

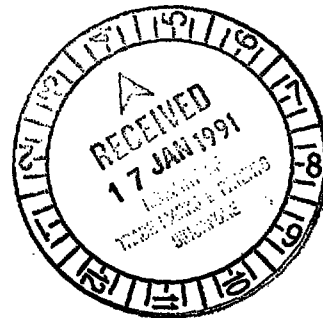
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I2000023 EP
HR/ep

TRANSCIVER

This invention relates to transceivers, and more particularly to a transceiver which is used to drive substantially square wave form signals having a high frequency onto a transmission line through a coupling transformer and of receiving through the transformer such high frequency signals transmitted by another such transceiver and a lower frequency signal having a quasi-sinusoidal wave form which signals are concurrently carried by the transmission line.

Transceivers which transmit and receive signals over a transmission line are well-known in the art. Where signals from two or more sources are to be transmitted, or carried concurrently over a single conductor, this is typically accomplished by some form of multiplexing or modulation system using a carrier signal.



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The environment in which the transceiver of this invention is to be used is a token passing local area network in which a limited number, sixty four being the maximum number in the preferred embodiment, communicate with one another over the network's transmission medium, a pair of coaxial cables. Communication between physical modules is at a frequency of 5.0 M Hz with the physical module of the network having the token at any given time and being enabled to transmit Manchester encoded binary data signals at a 5.0 M Hz frequency over the transmission medium to any one or all of the physical modules of the network. Each physical module must therefore have the capability of transmitting and of receiving data signals at the 5.0 M Hz frequency over both cables of the network medium.

In addition, there is a requirement, particularly where the network forms or constitutes, a control system for monitoring and controlling industrial and utility plants, to time the occurrence of events with a high degree of precision. In the environment in which the transceiver of this invention is utilized each physical module includes a timing subsystem. Selected ones of the timing subsystem of the physical modules prepare and transmit encoded timing information in the form of timing signals over the networks communication medium so that the timing subsystems of all physical modules can synchronize within a desired degree of accuracy, their internal senses of time with

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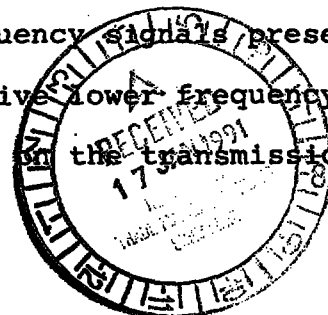
the internal sense of the timing subsystem of the physical module which timing subsystem is designated as the master.

In the preferred embodiment, the communication medium is dually redundant; i.e., it consists of two coaxial cables. In normal operation both cables carry the same data signals. The timing subsystem designated as a master is provided with a timing subsystem driver circuit which drives quasi-sinusoidal timing signals having a frequency of 12.5 K Hz onto one of the coaxial cables of the transmission medium. A second timing subsystem designated as a slave is also provided with such a driver circuit and applies timing signals onto the second coaxial cable of the dually redundant transmission medium. The length of the cables is limited to the length of substantially 1000 ft. in the preferred embodiment. For additional information on this environment, reference is made to the following concurrently filed patent applications:

1. Method and Apparatus for Synchronizing Timing Subsystems of the Physical Modules of a Local Area Network, EP application 85 , (I2000022), and
2. Driver Circuit, EP application , (I2000025).

It is therefore an object of this invention to provide a transceiver which can drive onto a transmission line through a coupling transformer high frequency substantially square wave signals and can receive from the transmission line through the coupling transformer any such high frequency signals present on the transmission line and also can receive lower frequency quasi-sinusoidal signals present concurrently on the transmission line.

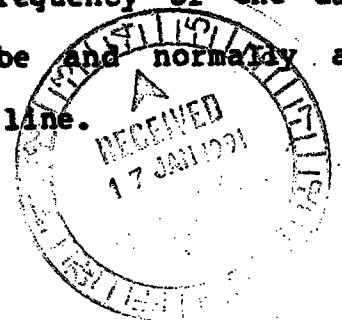
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This object is achieved by the invention as described in claim

1. The invention discloses a transceiver that can transmit and receive a first type of signal at one frequency over a single conductor and which can also receive a second type signal at a different frequency which is also concurrently carried by the single conductor in the absence of any modulators or demodulators, multiplexers or demultiplexers. Further details are characterized in the subclaims.

