

Description

BACKGROUND OF THE INVENTION

Field of the Invention

This invention relates generally to transducers which may be fabricated from semi-conductor material such as silicon by batch photolithographic techniques from a monocrystalline wafer. More particularly, the invention relates to a spring-mass-support structure for detecting a signal proportional to acceleration applied to the support structure. The invention also relates to electronic circuitry in combination with the structure for detecting displacement of the spring supported mass to produce analog or binary representations of acceleration applied to the support structure.

Description of the Prior Art

A publication by Kurt E. Petersen entitled "Silicon as a Mechanical Material" published in Vol. 70, No. 5 (May, 1982) of the Proceedings of the IEEE, surveys the art of micro-machining of silicon to fabricate a mechanical device. The Petersen publication describes etching techniques commonly used in fabricating micromechanical devices and describes several devices which have been made using such techniques. The publication also describes a cantilever beam light modulator array the beams of which are fabricated of silicon dioxide. Another survey article appeared in the April, 1983 publication of Scientific American entitled, Silicon Micromechanical Devices by Angell, et al.

An accelerometer having a micromachined cantilever beam intended for in vivo biomedical studies and which includes a piezoresistive sensing element is described in a publication entitled, "A Batch-Fabricated Silicon Accelerometer" by Roylance and Angell in IEEE Transactions on Electron Devices, Vol. ED-26, No. 12, December, 1979.

U.S. patent Re.31,459 (reissue of U.S. patent No. 4,071,838) reissued December 6, 1983 in the name of Barry Block discloses a silicon folded leaf spring force transducer fabricated by batch photolithographic and etching techniques. One embodiment of such transducer consists of four E-springs located at 90° angles about the central axis of sensitivity. Piezoresistors disposed on legs of the E-springs provide a measure of the displacement of the springs in response to a force applied to the transducer.

U.S. patent 4,155,516 issued March 13, 1979, in the name of Harry Aine discloses first and second leaf spring structures, as supported from support structures, which are coupled together in axially spaced mutually opposed relation so that the

first and second leaf spring structures are coupled together for deflection in unison relative to the surrounding support structures in response to deflection of either of the first and second spring structures. A capacitive detector structure is provided for detecting displacement of the leaf spring.

U.S. patent 4,155,516 discloses that a transducer with a movable mass suspended from a support structure and having plates fixed to the structure on each side of the mass may be micro-machined from a mono-crystalline wafer. Figures 13 and 14 of the patent disclose plates on each side of the mass which face conducting areas of opposing support structures. A displacement signal is produced as a function of capacitance difference in order to determine the distance that the mass has moved between opposing plates.

U.S. patent 4,597,003 issued June 24, 1986, in the names of Aine and Block discloses a method of chemical etching a silicon wafer by undercutting an etch stopped layer to fabricate a cantilever beam supported at one end by a frame structure with an integral mass formed by the etch. Etching from opposite etch stopped layers of a silicon wafer is described.

While the art described above has advanced the art of micro-machining of silicon wafers to fabricate sensors in general and accelerometers in particular, there remains a need for a micro-machined structure of silicon and other materials and associated electronics meeting extremely precise specifications. For example, extremely low distortion for an accelerometer is required for certain applications. Prior art spring configurations and materials are believed to produce spring constants which change unacceptably over a deflection range of interest. Very low sensitivity to accelerations orthogonal to the sensitivity axis is desired. Prior art spring mass configurations do not provide sufficient insensitivity to orthogonal acceleration and angular motion applied to such configurations. Very high dynamic range (requiring extremely low noise) is desired in an accelerometer which has not been provided by prior art accelerometers and associated detecting electronics systems. Shock impact insensitivity is required in an accelerometer intended for use in certain applications such as in the seismic, automotive, or aeronautics fields.

Identification of Objects of the Invention

Accordingly, it is a primary object of the invention to provide a micro-machined spring-mass sensor which has springs having spring constants which can be precisely controlled during the fabrication process.

It is another object of the invention to provide a geometry of a spring-mass-support transducing

element which results in an accelerometer which has very low distortion and has a very low sensitivity to motion orthogonal to its sensitivity axis and to angular motion.

It is still another object of the invention to provide an accelerometer sensor which when subjected to extremely high shock impacts will survive undamaged.

Another object of the invention is to provide a micro-machined sensing device and associated electronics resulting in an accelerometer with an extremely broad dynamic range.

Another object of the invention is to provide a batch fabricated micro-machined sensing element with associated electronics for producing a signal representative of acceleration applied along the sensitivity axis of the element.

Another object of the invention is to provide a batch fabricated micro-machined sensing element with associated electronics to produce a binary bit stream representation of acceleration applied along the sensitivity axis of the element.

SUMMARY OF THE INVENTION

The invention is as described in the accompanying claims. In particular Claim 1 has been divided into a two part form based on the assumption that US-A- 4 144 516 is the nearest state of the art. Thus the features that are recited in the precharacterizing part of the claim are already known from this prior art.

The objects identified above as well as other features and advantages of the invention are incorporated in a device including a mass-spring-support sensing structure and associated electronics. The sensing structure includes a micro-machined transducer having a support frame and a sensing mass supported from the frame by at least one, preferably eight, E-shaped leaf springs. Each of the eight E-springs have a base with two outer legs and an inner leg. Each leg is connected to the base at one of its ends. The two outer legs connect to the frame at their other ends. The inner leg is connected to the mass at its other end. The base is characterized by a coefficient of stiffness substantially greater than that of the legs. Each of the legs are preferably the same length. In order to precisely define the spring constant of the springs, the legs are fabricated of silicon dioxide, because the thickness of the silicon dioxide springs may be precisely controlled during the fabrication process.

The mass-spring-support sensing structure is fabricated by orienting four E-shaped leaf spring structure patterns on top and bottom (100) faces of a silicon wafer such that they are oriented at forty-five degree angles with respect to intersection lines of (111) planes of the wafer and the opposite faces

of the wafer. The opposite faces are masked and silicon dioxide grown on surfaces which will ultimately be etched away by an etching chemical. A thin layer which is a stop layer is formed. This layer can be formed by diffusing into this layer an impurity, such as arsenic, that will change the semiconductor type from P to N in this layer for an impurity such as arsenic which will form a P/N junction between the layer and the substrate. A voltage bias can then be applied to this layer to form an electromechanical etch stop layer. The arsenic is diffused into opposite surfaces of the wafer everywhere except the "opening" surfaces and the surfaces which will ultimately serve as the legs of the spring.

A thick oxide layer is then grown by diffusing a precise amount of oxygen into the silicon to form a precise thickness of silicon dioxide on the silicon wafer. Next, the opposite surfaces of the wafer are masked exposing only the regions which will not be the springs of the structure as well as some other selected areas. These exposed areas of silicon dioxide are then etched down to the silicon surfaces. A thin silicon dioxide layer is then grown, masked, and the exposed silicon dioxide etched down to the silicon surface in the regions that will be exposed to the silicon etchant. Chemical etch of the silicon material then is applied which acts through the openings of the silicon dioxide to etch silicon material away while undercutting the springs, leaving a central mass supported by silicon dioxide spring legs. Stiff etched-stop layers remain for the bases of the springs and the opposite faces of the suspended mass and support structures.

Force and sense conducting areas are then deposited on opposite faces of the suspended mass. Conductors are deposited which run from the conducting areas via spring legs to the support structure. First and second plates having corresponding force and sense plates with conductors running to external connecting pads are fixed to opposite support surfaces of the mass-spring-support structure such that each of the conducting surfaces of the mass precisely face a corresponding conducting surface of the first and second plates.

An electrostatic electronic feedback system is provided to detect displacement of the mass of the mass-spring-support structure in response to acceleration or force applied to the support. For frequencies of the applied acceleration below the resonant frequency of the mass-spring-support structure, displacement of the mass is proportional to acceleration. Displacement of the mass (and consequently the acceleration) applied to the support structure is measured by applying a first voltage between the sense conducting areas of the op-

posite plates and applying a second voltage between the force conducting areas of the opposite plates. The first voltage creates a sense electric field between the sense conducting areas of the suspended mass and induces a sense voltage on it proportional to the displacement of the mass between the opposite plates. The second voltage creates a force electric field between the force conducting areas of the mass.

