel signal for one-half the field interval and feeds its output to a right-channel time expander 105. Each of the time expanders comprises a charge couple device and a control circuit that operates in a manner inverse to the time compressors 90 and 93.

As shown in Fig. 12A, the time compressed right channel components CR_1 and CR_2 are thus delayed and time-expanded so that component CR_1 appears in the same interval as the original time compressed components CR_2 and CL_2 appear, and the time-compressed left-channel components CL_1 and CL_2 are expanded to appear in the same interval in which the expanded right-channel components CR_1 and CR_2 appear.

In the foregoing description, the tracking control circuit utilizes the different frequency components which are derived from opposite sides of a given track. The present invention could be modified to allow the transducer head to follow a desired track by utilizing the signal derived from an adjacent track scanned immediately prior to the desired track. This is particularly advantageous for tracking control during recording modes.

Claims

1. A helical scan tape recording and/or repro-

frequencies is separated into individual components A_5 to A_8 by band-pass filters 79, 80, 81 and 82 tuned to frequencies f_5 , f_7 , f_8 , f_6 respectively, fed to frequency demodulators 83, 84, 85, 86 respectively and applied to a rotary switch 87 to be multiplexed into a sequential signal for application to the audio heads. Preferably, the outputs of band-pass filters 79—82 are fed to a second audio tracking control circuit 300 identical to the first audio tracking circuit 100. The arrangements of Figs. 9A and 9B are preferred when a higher degree of precision control and a higher degree of crosstalk separation are desired.

In Fig. 10 is shown another embodiment of the present invention in which parts corresponding to those in Fig. 3 are indicated by the same numerals as used in Fig. 3. In this embodiment, left- and right-channel audio signals L and R are recorded in a time-division multiplexed format along each track by time-compressing the both channel signals into one-half the original and recording the time-compressed right-channel signal CR onto a first half section of each track and recording the time-compressed left-channel signal CL onto a second half section of the track as schematically shown in Fig. 11.

To this end, the sound recording circuit includes a left-channel time compressor 90 which typically comprises a charge-coupled device and a control circuit that determines the clock rate of the CCD. The time-compressed left-channel signal CL is delayed by one-half the field interval in a delay circuit 91, the delayed signal DCL being coupled to an input of a switch 92 through which it is applied to frequency modulators 8 to 11. The right-channel signal R is fed to a right-channel time compressor 93 identical in construction to time compressor 90, the compressed right-channel signal CR being coupled to a second input of switch 92 and through which it is fed to the frequency modulators 8—11. The switch 92 alternately couples the signals DCL and CR alternately in response to a control signal applied thereto from a frequency doubler 94. Frequency doubler 94 comprises a 1/2 field delay circuit 95, a NOR gate 96 and an AND gate 97. The 1/2 field delay 103 introduces a delay time of one-half the field interval to drum pulses, the delayed drum pulses being fed to NOR gate 96 and AND gate 97 to each of which nondelayed drum pulse is also applied. The outputs of the gates 96 and 97 are combined to produce a pulse train signal having a repetition frequency twice as high as the drum pulse. The output of switch 92 is a sequence of compressed right- and left-channel signals CR and CL which are frequency modulated upon respective carriers in a manner as described in Fig. 3 and fed to switch 12 which sequentially couples the frequency-modulated signals to amplifier 68 and thence to audio heads HA_1 , HA_2 in response to each head being shifted from one track to the next. Therefore, left-channel signal components L_1 and L_2 are compressed respectively into components CL_1 and CL_2 , as shown at Fig. 10A, and delayed for 1/2 field interval in

succession so that delayed components CL_1 and CL_2 appear respectively in the second half periods of the field intervals occupied by the original left-channel components L_2 and L_3 and the right-channel components R_1 and R_2 appear in the first half period of the same field intervals.

Since the frequency modulators 8—11 are commonly shared by left- and right-channel signals, the frequency modulated components fall in the same frequency range occupied by the frequency modulated signals A_1 to A_4 of the embodiment of Fig. 3. Thus, the time-division multiplexing relaxes the bandwidth requirement for audio signals.