The transceiver of this invention applies to a transmission line input signals which signals are applied to the transmitter circuit of the transceiver. The input signals are Manchester encoded binary data signals of a given frequency. Signals carried by the transmission line are applied to a data signal receiver circuit of the transceiver. The data signal receiver circuit of the transceiver amplifies and reshapes any data signals present on the transmission line, or applied to the transmission line by the transceiver's transmitter circuit, and produces such data signals as the data signal output of the transceiver. Timing signals representing encoded timing data at a different frequency from that of the data signals also present on the transmission line are applied to a timing signal receiver of the transceiver which amplifies and squares the quasi-sinusoidal timing signals present on the transmission line and produces squared timing signals as the timing signal output of the transceiver. The timing signals in the preferred embodiment have a significantly lower frequency than the frequency of the data signals, and both types of signals can be and normally are present at the same time on the transmission line.



Other objects features and advantages of the invention will be readily apparent from the following description of a preferred embodiment thereof, taken in conjunction with the accompanying drawing, although variations and modifications may be affected without departing from the spirit and scope and of the novel concepts of the disclosure, and in which the sole figure is a circuit diagram of the transceiver of this invention.

In the drawing, transceiver 110 is operationally connected by coupling transformer 112 to 75 ohm coaxial cable 114, the transmission medium, or line, of a local area network. One end of transmission line winding 116 of transformer 112 is connected to the central conductor of cable 114 and the other end is connected to the cylindrical outer shield of cable 114. Cable 114 carries, or as applied to it, two signals, a Manchester encoded 5.0 M Hz substantially square wave signal sometimes identified as a data signal, or data signals; and a 12.5 k Hz quasi-or pseudo-, sinusoidal signal, sometimes hereafter referred to as a timing signal, or timing signals since the timing signals represent, in the preferred embodiment encoded timing information.

Transformer coupling between coaxial cable 114 and transceiver 110 is used to isolate the ground potentials of transceiver 110

from transmission line 114 and from the ground potentials of the transceivers of other physical modules of the local area network of which cable 114 and transceiver 110 or elements as described in the above identified concurrently filed application.

5 The transceiver of each of the physical modules of such a network is substantially a duplicate of the transceiver 110 illustrated. The presence of 5.0 M Hz data signals and 12.5 k Hz timing signals on cable 114 produce or induce voltages across transceiver winding 18 of coupling transformer 112 in which both
10 the data and timing signals present on cable 14 exist. The signals across transceiver winding 118 are applied to the inverting and noninverting terminals of operational amplifier 20 through low pass R.C. filter 22 which filters out the 5.0 M Hz data signal from the composite signal applied to operational
15 amplifier 20. Capacitor 24 is connected between pins 1 and 8 of operational amplifier 20, to control its frequency response. Operational amplifier 20 removes from its output at pin 6 any of the 5.0 M Hz not removed by filter 22. The output of operational amplifier 20 thus, is the 12.5 k Hz timing signal.
20 The output of operational amplifier 20 is coupled through capacitors 25 and 26 which are connected in parallel to the inverting input of voltage comparator 28. The voltage on the noninverting input of comparator 28 is a reference voltage of approximately 10 to 20 millivolts (mV.) so that the output of
25 comparator 28 at its terminal 12 is normally high when no signal is being applied to its inverting input. Comparator 28 squares

the quasi-sinusoidal timing signals applied to it so that the output of comparator 28 is a square wave signal, the pulse width of which is representative of the pseudo-sinusoidal wave applied to its noninverting input.

5 The amplitude of the 12.5 k timing signals applied to a given transceivers timing signal receiver from the coaxial cable 114 varies as a function of the number of physical modules connected to cable 114 and the length of cable 114. As a result, the width of the square waves produced by comparator 28 may vary from 30-60
10 microseconds (μ sec.). Such undesirable variations in pulse width are eliminated by connecting the output of comparator 28 to one shot multivibrator 30, the time constant of which is set to 40 μ sec. Multivibrator 30 senses the leading edges of the output signals of comparator 28 so that each pulse produced by
15 comparator 28 is converted into a square wave signal having the period corresponding to a frequency of 12.5 k Hz at the output of one shot multivibrator 30. The output of multivibrator 30 is applied to differential driver 32 which produces amplified and reshaped 12.5 k Hz timing signals at TTL levels. The timing
20 signals produced by driver 32 are one of the two outputs of transceiver 110. The timing signal output of transceiver 110 is produced by its timing signal receiver which includes filter 22, operational amplifier 20, comparator 28, one shot 30 and driver 32.