A forward circuit responsive to the sense voltage generates a displacement signal which is representative of the displacement that the mass has moved from a reference position between the plates. A feedback circuit responsive to the displacement signal applies an amount of electric charge on the force conducting area of the mass surface such that a force is created on the mass proportional to the product of the amount of charge and the force electric field so as to substantially restore the mass to the reference position. The displacement signal is an analog representation of acceleration. The first and second voltages may be d.c. voltages, or may be square waves operating at a modulation frequency higher than expected frequencies of the acceleration to be measured. A demodulator is required to produce an unmodulated analog output signal representative of acceleration.

Another embodiment of the invention, like the ones described above, includes a sensor for generating a displacement signal representative of displacement of a mass from a reference position with respect to a support structure. A forward circuit is provided which, in response to the displacement signal, converts the displacement signal to an output binary stream. A feedback circuit, in response to such output binary stream, generates a binary force on the mass in a direction tending to restore the mass to the reference position. Consequently, the output binary stream is representative of a characteristic of motion, preferably acceleration, of the support structure.

BRIEF DESCRIPTION OF THE DRAWINGS

The objects, advantages and features of the invention will become more apparent by reference to the drawings which are appended hereto and wherein like numerals indicate like parts and wherein an illustrative embodiment of the invention is shown, of which:

Figure 1 is a transverse sectional view of a wafer of monocrystalline material from which a spring mass system is to be fabricated according to micro-machining techniques;

Figure 2 is a view similar to that of Figure 1 showing an oxide layer formed on opposite sides of the wafer;

Figure 3 is a plan view of the wafer of Figure 2 showing areas for which spring-mass-structures will be fabricated;

Figure 4A is an enlarged view of an area of Figure 3 as indicated by line 4A of Figure 3 which indicates the orientation of the crystal top and bottom surfaces and its relation to (111) planes of the crystal;

Figure 4B is a sectional view of the wafer showing a part of the layout taken along the lines 4B-4B of Figure 4A with photo-resist material having been applied to the opposite sides of the wafer;

Figures 5A and 5B are similar to Figures 4A and 4B but show a wafer portion after it has been masked with a pattern of a spring-mass-support structure, and after portions of the oxide layers have been removed, leaving a pattern of oxide partially defining ultimate etching openings on the opposite sides of the wafer;

Figures 6A and 6B are similar to Figures 5A and 5B but reflect further processing after etch stop material has been diffused into the crystal of the wafer on opposite surfaces in areas other than the etching openings and spring legs;

Figures 7A and 7B are similar to Figures 6A and 6B but reflect further processing after opposite wafer surfaces have been masked to allow further etch stop layer material onto wafer surfaces where surrounding support structure, central mass and bases of E-springs are patterned;

Figures 8A and 8B are similar to Figures 7A and 7B but reflect further processing after the etching openings on both sides of the wafer have had oxide previously covering them removed and show a relatively thick layer of silicon dioxide material having been grown over areas of the sides of the wafer defining E-spring legs;

Figure 9A shows a plan view of the one-half of the wafer after chemical etching of the wafer through etch openings has etched to (111) planes of the crystal of the wafer thereby undercutting the E-spring structures and severing the mass portion of the structure from the support portion while suspending the mass from the support and after further processing where the thin layer of etch stop material beneath oxide sides of the springs have been removed;

Figure 9B is a sectional view of the wafer illustrated in Figure 9A taken along section lines 9B-9B;

Figure 10 is a perspective view of the sensing element structure including support, E-springs and sensing mass, the view being partially cut-away to illustrate the central mass severed from the support structure to show the oxide springs and undercut etch stop layers of the mass, springs and support structure;

Figure 11 illustrates a cross-section through the sensing element showing the sensing mass supported by springs to undercut portions of the support member where gravity forces the mass to be disposed beneath the top plane of the support member, the view further showing a composite structure including top and bottom plates having force and sense conducting areas deposited on their faces corresponding to similar force and sense conducting areas on opposite surfaces of the sensing mass;

Figure 12 is a plan view of the sensing structure illustrating deposited conducting leads to the force and sense conducting areas of the mass surface from the support structure via the E-spring leads;

Figure 13 is a schematic illustration of an electrostatic feedback control system for detecting acceleration applied to the support system by substantially maintaining the sensing mass at a predetermined position between the opposite plates of the composite structure, the output of the system being an analog signal proportional to applied acceleration;

Figure 14 shows a schematic illustration of a system similar to that of Figure 13 but having an alternating voltage applied to the covering plate force and sense conducting areas, with a demodulator provided to develop an analog signal proportional to applied acceleration;

Figure 15 is a schematic illustration of a charge generator circuit used directly or with modification in the transducer systems of Figures 13, 14 and 16, 17, 18 and 19;

Figure 16 is a schematic illustration of a transducer having a binary bit stream output representative of the displacement or acceleration of a mass in a mass-spring-support structure with binary bit stream voltages applied to top and bottom force plates and with a constant charge applied to force plates of the mass with the result that a binary force is applied to the mass tending to restore it to its reference position;

Figure 16A is an illustration of a binary sampler used in the embodiments of Figures 16, 17, 18 and 19;

Figure 17 is similar to the transducer of Figure 16 except that a binary bit stream of charge is applied to the force plates of the mass and a constant electric field is applied across the force plates of the top and bottom plates;

Figure 18 is similar to the transducer of Figure 17 except that an electric field applied to the force plates of the top and bottom plates is modulated with an a.c. signal, and consequently the binary bit stream of charge applied to the force plates of the mass is also modulated; and

Figure 19 is similar to the transducer of Figure 18 except that the charge applied to the force plates of the mass is modulated and modulated bit streams of the output signal are fed back to the force plates of the top and bottom plates.

DESCRIPTION OF THE INVENTION

Two important motivating factors, among others, led to the accelerometer of this invention. The first is low manufacturing cost so that the resulting accelerometer may find wide application where many force sensing transducers are required. The second is an accelerometer having extremely high performance characteristics: wide dynamic range; low distortion; low sensitivity to motion orthogonal its sensitivity axis; and resistance to high shock impact. These requirements lead to the fabrication of the accelerometer by micro-machining techniques from a monocrystalline material such as silicon.

Referring now to Figure 1, a typical wafer 10 is shown from which a batch of accelerometers are to be fabricated. In a typical example, the wafer 10 is made of a nonmetallic monocrystalline material such as silicon, germanium, quartz, gallium arsenide, gallium phosphide, etc. In a preferred embodiment, the wafer 10 is made of a diamond cubic material, such as silicon. The wafer 10 has a thickness of about 500 microns, has polished top and bottom surfaces 12, 14 and has a convenient diameter, such as 3 to 8 inches. In the case of diamond cubic material, the crystallographic plane (100) is preferably formed at the upper 12 and lower 14 major surfaces of the wafer 10. Furthermore, the wafer 10, in the case of silicon, is preferably doped with a P type dopant to a conductivity of between 0.1 - 10.0 ohm-cm.

In the next step of the process, the wafer 10 is oxidized on opposite sides to form oxidized layers 16, 18 of 2500 angstroms in thickness as illustrated in Figure 2. This is conveniently achieved by putting the wafers in a furnace at 1100° C, in the presence of oxygen. Next, the oxide layers 16, 18 are each coated with a photoresist material such as KMER.

Figure 3 is a plan view of the wafer with multiple areas each of which are to be processed simultaneously to batch produce the mass-spring-support structures of the invention. One particular area is illustrated in Figure 3 with line 4A-4A, shown much enlarged in Figure 4A. The boundary edges 24 of the structure to be fabricated are oriented parallel to the (111) planes of the crystal as illustrated in Figure 4A. Each structure of the array of structures of Figure 3 will ultimately be cut along the boundary edges 24 parallel to the (111) planes of the crystal after batch processing. Figure

4B shows a cross-section through a portion of the wafer showing the KMER photoresist layers 20, 22 deposited over silicon dioxide layers 16, 18.