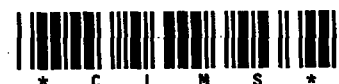
Fig. 12 is an illustration of the circuitry for reproducing the time-compressed audio signals, in which parts corresponding to those in Fig. 6 are marked with the same reference numerals as used therein and the description thereof being omitted for brevity. Additionally included in this circuitry is a demultiplexing switch 101 which is connected to the output of switch 33. The sequentially combined time-compressed signals CR plus CL are separated into different frequency components by band-pass filters 25—28 and applied respectively to frequency demodulators 29—32 on the one hand and to tracking control circuit 100 on the other. The frequency demodulated signals right- and left-channel compressed signals CR, CL derived from each track are selected by switch 33 and fed to switch 101. This switch is operated in response to an output signal from a frequency doubler 102 identical to that shown in Fig. 10. The compressed signals CL and CR on each track are separated from each other by switch 101 and respectively applied to a left-channel time expander 103 and to a 1/2 field delay 104 which delays the compressed right-channel signal for one-half the field interval and feeds its output to a right-channel time expander 105. Each of the time expanders comprises a charge couple device and a control circuit that operates in a manner inverse to the time compressors 90 and 93.

As shown in Fig. 12A, the time compressed right channel components CR_1 and CR_2 are thus delayed and time-expanded so that component CR_1 appears in the same interval as the original time compressed components CR_2 and CL_2 appear, and the time-compressed left-channel components CL_1 and CL_2 are expanded to appear in the same interval in which the expanded right-channel components CR_1 and CR_2 appear.

In the foregoing description, the tracking control circuit utilizes the different frequency components which are derived from opposite sides of a given track. The present invention could be modified to allow the transducer head to follow a desired track by utilizing the signal derived from an adjacent track scanned immediately prior to the desired track. This is particularly advantageous for tracking control during recording modes.

Claims

1. A helical scan tape recording and/or repro-



ducing apparatus having a transducer head proximate to a recording medium which is arranged to move so that the transducer head successively scans along each of a plurality of parallel tracks, wherein an information signal is modulated in frequency upon a carrier prior to recording in said medium through said head, characterized in that said information signal is modulated in frequency upon a plurality of carriers of different frequencies to generate a series of information bearing frequency-modulated carriers (A_1 — A_n) each of which occurs in synchronism with the shifting of said transducer head (HA_1 , HA_2) from one track to the next, and said series of frequency-modulated carriers is applied to said head to record each of said carriers on a different track so that there is a difference in frequency between carriers recorded on any pair of adjacent tracks.

2. A recording and/or reproducing apparatus as claimed in claim 1, characterized in that said information signal is modulated upon first, second, third and fourth different frequency carriers to generate a series of first, second, third and fourth frequency-modulated carriers in synchronism with the shifting of said head from one track to the next.

3. A recording and/or reproducing apparatus as claimed in claim 1 or 2, further characterized by means (25—28) for separating a series of frequency-modulated carriers detected by said head into different frequency components, and head control means (100, 52) for controlling said head in a direction transverse to the direction of track scan in response to at least one of the different frequency components which is detected from one side of the track which is being scanned by said head.

4. A recording and/or reproducing apparatus as claimed in claim 1 or 2, further characterized by means (25—28) for separating a series of frequency-modulated carriers detected by said head into different frequency components and head control means (100, 52) for controlling said head in a direction transverse to the direction of track scan in response to those different frequency components which are detected from opposite sides of the track which is being scanned by said head.

5. A recording and/or reproducing apparatus as claimed in claim 4, characterized in that said head control means (100, 52) comprises means (100) for detecting the difference in amplitude between said different frequency components detected from said opposite sides and generating a tracking control signal indicative of the amount and direction of deviation of said head from said scanned track, and means (52) for moving said head in response to said tracking control signal to reduce said deviation.