25 The signals across transceiver winding 18 are also applied to very high frequency differential amplifier 34 of the data signal

amplifier of transceiver 10 which amplifies the composite signal, both the 5.0 M Hz data signals and the 12.5 k Hz timing signals present. The output of differential amplifier 34 passes through high pass R.C. filter 36 which removes any 12.5 k Hz timing signals. The 5.0 M Hz data signals out of filter 36 are applied to a very high speed voltage comparator 38. The 5.0 M Hz signals applied to comparator 38 may not be a square wave as applied to cable 114 because the higher frequency components of the square wave signals are attenuated to some degree by the characteristics of cable 114. Comparator 38 squares the 5.0 M Hz data signals so that their wave form is once again that of a square wave with pulse widths of 100 ± 10 nanoseconds. The squared up and amplified data signals, the output of comparator 38 passes through Nor-Gates 39 and 40 when both are enabled. Another function of gates of 39 and 40 will be explained below. From gate 40 the output signals of comparator 38 are applied to differential driver 42 which produces amplified and reshaped, or squared, 5.0 M Hz data signals at TTL levels, the second output of transceiver 110. The data signal output of transceiver 110 is produced by its data signal receiver which includes differential amplifier 34, filter 36, voltage comparator 38 and driver 42.

Data signals, Manchester encoded binary square wave signals at a frequency of 5.0 M Hz, that are to be driven onto, or applied to, cable 114 are designated as TXDATA+ on the drawing and are the data input signal to the amplifier circuit of transceiver 110. The data signals TXDATA+ are input to differential driver

44, the two outputs of which are applied respectively to the bases of npn transistors 46, 47. The center tap of transceiver winding 118 of coupling transformer 112 is connected to a + 5 VDC power supply with the collector of transistor 46 connected to one end of winding 118 and collector of transistor 47 connected to the other. When transistor 46 is turned on, a voltage will be developed across transmission line winding 116 of one polarity and a voltage of the opposite polarity is developed across winding 116 when transistor 47 is turned on to drive or apply a bipolar square wave data signal having a peak to peak voltage of 800 m v and corresponding to TXDATA+ onto coaxial cable 114.

Differential driver 44 is a tri-state device, the outputs of which are forced to their high impedance state when the voltage on pin 4 of driver 44 is a logic zero. Driver 44 is enabled when the output of negative And gate 48, the signal XMITEN+ is positive. The inputs to gate 48 are the transmit enable control signal TXMITA- and the loop back control signal TXLOPA+. To enable driver 44, the control enable signal, TXMITA- must be negative and the loop back control signal TXLOPA+ must also be negative. When TXLOPA- is positive Nor gate 50 is disabled which prevents data signal TXDATA+ from being applied as one input to Nor Gate 40.

At such time as control signals TXLOPA+ is positive and TXLOPA- is negative, the signal XMITEN+ is also negative or a logic zero which disables driver 44 and puts it in its disabled high impedance state. Transistors 46 and 47 will both be off

when driver 44 is high in its impedance state. Nor gate 50 is enabled by the signal TXLOPA- and Nor gate 39 is disabled by the signal TXLOPA+. Data signals TXDATA+, will under these circumstances, be applied through Nor gates 50 and 40 to differential driver 42. Thus, when TXLOPA+ is positive, data signals TXDATA+ are applied to the differential driver 42 which causes the signals TXDATA+ as applied to transceiver 10 to be the data signal output of transceiver 110.

The reason for looping back data signal TXDATA+ is to permit TXDATA+ to be tested without driving it onto cable 114, since if TXDATA+ were not within specifications, it could interfere with the proper operations of other physical modules of the network receiving signals from cable 114.

Differential driver 44 includes a circuit which causes the outputs of driver 44 to be in their high impedance output state during the transitions associated with powering up and powering down. This high impedance state assures that transistors 46 and 47 will be off during such transitions. As a result the data transmitter circuit of transceiver 110 which includes differential driver 44, and transistors 46 and 47 will present a high impedance to transmission line 114 during such transitions as well as when no data signals are being driven onto the coaxial cable 114 by transceiver 110's transmitter circuit.

The data receiver of transceiver 110 produces data signals as its output whether the source of the data signals is from transmission line 114, the data signal transmitter circuit of

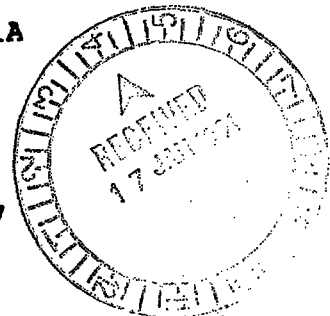
transceiver 110, or when transceiver 110 is in its loop back mode, the data signals TXDATA+ are applied directly to the line driver of its data receiver circuit. The timing signal circuit of transceiver 110 produces only timing signals as its output, which
 5 timing signals are received only from cable 114 of the local area network. It should be noted that in the preferred embodiment each physical module of the local control network is provided with two transceivers as illustrated. One for each of the coaxial cable of the transmission medium. As a result, each
 10 physical module can transmit and receive data signals over both cables and can also receive timing signals from both cables of the transmission medium.