Figure 5A shows one side of the wafer (for convenience called here the top side) after masking and exposure of the pattern as depicted thereon. The unexposed photoresist portion covers the etch opening patterns 30' and the spring legs 30", but the remainder of the photoresist portions are removed by exposure and development of the photoresist. The wafer is then placed in a silicon oxide etch solution such as buffered hydrofluoric acid, which attacks the oxide but not the photoresist or the underlying silicon. The photoresist protects the pattern 30 as illustrated in Figure 5A. The pattern 30 includes two parts: the regions indicated by 30' will ultimately be etch openings; the regions marked 30" will ultimately be spring legs. Figure 5B shows a cross-section through lines 5B-5B of Figure 5A and illustrates the oxide portions 30' and 30" on top and bottom surfaces 12, 14 which respectively ultimately form a slot opening behind the base of the F-spring and the middle leg of the E-spring. Oxide portions 30' and 30" are greatly enlarged in Figure 5B for ease of illustration.

The oxide pattern 30' of Figures 5A and 5B will ultimately serve as the etch openings to fabricate the three dimensional mass-spring-support structure of the accelerometer. The oxide patterns of 30" will ultimately be the location of the spring legs of the E-springs. A perspective view of the structure may be seen in cut-away sectional view of Figure 10 showing the structure at a certain stage in the processing.

Figure 5A illustrates particular features of the grown oxide patterns 30' and 30" which ultimately will result in the structure of Figure 10. The lines 32-32 of Figure 5B show a region in which an "E"-shaped spring 40 will be fabricated to partially support a mass 36 from a surrounding support structure 38. The E-spring 40 will include a base 42 and three legs 44, 46, 48. Slots or openings 43, 45, 47 will ultimately outline the shape of the E-spring. The inner leg 46 is twice the width of outer legs 44, 48. The lengths of the legs 44, 46, 48 are precisely the same.

Each E-shaped spring 40 is placed at the corner of intersecting (111) planes, that is, at 45° angles to intersecting lines of (111) planes which intersect the surfaces 12, 14 of the mono-crystalline wafer 10. The base 42 of each spring will ultimately be fabricated of stiff etch stop material while each leg 44, 46, 48 will ultimately be fabricated of silicon dioxide, as described below in more detail.

Two characteristics of the shape of the pattern of lines 32-32 are important. The first relates to the area identified by arrow 50, which connects the

inner leg 46 to the sensing mass 36. Solely for purposes of illustration and description, this region is called the "cat head with ears" region, and is pointed to by reference arrow 50. The top of the cat head is attached to inner leg 46 with its "neck" or "body" attached to mass region 36. The "ears" of the cat head and ears region extend into oxide areas which ultimately will be openings between sensing mass region 36 and support region 38. The cat head and ears geometry provides a stiff region of mass 36, ultimately to be fabricated of undercut etch-stop material, to which inner leg 46 is connected.

The second shape relates to an area pointed to by reference arrow 52 of Figure 5A. Solely for purposes of illustration and designation, this region is called the "hockey stick" region because of its general resemblance to the shape of a hockey stick. This hockey stick region, ultimately to be undercut etch stopped material between regions 43 and 56 (which ultimately will be openings formed by means of silicon etching), provides a relatively wide region at the top of the foot of the hockey stick by which outer leg 44 (or 48) attaches to etch stopped material. Such wide region provides a strong connection between the ends of the outer legs of the E-springs and the etch stopped material of the surrounding support structure.

The oxide areas 56 (which ultimately will be openings) have edges 58, 60 which are coexistent with intersecting lines of (111) planes of the crystal with the (100) surface 12 of the wafer. These areas 56 serve as "etch-time and undercut control" areas and are given the short-hand notation, "ETUC" areas. They ultimately serve as large open areas by which etchant chemical will undercut E-springs 40. Their area must be relatively large to allow etchant to etch to intersect (111) planes having intersecting lines 62, 64 with the surface 12. Areas 56 must be large enough to speed the etching process to completion to the intersection of the planes of the intersecting lines 62, 64, but must be shaped as shown to provide the "hockey stick" etch stopped areas 52 on its sides as discussed above. The inner edge 60 defines the intersecting line of a (111) plane of the crystal which will define the outer surfaces of the mass region 36.

Another factor dictating the shape of ETUC areas 56 is the requirement that during chemical etching, the mass region 36 must be severed from the surrounding support frame 38. Chemical etching of the silicon wafer occurs in one direction along (111) planes which intersect with lines such as 64 and 62 of the top and bottom surfaces which are extensions of edges 58 of ETUC areas. Chemical etching of the silicon wafer occurs in the opposite direction along (111) planes which intersect lines which extend from inner edges 60 of the

ETUC areas of the top and bottom surfaces. When the etching from the opposite directions along (111) planes meet, the mass 36 is separated from support frame 38 as can be seen in Figure 9B. Consequently, the distance between edges 58 and 60 must be great enough to allow the mass region 36 to be separated from support frame 36 during chemical etching.

Turning now to Figure 6A, the silicon dioxide regions 30' defining openings of Figure 5A and the spring leg regions 30'' are masked, and the uncovered silicon portions are implanted with a thin layer of arsenic. The arsenic is diffused a short distance, (6 to 7 microns) in the silicon to form N⁻ layers on top and bottom surfaces that are used as electrochemical etch stop layers in the silicon crystal. This stage of the processing is illustrated in the cross-section of Figure 6B. Next, the oxide of spring leg region 30'' (of legs 44, 46, 48) is removed and an additional two micron thickness of arsenic is diffused into the silicon. Figure 7B shows this stage of the processing. The thickness of the etch stop layer 70' is now about 8-9 microns thick.

In preferred processing of the mass-spring-support structure of this invention, the thick layer of arsenic is especially important to impart stiffness to certain regions of the top and bottom etch stop layers. The base of the springs 42, cat head region 50 and the hockey stick region 52 require stiffness and strength to connect with the spring legs 44, 46, 48 so that the mass region 36 may be adequately supported from support region 38.

Figures 8A and 8B show the condition of the wafer after the silicon dioxide regions 30' have been stripped from the wafer 10, a thick layer (1.2 microns) of silicon dioxide has been regrown everywhere over top and bottom surfaces, and finally such silicon dioxide has been stripped away everywhere except over spring leg portions 44, 46 and 48. Silicon dioxide layer 30''' remains for such legs as is shown in Figure 8B. At this point in the processing, conductive areas are provided on the opposite faces of the mass region 36 and conductive leads running from the support region via the legs 48 or 44, 46 and base 42 are deposited. For simplification of illustration, the results of providing such conductive region and leads are illustrated in Figure 12 and described below.

Next, the wafer is subjected to electro-chemical etching such that the wafer is etched away from the ETUC areas 56 and other open areas, such as slots 43. The etch chemical etches beneath etch stop layer 70' preferentially along (111) planes of the crystal of the wafer. The preferred etching process is similar to that described in U.S. patent 4,597,003, described above and incorporated herein by reference for its electro-chemical etching methods. Next, the thin layers 70''' (Figures 7B,

8B) of etch stop material are removed by subjecting the wafer to a silicon etch (either dip or plasma) everywhere. Such etching causes etch stop layer 70' to be reduced in thickness by about 2 microns, that is, to an ultimate thickness of about 6-7 microns.

The result of the processing described above is a three dimensional support-spring-mass structure illustrated in plan and cross-section views of Figures 9A and 9B and the perspective view partially broken away of Figure 10. The etching process has etched the silicon material away, along the (111) planes defined by the intersection of such planes with lines parallel to edges 58, 60 of each of the ETUC areas 56. Undercut etch stop layers 70' are visible in Figures 9B, and 10 as well as the suspended mass 36, surrounding support structure 38 and silicon dioxide spring layers 30'''. The bases 42 of the E-springs 40 are formed of undercut etch stop material 70' as seen in Figure 9B and 10.

