

METHOD AND APPARATUS FOR VARYING A CLOCK FREQUENCY ON A PHASE BY PHASE BASIS

BACKGROUND OF THE INVENTION

1. Field of the Invention

The present invention relates generally to the field of clock circuits, and specifically, to a method and apparatus for varying a clock frequency on a phase by phase basis.

2. Background Information

Component testers are often used to test and debug critical speed paths on newly designed microprocessors. A component tester is typically connected to the microprocessor and generates a test clock used to drive one or more functional units (also hereinafter referred to as the core) contained therein. The functional units include, for example, the data path, input units, execution units, cache, output units, and the like.

Clock shrinking is a mechanism by which the frequency of a clock (or group of clocks) is changed dynamically during the execution of a microprocessor. The term "shrinking" is used to denote that the frequency of a clock cycle of interest is reduced relative to other clock cycles. Clock shrinking is a debug tool for testing newly designed microprocessors and other types of integrated circuits. By shrinking a single clock (and leaving the other clocks at a lower, passing frequency), a single critical path can be isolated in a test or diagnostic that contains many critical speed paths.

Normally, in a high frequency microprocessor design, the core clock is generated from a phase-locked loop ("PLL"). The PLL is an analog circuit that generates a perfect, even duty cycle clock from an imperfect, uneven duty cycle

clock across a wide range of process skews, voltages, and temperatures. The PLL also guarantees that the generated clock's rising edge is synchronized with the rising edge of the input clock. The PLL is used primarily to "clean up" a noisy, system bus clock before being driven to the microprocessor core as well as ensure that the core clock is still synchronized to the external clock.

Figure 1 illustrates a prior art test system 100. The test system 100 includes a tester 110 coupled to a microprocessor 130 by way of signal line 120. Under test condition, the tester 110 transmits a test clock along signal line 120 to the microprocessor 130. The test clock typically includes one or more shrunk and/or stretched clocks for testing various functional units within the microprocessor 130. The test clock bypasses the PLL and drives the internal core clock directly. In this manner, a shrunk and/or stretched clock generated is passed directly into the functional units within the microprocessor. If the test clock is fed to the PLL, the PLL will attempt to "clean up" an intentional shrunk and/or stretched clock. The shrunk and/or stretched clock is treated as system noise and will either be re-adjusted back to the original frequency or break the lock of the PLL and cause it to spin out of control.

However, using an external tester to generate a varied clock and bypassing the PLL means that any system noise or inaccuracy is also passed directly into the core. For this reason, clock shrinking and/or stretching has only been possible on expensive component testers, which are capable of generating accurate, high frequency clocks. In recent years, however, the core frequency of microprocessors has been increasing at a faster rate than the capabilities of the component testers which makes it nearly impossible to provide an external clock at today's core frequencies.

Accordingly, there is a need in the technology for a method and apparatus for varying the clock frequency of one or more clocks on a cycle by cycle basis.

SUMMARY OF THE INVENTION

In one embodiment, the present invention is a circuit to vary a frequency of an input clock. The circuit includes a delay generator to generate at least two delayed clocks from the input clock and a select circuit coupled to receive the at least two delayed clocks and provide an output clock from one of the at least two delayed clocks. The select circuit switches the output clock from the one of the at least two delayed clocks to the other of the at least two delayed clocks on a first edge.

BRIEF DESCRIPTION OF THE DRAWINGS

The features and advantages of the present invention will become apparent from the following detailed description of the present invention in which:

Figure 1 illustrates a prior art test system.

Figure 2 illustrates an exemplary embodiment of a microprocessor implementing one embodiment of the present invention.

Figure 3 illustrates an exemplary embodiment of the clock variator circuit of Figure 2.

Figure 4 illustrates an exemplary embodiment of the delay generator circuit of Figure 3.

Figure 5 illustrates an exemplary timing diagram of the inputs and output signals of the multiplexer of Figure 3.

Figure 6A illustrates an exemplary embodiment of the interpolating circuit of Figure 3.

Figure 6B illustrates another exemplary embodiment of the interpolating circuit of Figure 3.

Figure 7 illustrates a timing diagram of an exemplary output clock signal of the interpolating circuit of Figure 6A.

Figure 8 illustrates an exemplary embodiment of a control circuit for controlling the control lines of the multiplexer of Figure 3.

DETAILED DESCRIPTION

The present invention is a method and apparatus for varying a clock frequency on a phase by phase basis. In one implementation, a clock variator circuit is implemented within an integrated circuit (e.g., a microprocessor) and includes a delay generator circuit which generates delayed clocks (versions) of an internal clock (e.g., a phase locked loop "PLL" clock) and a multiplexer which receives the internal clock and the delayed clocks and provides a clock signal having at least one clock phase that is varied from the remainder of the clock phases by selecting specific edges of the internal clock and the delayed clocks.

In one embodiment, the clock variator circuit provides a shrunk clock phase (e.g., a low phase or a high phase of a clock cycle) by switching an output clock from a delayed clock to a second delayed clock on a clock edge. In this embodiment, the clock variator circuit may provide a shrunk clock cycle by switching the output clock from the second delayed clock to a third clock on a second clock edge. The third clock may be the internal clock or a further delayed clock. In another embodiment, the clock variator circuit provides a stretched clock phase (e.g., a low phase or a high phase of a clock cycle) by switching the output clock from a delayed clock to a second delayed clock on a clock edge. In this embodiment, the clock variator circuit may provide a

stretched clock cycle by switching the output clock from the second delayed clock to a third clock on a second clock edge. The third clock may be the internal clock or a further delayed clock.