From the foregoing it is believe obvious that the transceiver of this invention can transmit to or receive data signals from a
 15 coaxial cable and simultaneously receive a lower frequency timing signal.

In the preferred embodiment the significant components of transceiver 10 are:

op amp 20 --	LM 301A
20 Voltage comparator 28	LM319
Multivibrator 30	96L02
Differential Driver 32, 42 & 44	DS3487
Diffential Amplifier 34	LM733
Voltage Comparator 38	LM361

25 which are commercially available products of the National Semiconductor Corp.



Claims:

1. A transceiver for applying signals to and receiving signals from a transmission line (114), characterized by:
 - a) transmitter means (44, 46, 47, 112) for applying a first type of signal having a first frequency to the transmission line (114);
 - b) first receiver means (34, 36, 38, 39, 40, 42) for receiving said first type of signal having said first frequency from the transmission line (114); and
 - c) second receiver means (22, 20, 28, 30, 32) for receiving a second type of signal having a second frequency from the transmission line (114), said second type of signal being present on the transmission line concurrently with the first type of signal.
2. The transceiver of claim 1, characterized in that
 - d) a coupling transformer (112) has a transmission line winding (116) and a transceiver winding (118),
 - e) the driver transmitter means (44, 46, 47) are coupled to the transceiver winding (118) for driving the first type signal having a first frequency onto the transmission line (114);
 - f) the first receiver means (34, 36, 38, 39, 40, 42) are coupled to the transceiver winding (118) for receiving first type signals having said first frequency; and
 - g) the second receiver means (22, 20, 28, 30, 32) are coupled to the transceiver winding (118) for receiving the second type signal having a second frequency which is carried by the transmission line concurrently with the first type signal.
3. The transceiver of claim 2 for applying substantially square wave data signals having a first frequency to a coaxial cable (114) and for receiving data signals carried by the coaxial cable and quasi-sinusoidal timing signals having a second frequency, which data and timing signals



are carried concurrently by the coaxial cable,
c h a r a c t e r i z e d i n t h a t:

- h) the transmitter means (44, 46, 47) induce a voltage across the transmission line winding (118) to apply data signals to the coaxial cable (114);
- i) the first receiver means (34, 36, 38, 39, 40, 42) receives and reshapes data signals present in voltages across the transceiver winding (118); and
- j) the second receiver means (22, 20, 28, 30, 32) receives, amplifies and reshapes the quasi-sinusoidal timing signals into substantially square wave signals of the second frequency.

4. The transceiver of one of the preceding claims, c h a r a c t e r i z e d i n t h a t

- k) the first type signals are substantially square wave Manchester encoded data signals having a first frequency and a predetermined maximum peak to peak voltage and the second type signals are quasi-sinusoidal timing signals, said timing signals having a second frequency, which data and timing signals are present at the same time on the coaxial cable;
- l) the first data signals receiver (34, 36, 38, 39, 40, 42) includes a driver circuit (42) for making the data signals available as an output of the transceiver; and
- m) the timing signal receiver (22, 20, 28, 30, 32) includes a driver circuit (32) for making the timing signals available as an output of the transceiver.

5. The transceiver of Claim 3 or 4, c h a r a c t e r i z e d i n t h a t the frequency of the data signals is 5.0 MHz.

6. The transceiver of Claim 5, c h a r a c t e r i z e d i n t h a t the frequency of the timing signals is 12.5 KHz.



7. The transceiver of one of the claims 4 to 6, c h a r -
a c t e r i z e d b y a loop back circuit (50, 40)
which when enabled in response to a loop back enabling
signal applies data signals applied to the transceiver
5 directly to the driver circuit (42) of the data signal
receiver (34, 36, 38, 39, 40, 42), said loop back enabling
signal when present disabling the transmitter circuit
(44, 46, 47).
- 10 8. The transceiver of one of the preceding claims, c h a r -
a c t e r i z e d i n t h a t a low pass filter (22)
is connected between the transceiver winding (118) and an
operational amplifier (20) of the second receiver means
(22, 20, 30, 32).
- 15 9. The transceiver of claim 8, c h a r a c t e r i z e d
i n t h a t a comparator (28) with its one input (5)
is connected to the output of the operational amplifier
(20), at its other input (4) receives a reference voltage
20 and at its output delivers a square shaped version of the
timing signal.
10. The transceiver of claim 9, c h a r a c t e r i z e d
i n t h a t one shot multivibrator (30) is connected
25 to the output of the comparator (28) for delivering a re-
shaped timing signal.