An important feature of the invention is apparent from Figures 9A and 9B. That feature is the fact that bases 42 of the E-springs are fabricated of etch stop material 70' while the legs 44, 46, 48 of the E-springs are substantially entirely of silicon dioxide. This difference in materials allows the stiffness of the base of each E-spring to be substantially greater than that of the legs.

The coefficient of stiffness (Young's modulus) is approximately three times greater for silicon than it is for silicon dioxide. The thickness of the base 42 at each E-spring is controlled during processing to be greater than seven times as thick as that of the silicon dioxide legs. The resulting stiffness of the base, consequently, is seven cubed times three, or more than 1000 times stiffer than the legs for similar lengths and widths of material.

Each spring leg 42, 46, 48 of each E-spring 40 is precisely the same length from its connection at one end of its respective base 42 to its respective connection to either the top of the foot of the hockey stick area or to the head of the "cat head with ears" area. It is fabricated of silicon dioxide, as described above, because the thickness of the silicon dioxide legs may be very accurately controlled. The length of each leg and its thickness must be accurately controlled so as to control the spring constant of the spring member. It can be shown that the spring constant of the E-spring is proportional to the cube of the thickness of its legs. Another reason for fabricating the legs of each E-spring of silicon dioxide is that silicon dioxide is a well-behaved material in that it has no measurable hysteresis and it does not fatigue over many cycles of flexing.

Legs 44 and 48 are precisely the same width. Leg 46 is made up of one or more side by side

legs whose composite width is precisely twice the width of legs 44 or 48. Although not illustrated, leg 46 is preferably a composite of three side by side legs, each leg of course having a smaller width than the composite width. Dividing the total width of leg 46 into three parts has been found to prevent buckling or crinkling of leg 46 after processing.

Figure 11 shows a cross-section of sensing element 80 sandwiched between opposite supporting members 82, 84. Sensing element 80 shows central mass 36 supported via spring legs 46, 48 and legs 46, 44 via bases 42 from surrounding support member 38. The mass 36 is shown displaced vertically by the force of gravity. The sense conducting areas 90 formed on opposite sides of the faces of mass 36 surround force conducting areas 92. Figure 12 shows a plan view of the support-spring-mass structure 80 with the force and sense conducting areas 92, 90 formed on the top face of the mass 36. Such areas 90, 92 are conductively isolated from each other. Although not shown in Figure 12 for simplification of illustration, guard regions are provided between areas 90, 92 to reduce the capacitive coupling between such areas. Corresponding force and sense conducting areas 92', 90' are provided on plate surfaces formed on opposite supporting members 82, 84. Such plate surfaces and force and sense conducting areas 92', 90' correspond in size and pattern to the opposite surfaces of mass 36 and its force and sense conducting areas 92, 90. Opposite supporting members are bonded to the support member 38 by conventional means.

Figure 12 shows conductors running from external connection pads to the force and sense plates. Such conductors are illustrated for one surface of the support-spring-mass structure 80 as an illustration of such conductors provided to the sense and force conducting areas of the top and bottom plates 82, 84 and the opposite surfaces of mass 36. Conductors 94 and 96, preferably gold, are deposited on opposite surfaces of the wafer and structure 80 such that they run respectively from connecting pads 98 and 100 via separate E-springs 40A and 40B. As illustrated, lead 94 connects to force conducting area 92 via legs 48 and 46 and base 42 of E-spring 40A, while lead 96 connects to sense conducting area 90 via legs 44 and 46 and base 42 of E-spring 40B.

The leads 94 and 96 respectively pass through tunnels 102, 104 provided for a seal (not shown) between sandwiched surfaces of support member 38 and opposite supporting members 82, 84. Conductor 94' is deposited on opposite leg 44 and a portion of base 42 of spring 40A to balance the weight and stiffness of conductor 94 on leg 48 and base 42. Likewise, conductor 96' is deposited on opposite leg 48 and a portion of base 42 of spring

40B to balance the weight and stiffness of conductor 96 or leg 44 and base 42. Conductor 94' is not electrically connected to conductor 94. Likewise, conductor 96' is not electrically connected to conductor 96.

Conductors 106 and 108 are deposited on E-springs 40C and 40D, but have no external electrical connection. Such conductors 106 and 108 provide weight and stiffness to springs 40D and 40C to balance the weight and stiffness of conductors deposited on springs 40A and 40B. Each of the four surfaces, two opposite faces of mass 36 and opposing surfaces of opposite support members 82 and 84, have similar conducting leads attached to their respective force and sense conducting areas. Such leads are connected to appropriate pads illustrated adjacent pads 98, 100 of Figure 12 which communicate with electronic circuitry described below.

Turning now to the electronic circuitry associated with the spring mass transducer described above, three alternative embodiments of transducer systems, or accelerometers, are illustrated respectively in Figures 13, 14 and 15. Figure 13 shows a mechanical schematic of the spring mass transducer of Figure 11.

Mass 36 is supported from a support 38 via springs 40 having substantially equal spring constants of 2ks. A d.c. voltage of magnitude V_{cc} is applied between sense conducting areas 90' on plates of opposite support members 82, 84, which of course are fixed with support 38. A d.c. voltage of magnitude V_{qq} (which may be the same as or different from magnitude V_{cc}) is applied between force conducting areas 92'. Leads 96 connected to top and bottom sense conducting areas 90 of mass 36 are connected to lead 110. Leads 94 connected to top and bottom force conducting areas 92 of mass 36 are connected to lead 112.

A differential amplifier 120 has one of its two inputs connected to lead 110 with another of its leads connected to a reference voltage which is provided with some fraction of magnitude V_{cc} , preferably, one half V_{cc} . The voltage appearing on lead 110 is proportional to the position of mass element 36 between the plates of support members 82 and 84, because a sense electric field is created between sense conducting plates 90' by voltage V_{cc} , and because sense conducting plates 90 of mass 36 are disposed in the path of that sense electric field. When the mass element 36 moves toward the plate of top support member 82, the voltage on sense plates 90 and lead 110 approach the magnitude V_{cc} ; conversely, when the mass element 36 moves toward the plate of bottom support member 84, the voltage on sense plate 90 and lead 110 approaches zero or ground magnitude.

Consequently, with the reference voltage $V_{cc}/2$ applied to differential amplifier 20, its output on lead 122 is a sense displacement signal proportional to the distance that mass element 36 has moved from a reference position, halfway between the reference plate of supports 82, 84. The displacement signal produced on lead 122 is smoothed and filtered by filter circuitry 124, resulting in an output signal on lead 126.

The voltage V_{qq} applied between force conducting areas of plates of top and bottom support members 82, 84, creates a force electric field across the force conducting areas 92 of mass 36. A negative feedback circuit, labelled as charge generator 130 (described below and illustrated in Figure 15) produces an output on its output lead 112 in response to the output displacement signal from lead 128 to apply an amount and magnitude of electric charge on force conducting areas 92 to move the mass toward its reference position, preferably halfway between support members 82, 84. The force on mass 36 is proportional to the numerical product of the amount of charge deposited on plates 92 times the force electric field, provided the mass is maintained close to or at its reference position.

In other words, force or acceleration applied to support 38 causes displacement of the mass 36 via the springs 40. Displacement of the mass from a reference position is sensed as a displacement signal which causes a charge generator 130 to generate a charge which is applied to conducting areas 92 so as to drive the mass back to its reference position. Consequently, the difference between the sense displacement signal on lead 110 and the reference potential $V_{cc}/2$ applied to lead 120 is extremely small, yet the output displacement signal on lead 126 after amplification, and filtering of the output of differential amplifier 120 is representative of the displacement of mass 36 in response to acceleration applied to support 38.

As long as the frequency of applied acceleration is below the effective resonant frequency of the spring-mass-support transducer, displacement of the mass of the transducer is directly proportional to acceleration, and the system of Figure 13 can be described as an accelerometer. Consequently, the output voltage signal of lead 126 is an analog signal proportional to acceleration for acceleration frequencies below the effective resonant frequency of the transducer.