6. A recording and/or reproducing apparatus as claimed in claim 4, characterized in that said separating means comprises first, second, third and fourth band-pass filters (25—28) for separating the series of frequency-modulated carriers into first, second, third and fourth different fre-

quency components, and in that said head control means comprises a differential amplifier (51) having a first and a second input terminal for generating said tracking control signal, and coupling means (53, 48) for coupling said first and third signals to said first and second input terminals respectively when said head scans a first track, coupling said second and fourth signals to said first and second terminals respectively when said head subsequently scans a second track, coupling said first and third signals to said second and first input terminals respectively when said head subsequently scans a third track and coupling said second and fourth signals to said second and first input terminals respectively when said head subsequently scans a fourth track.

7. A recording and/or reproducing apparatus as claimed in any one of the preceding claims, wherein said information signal is an audio signal and a second transducer head (H_1 , H_2) is provided so that the second head follows each of said tracks in which said frequency-modulated carriers are recorded, and a video signal is modulated upon a carrier having a frequency differing from the frequencies of said carriers modulated with said audio signal and applied to said second head so that the modulated video signal is recorded in an area overlying the area in which said audio signal is recorded, characterized by means (54—57) for separating a series of frequency-modulated carriers detected by said second head into different frequency components, and head control means (200) for controlling said second head in a direction transverse to the direction of track scan in response to those different frequency components which are detected from opposite sides of the track which is being scanned by said second head.

8. An apparatus for reproducing a record medium having a plurality of successively arranged tracks in each of which is recorded a carrier modulated in frequency with an information signal in a manner as claimed in claim 1 or 2, comprising a transducer head arranged to successively scan each of said tracks for sequentially detecting said stored carriers, means (100) for detecting the deviation of said head from the intended track to generate a tracking control signal and means (52) for moving said head in response to said tracking control signal, characterized by means (25—28) for separating the detected carriers into different frequency components, and said deviation detecting means (100) is responsive to those different frequency components which are derived from opposite sides of the track which is being scanned by said head.

9. A reproducing apparatus as claimed in claim 8, characterized in that said deviation detecting means comprises a differential amplifier (51) for detecting the difference in amplitude between said different frequency components derived from said opposed sides and generating a tracking control signal indicative of the amount and direction of deviation of said head from said



scanned track, and means (52) for moving said head in response to said tracking control signal to reduce said deviation.

10. A reproducing apparatus as claimed in claim 9, characterized in that said separating means comprises first, second, third and fourth band-pass filter means (25—28) for separating a series of carriers detected by said head into first, second third and fourth different frequency components, and in that said deviation detecting means (100) comprises coupling means (53, 48) for coupling said first and third components to the first and second input terminals of said differential amplifier (51) respectively when said head scans a first track, coupling said second and fourth components to said first and second terminals respectively when said head subsequently scans a second track, coupling said first and third components to said second and first input terminals respectively when said head subsequently scans a third track and coupling said second and fourth components to said second and first input terminals respectively when said head subsequently scans a fourth track.

Patentansprüche

1. Schrägspurbandaufzeichnungs-und/oder Wiedergabegerät mit einem Aufzeichnungs-und Wiedergabekopf, der an einem Aufzeichnungsmedium anliegt, welches bewegbar angeordnet ist, so daß der Aufzeichnungs-und Wiedergabekopf aufeinanderfolgend jede von einer Vielzahl von parallelen Spuren abtastet, wobei ein Informationssignal vor dem Aufzeichnen auf das Medium durch den Kopf auf einen Träger frequenzmoduliert ist, dadurch gekennzeichnet, daß das Informationssignal auf eine Vielzahl von Trägern unterschiedlicher Frequenz frequenzmoduliert ist, um eine Reihe von Information tragenden frequenzmodulierten Trägern (A_1 — A_n) zu erzeugen, von denen jeder synchron mit der Verschiebung des Aufzeichnungs-und Wiedergabekopfes (HA_1 , HA_2) von einer Spur zur nächsten auftritt, und daß die Reihe von frequenzmodulierten Trägern an den Kopf angelegt ist, um jeden der Träger auf eine unterschiedlichen Spur aufzuzeichnen, so daß eine Frequenzdifferenz zwischen Trägern vorliegt, die auf jedem Paar von benachbarten Spuren aufgezeichnet sind.