Figure 14 is a block diagram of an accelerometer similar to that of Figure 13 but has an alternating voltage, preferably a square wave, applied between sense conducting areas 90' and force conducting areas 92' of the plates of opposite support members 82, 84. Different amplitudes of

the square waves may be applied to the force and conducting areas. A square wave train, labelled V_{o1} , is illustrated in Figure 14 representative of the alternating voltage applied to the force and sense conducting areas 92', 90' of the opposite plates. The frequency of the square wave-train is high compared to that of an illustration of the displacement of mass element 36 in response to acceleration applied to support 38. The voltage appearing on lead 110 is an amplitude modulated signal (suppressed carrier) representative of the displacement of mass 36 from a reference position.

The difference signal of differential amplifier 120 is applied to band pass filter 124 via lead 122 which results in a modulated displacement signal on lead 126. The charge generator circuit 130 produces modulated voltages to apply charge to force plates 92 of mass 36 in synchronism with the square wave voltage applied to force conducting areas 92' of opposite support members 82 and 84 to drive the mass to its reference position.

The voltage on lead 126 is applied to demodulator 134 along with V_{o1} and V_{o1} inputs to produce an analog voltage on lead 136 representative of acceleration applied to supports 38 of the transducer. The advantage of the carrier system of Figure 14, over the analog system of Figure 13, is that the mass displacement signal is shifted out of the base band of the acceleration signal to a higher carrier signal of the square wave train thereby minimizing 1/f noise and d.c. offset problems.

The charge generator 130 of Figure 13 is illustrated in detail in Figure 15. The voltage from the output of the system of Figure 13 is applied via lead 128 to charge generator 130. Two control signal voltages, ϕ_1 and ϕ_2 as illustrated in Figure 15, are used to control conduction of MOS transistor pairs Q_1 , Q_2 and Q_3 , Q_4 respectively. Control signal voltages ϕ_1 and ϕ_2 provide non-coincident turn on signals of a period which is small compared to the period T of the highest frequency acceleration input to the system.

Capacitor C_{REF} is disposed in the conduction path of transistors Q_1 and Q_2 . Consequently when Q_1 and Q_2 are turned on by control voltage ϕ_1 , a charge is placed on C_{REF} proportional to the output voltage V_{out} and lead 128. A positive charge is placed on capacitor C_{REF} if the output voltage V_{out} is positive, and vice versa. After control voltage ϕ_1 is turned off, transistor Q_3 and Q_4 are turned on by control voltage ϕ_2 thereby placing the charge stored on capacitor C_{REF} across the inputs of very high gain operational amplifier 137. In response, the output of amplifier 137 changes in a direction that causes the charge stored on capacitor C_{REF} to discharge. Lead 112 is connected between force plates 92 of the mass 36 of the sensor. Consequently, the discharge path for the charge stored

on capacitor C_{REF} is through the capacitance of the force electrodes 92 via lead 112. The charge on the reference capacitance C_{REF} is totally transferred to force electrodes 92.

As long as control voltage ϕ_1 is in the off condition, or the ϕ_1 signal is applied to transistor pair Q_7 , Q_8 , voltages V_{gg} and ground are applied respectively to corresponding force plates 92' of the top and bottom support plates 82, 84. During the time when control signal ϕ_1 is positive and ϕ_1 is not present, transistor pair Q_5 , Q_6 conduct and cause top and bottom pairs of plates 92' and 92 to be connected together, thereby equalizing charge on the respective plates.

The circuit 130 could be modified such that the voltage V_{gg} is constantly applied to Q_1 on lead 128 and the output signal from lead 126 is applied to transistor Q_7 . This causes a fixed value of charge to be applied to plates 92 while the voltage across plates 92' is equal to the output voltage on lead 126.

The circuit could also be modified such that positive and negative voltages are applied respectively to plates 92' (or 92 as indicated above) such that the electric field across plates 92' is the difference between the positive voltage and the negative voltage. Likewise, the voltage applied across capacitor C_{REF} may be a negative voltage rather than ground as illustrated. Such modification will be apparent to one of skill in the electronic design art.

The charge generator 130 of Figure 14 is similar to that illustrated in Figure 15 except that the control signal voltages ϕ_1 and ϕ_2 are in synchronism with the modulation signal V_{01} .

Figures 16 through 19 illustrate different arrangements of the embodiment of the invention where a binary bit stream representation of the displacement of the mass 36 is fed back to the mass-spring-support structure to generate a binary force on the mass in a direction tending to restore the mass to its reference position.

Figure 16 illustrates a support 38-spring-mass 36 system which is similar to the d.c. system of Figure 13 in that a displacement signal on lead 110 is applied to a differential amplifier 120 having a reference potential $V_{01/2}$ applied to it corresponding to a reference position of mass 36 between support plates 82 and 84. A difference signal on lead 112 is applied to a filter 124. The output of filter 122 on lead 126 is then applied to a binary sampler 150 which is illustrated in Figure 16A. The filtered differential displacement signal on lead 126 is applied to a threshold circuit 160 which produces a "1" ("high") or "0" ("low") signal on lead 161 depending on whether the displacement signal on lead 126 is greater than or less than a voltage threshold signal applied to lead 159. Consequently, the signal on lead 161 is either a "1" or a "0".

The binary signal on lead 161 is applied to flip flop circuit 162 to which is applied a clock signal labeled T_{CLOCK} and which is illustrated in Figure 16. The flip flop circuit produces outputs on its Q and Q leads 164, 166 which is a binary stream of pulses, as illustrated in Figure 16.

When the signal on lead 161 is a "1", the binary stream on the Q output (lead 164) is a stream of "1" pulses each of the same period as the pulses of the T_{CLOCK} signal. Simultaneously, the Q output (lead 166) is a binary stream of "0" pulses each of the same period. The bit streams are reversed when the signal on lead 161 is a "0". The illustration of "1"s and "0"s shown in Figures 16, 17, 18 and 19 is an accurate representation of how this signal will appear for a sigma delta modulation encoder.

It is preferred that the accelerometer of Figure 16, as well as those of Figures 17-19, have very high noise reduction characteristics. In order to achieve such noise reduction the preferred embodiment of Figure 16 and those of Figures 17-19 includes one or more integrators in the forward loop. With such an integrator, for example in conjunction with filter 124, the filter-integrator 124 and binary sampler 150, with feedback of the output signal to the mass 36 to be summed with the input signal applied to support 38, is defined as a sigma-delta-modulation system. It is preferred to have a single integrator in the forward loop, thus achieving a first order sigma-delta-modulation system. Lower or higher order systems may find application with the invention of Figures 16-19. Sigma delta modulation systems are described in an IEEE publication, "Stability Analysis of High-Order Sigma-Delta Modulators", by Ardan and Paulos, CH 2255 - 8/86/0000-0715, 1986.

The output binary bit stream of a sigma delta modulation system is mostly alternating bit patterns with the density of "1"s versus "0"s being the representation of the acceleration input. In the embodiment of Figure 16, the voltage bit streams on leads 164 and 166 are applied to force conducting plates 92' of the top and bottom support members 82, 84.

A constant charge is applied to force conducting plates 92 of mass 36 via charge generator 130 and leads 112, 94. Consequently, the force electric field across the mass 36 is applied in a series of binary voltage pulses to the support member conducting plates to drive mass 36 to its reference position as determined by the d.c. reference potential $V_{01/2}$. The binary bit stream applied to plates 92' reverses in polarity of its bit density as the displacement signal on lead 126 crosses the threshold voltage level applied to threshold circuit 160. Such reversal of polarity of the bit streams on leads 164, 166 causes the mass 36 to be driven in

the opposite direction.

The charge generator circuit 130 is similar in construction to that illustrated in Figure 15 except that a constant voltage V is applied to lead 128 which causes a constant amount of charge to be applied to conducting plates 92 of mass 36.

The output of the transducer of Figure 16 on lead 168 is a serial binary bit stream of 0's or 1's representative of the displacement of the mass 36 with respect to a reference position as specified by the reference potential $V_{01/2}$ applied to amplifier 120. As discussed previously, displacement of the mass 36 is proportional to acceleration applied to support 38 of the mass 36 spring support 38 system.

Figure 17 is an alternative binary bit stream embodiment of the transducer, similar to the transducer of Figure 16, except that a d.c. potential difference $V_{01} - V_{01}$ is applied across force conducting plates 92' of top and bottom plates 82, 84. The binary bit stream output from the Q output of binary sampler 150 is applied to lead 128 to charge generator 130, which in response, applies "1"s or "0"s of charge to force conducting plates 92 via leads 112 and 94. In other words, the embodiment of Figure 17 is the inverse arrangement of that of Figure 16 in that a constant voltage is applied to force plates 92' of the top and bottom plates, but binary charge proportional to displacement of the mass 36 (in response to acceleration applied to support 38) is applied to the force plates 92. As before, mass 36 is driven in a direction toward its reference position as determined by the voltage $(V_{01} - V_{01})/2$ applied to amplifier 120. The output on lead 168, like that of Figure 16, is a serial binary bit stream proportional to acceleration applied to the support 38 of the transducer.

Figure 18 is another alternative arrangement of the embodiment of the invention having a binary bit stream output and feedback which is representative of displacement or acceleration applied to the transducer. The arrangement of Figure 18 is similar to the analog embodiment of Figure 14 in that it applies an a.c. carrier voltage (square wave) to modulate the force electric field on force plates 92' and to sense plates 90' of top and bottom supports 82, 84. The arrangement of Figure 18 is also similar to the binary output arrangement of Figure 17 in that a binary stream of charge pulses, (but in this arrangement modulated by the carrier signal V_{01}) is fed back to force plates 92 of mass 36 in response to the operation of digital sampler 150. Like in Figure 17, the digital sampler 150 produces a binary bit stream on lead 168, but modulated by signal V_{01} , representative of displacement signal 126. Charge generator 130 produces a modulated binary bit stream on leads 112 and 94 in response to voltage bits applied on lead 128 from lead 168.

The modulated bit stream on lead 168 is demodulated by demodulator 153 (in a similar fashion as the analog signal on lead 126 of Figure 14 is demodulated) to produce a binary bit stream on lead 170 representative of displacement or acceleration applied to the transducer.

Figure 19 is similar to Figure 18, but feeds a modulated binary bit stream of voltage pulses back to plates 92' of top and bottom support plates 82, 84 and applies modulated charge, but of constant magnitude, to force plates 92 of mass 36. The result, after demodulation of the modulated binary bit stream signal on lead 152 by demodulator 153 is a serial binary bit stream representation of the displacement or acceleration applied to the transducer system.

Various modifications and alternatives in the described structures will be apparent to those skilled in the art. For example, although the preferred method of fabricating the force transducer is by chemical etching of a semi-conductor wafer, plasma etching may also be used effectively. Although the preferred N⁻ doping of the spring bases and other regions is arsenic, boron or phosphorous may also be used depending upon the original crystal type. For this reason, these changes are desired to be included in the appended claims. The appended claims recite the only limitations to the present invention and the descriptive manner which is employed for setting forth the embodiments is to be incorporated as illustrative and not limitative.

Claims

1. A transducer comprising a movable mass (36) suspended from a support structure (38) and having plates (82,84) fixed to the structure on each side of said mass, each surface of said mass facing one of the plates (82,84) carrying a sense conducting area (90), said plates carrying first and second sense conducting areas (90') each of which faces a corresponding sense conducting area (90) of said mass, whereby a voltage may be applied between said first and second sense conducting areas (90') of said plates so as to induce a voltage on said sense conducting area (90) of said mass indicative of the position of said mass between said first and second conducting areas of said plates,

the movable mass (36) and the support structure (38) being micromachined from a mono-crystalline wafer with said mass being suspended from said support structure (38) by at least two E-leaf springs, said mass being movable between said plates in a direction perpendicular to the plane of the wafer,

means for applying a first voltage (V_{cc})

the opposite direction.

The charge generator circuit 130 is similar in construction to that illustrated in Figure 15 except that a constant voltage V is applied to lead 128 which causes a constant amount of charge to be applied to conducting plates 92 of mass 36.

The output of the transducer of Figure 16 on lead 168 is a serial binary bit stream of 0's or 1's representative of the displacement of the mass 36 with respect to a reference position as specified by the reference potential $V_{01/2}$ applied to amplifier 120. As discussed previously, displacement of the mass 36 is proportional to acceleration applied to support 38 of the mass 36 spring support 38 system.

Figure 17 is an alternative binary bit stream embodiment of the transducer, similar to the transducer of Figure 16, except that a d.c. potential difference $V_{01} - V_{01}$ is applied across force conducting plates 92' of top and bottom plates 82, 84. The binary bit stream output from the Q output of binary sampler 150 is applied to lead 128 to charge generator 130, which in response, applies "1"s or "0"s of charge to force conducting plates 92 via leads 112 and 94. In other words, the embodiment of Figure 17 is the inverse arrangement of that of Figure 16 in that a constant voltage is applied to force plates 92' of the top and bottom plates, but binary charge proportional to displacement of the mass 36 (in response to acceleration applied to support 38) is applied to the force plates 92. As before, mass 36 is driven in a direction toward its reference position as determined by the voltage $(V_{01} - V_{01})/2$ applied to amplifier 120. The output on lead 168, like that of Figure 16, is a serial binary bit stream proportional to acceleration applied to the support 38 of the transducer.

Figure 18 is another alternative arrangement of the embodiment of the invention having a binary bit stream output and feedback which is representative of displacement or acceleration applied to the transducer. The arrangement of Figure 18 is similar to the analog embodiment of Figure 14 in that it applies an a.c. carrier voltage (square wave) to modulate the force electric field on force plates 92' and to sense plates 90' of top and bottom supports 82, 84. The arrangement of Figure 18 is also similar to the binary output arrangement of Figure 17 in that a binary stream of charge pulses, (but in this arrangement modulated by the carrier signal V_{01}) is fed back to force plates 92 of mass 36 in response to the operation of digital sampler 150. Like in Figure 17, the digital sampler 150 produces a binary bit stream on lead 168, but modulated by signal V_{01} , representative of displacement signal 126. Charge generator 130 produces a modulated binary bit stream on leads 112 and 94 in response to voltage bits applied on lead 128 from lead 168.

The modulated bit stream on lead 168 is demodulated by demodulator 153 (in a similar fashion as the analog signal on lead 126 of Figure 14 is demodulated) to produce a binary bit stream on lead 170 representative of displacement or acceleration applied to the transducer.

Figure 19 is similar to Figure 18, but feeds a modulated binary bit stream of voltage pulses back to plates 92' of top and bottom support plates 82, 84 and applies modulated charge, but of constant magnitude, to force plates 92 of mass 36. The result, after demodulation of the modulated binary bit stream signal on lead 152 by demodulator 153 is a serial binary bit stream representation of the displacement or acceleration applied to the transducer system.

Various modifications and alternatives in the described structures will be apparent to those skilled in the art. For example, although the preferred method of fabricating the force transducer is by chemical etching of a semi-conductor wafer, plasma etching may also be used effectively. Although the preferred N⁻ doping of the spring bases and other regions is arsenic, boron or phosphorous may also be used depending upon the original crystal type. For this reason, these changes are desired to be included in the appended claims. The appended claims recite the only limitations to the present invention and the descriptive manner which is employed for setting forth the embodiments is to be incorporated as illustrative and not limitative.