2. Aufzeichnungs-und/oder Wiedergabegerät nach Anspruch 1, dadurch gekennzeichnet, daß das Informationssignal auf erste, zweite, dritter und vierte Träger unterschiedlichen Frequenz aufmoduliert ist, um eine Reihe von ersten, zweiten, dritten und vierten frequenzmodulierten Trägern synchron mit der Verschiebung des Kopfes von einer Spur zur nächsten zu erzeugen.

3. Aufzeichnungs-und/oder Wiedergabegerät nach Anspruch 1 oder 2, gekennzeichnet durch Mittel (25—28) zum Trennen einer Reihe vom Kopf erfaßter frequenzmodulierter Träger in unterschiedliche Frequenzkomponenten, und durch Kopfsteuermittel (100, 52) zum Steuern des Kopfes in einer Richtung quer zu der Richtung der

Spurabtastung in Abhängigkeit von wenigstens einer von den unterschiedlichen Frequenzkomponenten, welche von einer Seite der vom Kopf abgetasteten Spur erfaßt ist.

4. Aufzeichnungs-und/oder Wiedergabegerät nach Anspruch 1 oder 2, gekennzeichnet durch Mittel (25—28) zum Trennen einer Reihe vom Kopf erfaßter frequenzmodulierter Träger in unterschiedliche Frequenzkomponenten und durch Steuermittel (100, 52) zum Steuern des Kopfes in einer Richtung quer zu der Richtung der Spurabtastung in Abhängigkeit von solchen unterschiedlichen Frequenzkomponenten, welche von entgegengesetzten Seiten der vom Kopf abgetasteten Spur erfaßt sind.

5. Aufzeichnungs-und/oder Wiedergabegerät nach Anspruch 4, dadurch gekennzeichnet, daß die Kopfsteuermittel (100, 52) Mittel (100) aufweisen zum Erfassen von Amplitudendifferenzen zwischen den von entgegengesetzten Seiten erfaßten unterschiedlichen Frequenzkomponenten und zum Erzeugen von Spurführungssteuersignalen, die die Größe und die Richtung der Abweichung des Kopfes von der abgetasteten Spur anzeigen, und Mittel (52) aufweisen, zum Bewegen des Kopfes in Abhängigkeit von dem Spurführungssteuersignal zum Vermindern der Abweichung.

6. Aufzeichnung-und/oder Wiedergabegerät nach Anspruch 4, dadurch gekennzeichnet, daß die Trennmittel erste, zweite, dritte und vierte Bandpaßfilter (25—28) umfassen zum Trennen der Reihe von frequenzmodulierten Trägern in erste, zweite, dritte und vierte unterschiedliche Frequenzkomponenten, und daß die Kopfsteuermittel einen Differenzverstärker (51) mit einem ersten und einem zweiten Eingang zu Erzeugung des Spurführungssteuersignals und Verbindungsmittel (53, 48) aufweisen zum Anlegen der ersten und dritten Signale an den ersten bzw. zweiten Eingang, wenn der Kopf eine erste Spur abtastet, zum Anlegen der zweiten und vierten Signale an den ersten bzw. zweiten Eingang, wenn der Kopf nachfolgend eine zweite Spur abtastet, zum Anlegen der ersten und dritten Signale an den zweiten bzw. ersten Eingang, wenn der Kopf nachfolgende eine dritte Spur abtastet, und zum Anlegen der zweiten und vierten Signale an den zweiten bzw. ersten Eingang, wenn der Kopf nachfolgend eine vierte Spur abtastet.