Claims

1. A transducer comprising a movable mass (36) suspended from a support structure (38) and having plates (82,84) fixed to the structure on each side of said mass, each surface of said mass facing one of the plates (82,84) carrying a sense conducting area (90), said plates carrying first and second sense conducting areas (90') each of which faces a corresponding sense conducting area (90) of said mass, whereby a voltage may be applied between said first and second sense conducting areas (90') of said plates so as to induce a voltage on said sense conducting area (90) of said mass indicative of the position of said mass between said first and second conducting areas of said plates,

the movable mass (36) and the support structure (38) being micromachined from a mono-crystalline wafer with said mass being suspended from said support structure (38) by at least two E-leaf springs, said mass being movable between said plates in a direction perpendicular to the plane of the wafer,

means for applying a first voltage (V_{00})



between said first and second sense conducting areas (90") of said plates (82, 84) thereby creating a sense electric field between said sense conducting areas and inducing a sense voltage on said sense conducting area (90) of said mass (36) proportional to the displacement of said mass between said plates,

means (120, 111, 122, 124, 126) responsive to said sense voltage for generating a displacement signal representative of the distance said mass has moved from a reference position between said plates, and
CHARACTERIZED IN THAT,

each surface of said mass (36) facing one of the plates (82,84) further includes a force conducting area (92), electrically isolated from the sense conducting area (90), said plates carrying first and second force conducting areas (92") each of which faces a corresponding force conducting area (92) of said mass, and

in that feedback circuit means responsive to said displacement signal are provided for producing a return force to said mass (36) by applying an amount of charge to said force conducting area (92) of said mass and a force electric field across said first and second force conducting areas (92") of said plates such that a force is created on said mass proportional to the numerical product of said amount of charge and said electric field in a direction to restore said mass to said reference position.

2. A transducer according to claim 1 wherein said mass (36) has opposing surfaces, each surface having a sense conducting area (90) and a force conducting area (92), with said sense conducting areas of said opposing surfaces, being conductively connected and with said force conducting areas of said opposing surfaces being conductively connected.

3. A transducer according to claim 1 or 2 wherein said first voltage and said force electric field applied respectively between said first and second sense and force conducting areas of said plates which face said mass are d.c.

4. A transducer according to claim 1 or 2 wherein said first voltage and said force electric field applied respectively between said first and second sense and force conducting areas of said plates which face said surface of said mass area a.c.

5. A transducer of any of the preceding claims wherein said first voltage and said force electric field are square wave signals.

6. A transducer according to any of the preceding claims wherein said displacement signal is an a.c. signal, the peak to peak amplitude of which is proportional to the distance said mass (36) has moved from said reference position between said plates (82, 84), and for accelerations of frequency less than the resonant frequency of said springs and said mass is substantially proportional to acceleration of said support structure.

7. A transducer of any of the preceding claims further comprising circuit means responsive to said displacement signal for generating a binary output signal representative of said displacement signal.

8. A transducer according to claim 1 wherein said means (120, 111, 122, 124, 126) responsive to the sense voltage includes means responsive to said displacement signal for converting said displacement signal to an output binary bit stream (168), including a binary sampler means (150) for producing said output binary bit stream as a sequence of equal time length bi-level pulses representing "1s" and "0s", and said feedback circuit means includes means responsive to said output binary bit stream for applying bi-level electrostatic force pulses to said mass (36) in synchronism with said "1s" and "0s" of said output binary bit stream and in a direction tending to restore said mass to said reference position, said force pulses having an approximately constant level during each time length of said bi-level pulses, said bi-level force pulses resulting from bi-level charge electrostatic field forces applied between said force conducting area (92) of said mass (36) and said force conducting areas (92") of said plates (82, 84) which face said force conducting area of said mass.

Patentansprüche

1. Meßwandler mit einer beweglichen Masse (36), die von einer Tragkonstruktion (38) hängt und mit Platten (82, 84), die an jeder Seite der genannten Masse an der Konstruktion befestigt sind, wobei jede Oberfläche der genannten Masse zu einer der Platten (82, 84) zeigt und einen Meßdurchlaßbereich (90) trägt, wobei die genannten Platten erste und zweite Meßdurchlaßbereiche (90") tragen, von denen jeder zu einem entsprechenden Meßdurchlaßbereich (90) der genannten Masse zeigt, wobei zwischen den genannten ersten und zweiten Meßdurchlaßbereichen (90") der genannten Platten eine Spannung angelegt werden kann, um so



dem genannten Meßdurchlaßbereich (90) der genannten Masse eine Spannung zu induzieren, welche die Position der genannten Masse zwischen den genannten ersten und zweiten Durchlaßbereichen der genannten Platten anzeigt,

dadurch gekennzeichnet, daß die genannte bewegliche Masse (36) und die Tragkonstruktion (38) aus einer Monokristallin-Halbleiterscheibe mikrobearbeitet sind, wobei die genannte Masse durch mindestens zwei E-Blattfedern von der genannten Tragkonstruktion (38) hängt, wobei die genannte Masse zwischen den genannten Platten in senkrechter Richtung zu der Ebene der Halbleiterscheibe beweglich ist,

gekennzeichnet durch eine Einrichtung zum Anlegen einer ersten Spannung (V_{∞}) zwischen den genannten ersten und zweiten Meßdurchlaßbereichen (90') der genannten Platten (82, 84), wodurch ein elektrisches Meßfeld zwischen den genannten Meßdurchlaßbereichen erzeugt wird und an dem genannten Meßdurchlaßbereich (90) der genannten Masse (36) eine Meßspannung induziert wird, die proportional zu der Verschiebung der genannten Masse zwischen den genannten Platten ist,

gekennzeichnet durch eine Einrichtung (120, 111, 122, 124, 126), die auf die genannte Meßspannung anspricht, um so ein Verschiebungssignal zu erzeugen, welches die Strecke anzeigt, um welche die Masse sich von einer Bezugsposition zwischen den Platten bewegt hat, und

dadurch gekennzeichnet, dass

jede Oberfläche der genannten Masse (36), die zu einer der Platten (82, 84) zeigt, ferner einen Kraftdurchlaßbereich (92) umfaßt, der elektrisch von dem Meßdurchlaßbereich (90) isoliert ist, wobei die genannten Platten erste und zweite Kraftdurchlaßbereiche (92') tragen, von denen jeder zu einem entsprechenden Kraftdurchlaßbereich (92) der genannten Masse zeigt und dadurch gekennzeichnet,

daß auf das genannte Verschiebungssignal ansprechende Rückkopplungskreise bereitgestellt sind, zur Erzeugung einer Rücklaufkraft zu der genannten Masse (36), dadurch, daß dem genannten Kraftdurchlaßbereich (92) der genannten Masse eine Ladung zugeführt wird, und daß ein elektrisches Kraftfeld an den genannten ersten und zweiten Kraftdurchlaßbereichen (92') der genannten Platten angelegt wird, so daß an der genannten Masse eine Kraft erzeugt wird, die proportional zu dem numerischen Produkt der genannten Ladung ist und dem genannten elektrischen Feld, in eine solche Richtung, um die genannte Masse an

die genannte Bezugsposition zurückzuführen.