7. Aufzeichnungs-und/oder Wiedergabegerät nach einem der vorhergehenden Ansprüche, wobei das Informationssignal ein Audiosignal ist und ein zweiter Aufzeichnungs-und Wiedergabekopf (H_1 , H_2) vorgesehen ist, so daß der zweite Kopf jeder von den Spuren folgt, in denen die frequenzmodulierten Träger aufgezeichnet sind, und ein Videosignal auf einen Träger aufmoduliert ist, der eine Frequenz unterschiedlich von den Frequenzen der mit dem Audiosignal modulierten Trägern aufweist, und an den zweiten Kopf angelegt ist, so daß das Modulierte Videosignal in einem Bereich aufgezeichnet ist, der den Bereich überdeckt, in dem die Audiosignale aufgezeichnet sind, gekennzeichnet durch Mittel (54, 57) zum



Trennen einer Reihe vom zweiten Kopf erfaßter frequenzmodulierter Träger in unterschiedliche Frequenzkomponenten und durch Kopfsteuermit-
tel (200) zum Steuern des zweiten Kopfes in einer Richtung quer zu der Richtung der Spurbastung in Abhängigkeit von solchen unterschiedlichen Frequenzkomponenten, welche von entgegengesetzten Seiten der vom zweiten Kopf abgetasteten Spur erfaßt sind.

8. Vorrichtung zum Reproduzieren eines Aufzeichnungsmediums mit einer Vielzahl von aufeinanderfolgend angeordneten Spuren, in jeder von denen ein Träger aufgezeichnet ist, auf den ein Informationssignal in der Art nach Anspruch 1 oder 2 frequenzmoduliert ist, mit einem Aufzeichnungs- und Wiedergabekopf, der zum aufeinanderfolgenden Abtasten jeder der Spuren angeordnet ist zum nacheinanderfolgenden Erfassen der gespeicherten Träger, mit Mitteln (100) zum Erfassen der Abweichung des Kopfes von der vorgesehenen Spur, um ein Spurführungssteuersignal zu erzeugen, und mit Mitteln (52) zum Bewegen des Kopfes in Abhängigkeit von dem Spurführungssteuersignal, gekennzeichnet durch Mittel (25—28) zum Trennen der erfaßten Träger in unterschiedliche Frequenzkomponenten und dadurch, daß die Abweichungserfassungsmittel (100) von solchen unterschiedlichen Frequenzkomponenten Abhängig sind, die abgeleitet sind von entgegengesetzten Seiten der Spur, welche vom Kopf abgetastet ist.

9. Reproduziergerät nach Anspruch 8, dadurch gekennzeichnet, daß die Abweichungserfassungsmittel einen Differenzverstärker (51) aufweisen zum Erfassen der Amplituden differenz zwischen den unterschiedlichen Frequenzkomponenten, die abgeleitet sind von den entgegengesetzten Seiten, und zum Erzeugen eines Spurführungssteuersignals, das die Größe und Richtung der Abweichung des Kopfes von der abgetasteten Spur anzeigt, und Mittel (52) aufweisen zum Bewegen des Kopfes in Abhängigkeit von dem Spurführungssteuersignal, um die Abweichung zu vermindern.

10. Reproduziergerät nach Anspruch 9, dadurch gekennzeichnet, daß die Trennmittel erste, zweite, dritte und vierte Bandpaßfiltermittel (25—28) aufweisen zum Trennen einer Reihe von durch den Kopf erfaßten Trägern in erste, zweite, dritte und vierte unterschiedliche Frequenzkomponenten, und daß die Abweichungserfassungsmittel (100) Verbindungsmittel (53, 48) aufweisen zum Anlegen der ersten und dritten Komponente an den ersten bzw. zweiten Eingang des Differenzverstärkers (51), wenn der Kopf eine erste Spur abtastet, zum Anlegen der zweiten und vierten Komponente an den ersten bzw. zweiten Eingang, wenn der Kopf nachfolgend eine zweite Spur abtastet, zum Anlegen der ersten und dritten Komponente an den zweiten bzw. ersten Eingang, wenn der Kopf nachfolgend eine dritte Spur abtastet, und zum Anlegen der zweiten und vierten Komponente an den zweiten bzw. ersten Eingang, wenn der Kopf nachfolgend eine vierte Spur abtastet.

Revendications

1. Appareil d'enregistrement et/ou de reproduction à bande à balayage hélicoïdal ayant une tête de transducteur voisine d'un support d'enregistrement qui est disposé de façon à se déplacer de telle sorte que la tête de transducteur balaye successivement chacune d'une pluralité de pistes parallèles, dans lequel un signal d'information est modulé en fréquence sur une porteuse avant de l'enregistrer dans ledite support par l'intermédiaire de ladite tête, caractérisé en ce que ledit signal d'information est modulé en fréquence sur une pluralité de porteuses de fréquences différentes de façon à générer une série de porteuses (A_1 à A_n) modulées en fréquence et portant des informations, chacune d'entre elles apparaissant en synchronisme avec le passage de ladite tête de transducteur (HA_1 , HA_n) d'une piste à la suivante, et en ce que ladite série de porteuses modulées en fréquence est appliquée à ladite tête de façon à enregistrer chacune desdites porteuses sur une piste différente de telle sorte qu'il y aie une différence de fréquence entre les porteuses enregistrées sur n'importe quelle paire de pistes adjacentes.

2. Appareil d'enregistrement et/ou de reproduction selon la revendication 1, caractérisé en ce que ledit signal d'information est modulé sur des première, deuxième, troisième et quatrième porteuses de fréquences différentes de façon à générer une série de première, deuxième, troisième et quatrième porteuses modulées en fréquence en synchronisme avec le passage de ladite tête d'une piste à la suivante.

3. Appareil d'enregistrement et/ou de reproduction selon la revendication 1 ou 2, caractérisé de plus par des moyens (25 à 28) pour séparer une série de porteuses modulées en fréquence détectées par ladite tête en différentes composantes de fréquence, et par des moyens de commande de tête (100, 52) pour commander ladite tête dans une direction transversale à la direction de balayage de la piste en réponse à au moins l'une des différentes composantes de fréquence que est détectée à partie d'un côté de la piste qui est en train d'être balayée par ladite tête.

4. Appareil d'enregistrement et/ou de reproduction selon la revendication 1 ou 2, caractérisé de plus par des moyens (25, 28) pour séparer une série de porteuses modulées en fréquence détectées par ladite tête en différentes composantes de fréquence, et par des moyens de commande de tête (100, 52) pour commander ladite tête dans une direction transversale à la direction de balayage de la piste en réponse à ces différentes composantes de fréquence qui sont détectées à partir des côtés opposés de la piste que est en train d'être balayée par ladite tête.

5. Appareil d'enregistrement et/ou de reproduction selon la revendication 4, caractérisé en ce que lesdits moyens de commande de tête (100, 52) comportent des moyens (100) pour détecter la différence d'amplitude entre lesdites différentes composantes de fréquences détectées à partir



desdits côtés opposés et pour générer un signal de commande d'alignement indicatif de l'amplitude et de la direction de l'écart de ladite tête par rapport à ladite piste balayée, et des moyens (52) pour déplacer ladite tête en réponse audit signal de commande d'alignement afin de réduire ledit écart.

6. Appareil d'enregistrement et/ou de reproduction selon la revendication 4, caractérisé en ce que lesdits moyens de séparation comportent des premier, deuxième, troisième et quatrième filtres passe-bande (25 à 28) pour séparer la série de porteuses modulées en fréquence en première, deuxième, troisième et quatrième composantes de fréquence différentes, et en ce que lesdits moyens de commande de tête comportent un amplificateur différentiel (51) ayant une première et une deuxième bornes d'entrée pour générer ledit signal de commande d'alignement, et des moyens de couplage (53, 48) pour coupler lesdits premier et troisième signaux auxdites première et deuxième bornes d'entrée, respectivement, lorsque ladite tête balaye une première piste, pour coupler lesdits deuxième et quatrième signaux auxdites première et deuxième bornes, respectivement, lorsque ladite tête balaye ensuite une deuxième piste, pour coupler lesdits premier et troisième signaux auxdites deuxième et première bornes d'entrée, respectivement, lorsque ladite tête balaye ensuite une troisième piste, et pour coupler lesdits deuxième et quatrième signaux auxdites deuxième et première bornes d'entrée, respectivement, lorsque ladite tête balaye ensuite une quatrième piste.