2. Meßwandler nach Anspruch 1, dadurch gekennzeichnet, daß die genannte Masse (36) gegenüberliegende Oberflächen aufweist, wobei jede Oberfläche einen Meßdurchlaßbereich (90) und einen Kraftdurchlaßbereich (92) aufweist, dadurch gekennzeichnet, daß die genannten Meßdurchlaßbereiche der genannten gegenüberliegenden Oberflächen leitfähig verbunden sind und wobei die genannten Kraftdurchlaßbereiche der genannten gegenüberliegenden Oberflächen leitfähig verbunden sind.
3. Meßwandler nach Anspruch 1 oder 2, dadurch gekennzeichnet, daß die genannte erste Spannung und das genannte elektrische Kraftfeld, die entsprechend zwischen den genannten ersten und zweiten Meß- und Kraftdurchlaßbereichen der genannten Platten angelegt werden, die zu der genannten Masse zeigen, gleichstromartig sind.
4. Meßwandler nach Anspruch 1 oder 2, dadurch gekennzeichnet, daß die genannte erste Spannung und das genannte elektrische Kraftfeld, die entsprechend zwischen den genannten ersten und zweiten Meß- und Kraftdurchlaßbereichen der genannten Platten angelegt werden, die zu der genannten Oberfläche der genannten Masse zeigen, wechselstromartig sind.
5. Meßwandler nach einem der vorstehenden Ansprüche, dadurch gekennzeichnet, daß die genannte erste Spannung und das genannte elektrische Kraftfeld Rechtecksignale darstellen.
6. Meßwandler nach einem der vorstehenden Ansprüche, dadurch gekennzeichnet, daß das Verschiebungssignal ein Wechselstromsignal darstellt, dessen Doppelamplitude proportional zu der Strecke ist, um die sich die genannte Masse (36) zwischen den genannten Platten (82, 84) von der genannten Bezugsposition bewegt hat und für Frequenzbeschleunigungen, die geringer als die Resonanzfrequenz der genannten Feder sind und wobei die genannte Masse im wesentlichen proportional zu der Beschleunigung der genannten Tragkonstruktion ist.
7. Meßwandler nach einem der vorstehenden Ansprüche, ferner umfassend eine Schaltungseinrichtung, die auf das genannte Verschiebungssignal anspricht, um so ein binäres Ausgangssignal zu erzeugen, welches das genannte Verschiebungssignal darstellt.



8. Meßwandler nach Anspruch 1, dadurch gekennzeichnet, daß die genannte Einrichtung (120, 111, 122, 124, 126), die auf die Meßspannung anspricht, eine Einrichtung umfaßt, die auf das genannte Verschiebungssignal anspricht, um das genannte Verschiebungssignal in einen binären Bitausgangsstrom (168) umzuwandeln, umfassend eine binäre Abtasteinrichtung (150) zur Erzeugung des genannten binären Bitausgangsstroms als eine Folge von zweistufigen Impulsen mit gleicher Dauer, welche "1er" und "0er" darstellen und dadurch gekennzeichnet, daß der genannte Rückkopplungskreis eine Einrichtung aufweist, die auf den genannten binären Bitausgangsstrom anspricht, um der genannten Masse (36) zweistufige elektrostatische Kraftimpulse synchron mit "1ern" und "0ern" des genannten binären Bitausgangsstroms zuzuführen, und zwar in eine Richtung, die dazu neigt, die genannte Masse zu der genannten Bezugsposition zurückzuführen, wobei die genannten Kraftimpulse während jeder Dauer der genannten zweistufigen Impulse eine ungefähr gleichen Pegel aufweisen, wobei die genannten zweistufigen Kraftimpulse daraus resultieren, daß zweistufige elektrostatische Ladungsfeldkräfte zwischen dem genannten Kraftdurchlaßbereich (92) der genannten Masse (36) und den genannten Kraftdurchlaßbereichen (92') der genannten Platten (82, 84), die zu dem genannten Kraftdurchlaßbereich der genannten Masse zeigen, angelegt werden.

Revendications

1. Transducteur comprenant une masse mobile (36), suspendue à une structure porteuse (38), et comportant des plaques (82, 84) fixées à la structure de chaque côté de la masse, chaque surface de la masse, faisant face à une des plaques (82, 84) portant une zone conductrice de sens (90), les plaques portant des première et seconde zones conductrices de sens (90') qui font face chacune à une zone conductrice (90) correspondante de la masse, une tension pouvant être appliquée entre les première et seconde zones conductrices de sens (90') des plaques de façon à induire une tension, sur la zone conductrice de sens (90) de la masse, qui indique la position de la masse entre les première et seconde zones conductrices des plaques,

la masse mobile (36) et la structure porteuse (38) étant micro-usinées à partir d'une plaquette monocristalline, la masse étant suspendue à la structure porteuse (38) par au moins deux ressorts en feuille en E, la masse

étant mobile entre les plaques dans une direction perpendiculaire au plan de la plaquette,

des moyens pour appliquer une première tension (V_{cc}) entre les première et seconde zones conductrices de sens (90') des plaques (82, 84) en produisant par cela un champ électrique de sens entre les zones conductrices de sens et en induisant une tension de sens, sur la zone conductrice de sens (90) de la masse (36), proportionnelle au déplacement de la masse entre les plaques,

des moyens (120, 111, 122, 124, 126) sensibles à la tension de sens susdite pour produire un signal de déplacement qui représente la distance dont s'est déplacée la masse à partir d'une position de référence entre les plaques, caractérisé en ce que,

chaque surface de la masse (36) qui fait face à une des plaques (82, 84) comporte en outre une zone conductrice de force (92) électriquement isolée de la zone conductrice de sens (90), lesdites plaques portant des première et seconde zones conductrices de force (92') dont chacune est en face d'une zone conductrice de force (92) correspondante de ladite masse, et

des moyens à circuit de rétroaction sensibles au signal de déplacement susdit sont prévus pour produire une force de retour sur la masse (36) en appliquant une quantité de charge sur la zone conductrice de force (92) de la masse et un champ électrique de force entre les première et seconde zones conductrices de force (92') des plaques susdites de sorte que soit produite sur la masse une force proportionnelle au produit numérique de la quantité de charge et du champ électrique, dans un sens pour ramener la masse à la position de référence.

2. Transducteur suivant la revendication 1, caractérisé en ce que la masse (36) comporte des surfaces opposées, chaque surface ayant une zone conductrice de sens (90) et une zone conductrice de force (92), les zones conductrices de sens des surfaces opposées étant reliées de façon conductrice, les zones conductrices de force des surfaces opposées étant reliées de façon conductrice.

3. Transducteur suivant l'une ou l'autre des revendications 1 et 2, caractérisé en ce que la première tension et le champ électrique de force appliqués respectivement entre les première et seconde zones conductrices de sens et de force des plaques qui font face à la masse sont continus.



4. Transducteur suivant l'une ou l'autre des revendications 1 et 2, caractérisé en ce que la première tension et le champ électrique de force appliqués respectivement entre les première et seconde zones conductrices de sens et de force des plaques qui font face auxdites surfaces de la masse sont alternatifs. 5
5. Transducteur suivant l'une quelconque des revendications précédentes, caractérisé en ce que la première tension et le champ électrique de force sont des signaux à onde carrée. 10
6. Transducteur suivant l'une quelconque des revendications précédentes, caractérisé en ce que le signal de déplacement est un signal alternatif dont l'amplitude de crête à crête est proportionnelle à la distance dont la masse (36) s'est déplacée à partir de la position de référence entre les plaques (82, 84), et pour des accélérations d'une fréquence inférieure à la fréquence de résonance des ressorts et de la masse, est sensiblement proportionnelle à l'accélération de ladite structure porteuse. 15
20
25
7. Transducteur suivant l'une quelconque des revendications précédentes, caractérisé en ce qu'il comprend des moyens de circuit sensibles au signal de déplacement susdit afin de produire un signal de sortie binaire qui représente le signal de déplacement. 30
8. Transducteur suivant la revendication 1, caractérisé en ce que les moyens (120, 111, 122, 124, 126) susdits sensibles à la tension de sens comportent des moyens sensibles au signal de déplacement susdit afin de transformer le signal de déplacement en un train de bits binaire de sortie (168), tout en comprenant des moyens d'échantillonnage binaire (150) pour produire le train de bits binaire de sortie comme séquence d'impulsions à deux niveaux de longueur temporelle égale qui représentent des "1" et des "0", et en ce que les moyens de circuit de rétroaction comportent des moyens sensibles au train de bits binaire de sortie afin d'appliquer des impulsions de force électrostatique à deux niveaux à la masse (36) en synchronisme avec les "1" et les "0" du train de bits binaire de sortie et dans un sens qui tend à ramener la masse à sa position de référence, les impulsions de force ayant un niveau approximativement constant pendant chaque durée de temps des impulsions à deux niveaux, les impulsions de force à deux niveaux découlant de forces de champ électrostatique de charge à deux niveaux appliquées entre la zone conductrice de force (92) de la

masse (36) et les zones conductrices de force (92) des plaques susdites (82, 84) qui sont en face de la zone conductrice de force de la masse.