7. Appareil d'enregistrement et/ou de reproduction selon l'une quelconque des revendications précédentes, dans lequel ledit signal d'information est un signal audio et une deuxième tête de transducteur (H_1 , H_2) est disposée de telle sorte que la deuxième tête suive chacune desdites pistes dans lesquelles lesdites porteuses modulées en fréquence sont enregistrées, et un signal vidéo est modulé sur une porteuse ayant une fréquence différant des fréquences desdites porteuses modulées par ledit signal audio et appliquée à ladite deuxième tête de telle sorte que le signal vidéo modulé soit enregistré dans une zone recouvrant la zone dans laquelle est enregistré ledit signal audio, caractérisé par des moyens (54 à 57) pour séparer une série de porteuses modulées en fréquence détectées par ladite deuxième tête en différentes composantes de fréquence, et par des moyens de commande de tête (200) pour commander ladite deuxième tête dans une direction transversale à la direction de balayage de la piste en réponse à ces différentes composantes de fréquence qui sont détectées à partir des côtés opposés de la piste qui est en train d'être balayée par ladite deuxième tête.

8. Appareil pour lire un support d'enregistrement ayant une pluralité de pistes disposées successivement dans chacune desquelles est enregistrée une porteuse modulée en fréquence par un signal d'information d'une façon telle que celle revendiquée dans la revendication 1 ou 2, comportant une tête de transducteur disposée de façon à balayer successivement chacune desdites pistes pour détecter séquentiellement lesdites porteuses mémorisées, des moyens (100) pour détecter l'écart de ladite tête par rapport à la piste visée de façon à générer un signal de commande d'alignement et des moyens (52) pour déplacer ladite tête en réponse audit signal de commande d'alignement, caractérisé par des moyens (25 à 28) pour séparer les porteuses détectées en différentes composantes de fréquence, et en ce que lesdits moyens de détection de l'écart (100) réagissent à ces différentes composantes de fréquence que proviennent des côtés opposés de la piste qui est en train d'être balayée par ladite tête.

9. Appareil de reproduction selon la revendication 8, caractérisé en ce que lesdits moyens de détection de l'écart comportent un amplificateur différentiel (51) pour détecter la différence d'amplitude entre lesdites différentes composantes de fréquence provenant desdits côtés opposés et pour générer un signal de commande d'alignement indicatif de l'amplitude et de la direction de l'écart de ladite tête par rapport à ladite piste balayée, et des moyens (52) pour déplacer ladite tête en réponse audit signal de commande d'alignement afin de réduire ledit écart.

10. Appareil de reproduction selon la revendication 9, caractérisé en ce que lesdits moyens de séparation comportent des premiers, deuxième, troisième et quatrième moyens formant filtre passe-bande (25 à 28) pour séparer une série de porteuses détectées par ladite tête en première, deuxième, troisième et quatrième composantes de fréquence différentes, et en ce que lesdits moyens de détection de l'écart (100) comportent des moyens de couplage (53, 48) pour coupler lesdites première et troisième composantes aux première et deuxième bornes d'entrée dudit amplificateur différentiel (51), respectivement, lorsque ladite tête balaye une première piste, pour coupler lesdites deuxième et quatrième composantes auxdites première et deuxième bornes, respectivement, lorsque ladite tête balaye ensuite une deuxième piste, pour coupler lesdites première et troisième composantes auxdites deuxième et première bornes d'entrée, respectivement, lorsque ladite tête balaye ensuite une troisième piste et pour coupler lesdites deuxième et quatrième composantes auxdites deuxième et première bornes d'entrée, respectivement, lorsque lesdites têtes balayent ensuite une quatrième piste.