In the following description, for purposes of explanation, numerous details are set forth in order to provide a thorough understanding of the present invention. However, it will be apparent to one skilled in the art that these specific details are not required in order to practice the present invention.

Figure 2 illustrates an exemplary embodiment of a microprocessor 200 implementing one embodiment of the present invention. Referring to Figure 2, the microprocessor 200 includes a clock generator circuit 210 which receives an external clock XCLK on signal line 214 (typically from an oscillator or a crystal) and provides a CORE CLOCK on signal line 232. The CORE CLOCK is fed to a plurality of functional units 240₁-240_X (where "X" is a positive whole number). The functional units, include, for example, the data path unit, ALU, execution unit, cache, and the like.

The clock generator circuit 210 includes a PLL 202 which receives XCLK and generates a perfect, even duty cycle PLL clock on signal line 216. Since the PLL clock may not guarantee a 50% duty cycle, the PLL clock may optionally be fed to a divider circuit 220 (shown by dashed lines) which divides the frequency of the PLL clock by two, in which case the PLL 202 may generate a clock having twice the frequency. The divider circuit 220 switches every other rising edge of the PLL clock to a falling edge to ensure a 50% duty cycle. The output of the divider circuit 220 is fed back to the PLL 202 to ensure that the rising edge of the clock on signal line 222 (and the CORE CLOCK 232) is synchronous with the rising edge of XCLK on signal line 214. The output of the divider circuit 220 is coupled to a multiplexer 230 which, in normal mode, is selected by control line 234 to provide the PLL clock as the CORE CLOCK.

The PLL 202 is also coupled to a clock variator circuit 218 which receives the PLL clock on signal line 216 and generates a clock on signal line 224 which may have one or more shrunk and/or stretched clock phases. The clock variator circuit 218 internally generates one or more discrete delayed versions of the PLL clock and provides a TEST CLOCK on signal line 224 having at least one clock phase that is varied from the remainder of clock phases by selecting, on an edge basis, from the discrete delayed version and the PLL clock and/or the PLL clock. The TEST CLOCK may optionally be fed to a divider circuit 226 (shown by dashed lines) which divides the TEST CLOCK by two. This reduces the complexity of the clock variator circuit 218 and a control circuit (see, e.g., Figure 8) because only one clock edge (e.g., the rising edge) needs to be varied instead of both the rising and falling edges. The edge that is varied is fed to the divider circuit which makes every effectively varies both edges. In a test mode, (e.g., validation testing, speed path testing/debugging, etc.) the multiplexer 230 is controlled by way of control line 234 to provide the TEST CLOCK as the CORE CLOCK.

Figure 3 illustrates an exemplary embodiment of the clock variator circuit 218 of Figure 2. Referring to Figure 3, the clock variator circuit 218 includes a delay generator circuit 310 which receives the PLL clock on signal line 216 and generates one or more clock signals D_1 - D_M on signal lines 315₁-315_M (where "M" is a positive whole number) that are fed to a multiplexer 320. The one or more clock signals D_1 - D_M are delayed versions of the PLL clock. In addition to the one or more clock signals D_1 - D_M , the PLL clock may also be fed to the multiplexer 320 (not shown in this Figure). In one embodiment, the multiplexer 320 provides a TEST CLOCK on signal line 325 by selecting, on an edge-by-edge basis, one or more of the clock signals D_1 - D_M . It is contemplated that other circuitry may be used in lieu of the multiplexer 320 to emulate its functionality. Thus, circuitry having the functionality of the multiplexer 320 itself may be generally referred to as "select circuitry". The multiplexer 320 is

controlled through control lines 330 by a state machine (not shown) which selects the TEST CLOCK (discussed below). The TEST CLOCK may optionally be fed to an interpolating circuit 340, as shown by dashed lines. An exemplary embodiment of the interpolating circuit 340 and a corresponding timing diagram will be discussed with reference to Figures 6 and 7, respectively.

The control lines 330 may be controlled in various ways (see Figure 8 for an exemplary embodiment of a control circuit for controlling the control lines 330). In one embodiment, the control lines 330 are user programmable where a user can write to a register and select the clock frequency, the clock phase(s) or clock cycle(s) to be shrunk and/or stretched, the duty cycle, etc. In another embodiment, the control lines 330 may be controlled by an internal interrupt routine where if a test pin of the microprocessor is enabled, the microprocessor enters a test mode and executes the internal interrupt routine. In yet another embodiment, the control lines 330 may be controlled by microcode where depending on the instruction to be executed, one or more clock phases and/or clock cycles may be varied. It is to be appreciated that various other techniques may be employed to control the control lines 330 and that the techniques described herein are only exemplary.

Figure 4 illustrates an exemplary embodiment of the delay generator circuit 310 of Figure 3. Referring to Figure 4, the delay generator circuit 310 includes one or more delay circuits 410_1 - 410_M which provide one or more delay clock signals D_1 - D_M on signal lines 315_1 - 315_M . In this embodiment, the delay circuits are buffers. The delay clock signals D_1 - D_M are applied to the multiplexer 320 and to a subsequent delay circuit, with the exception of delay clock signal D_M which is only fed to the multiplexer 320. A dummy load may be coupled to the output of the delay circuit 410_M to match the load seen by the other delay circuits. In this embodiment, the PLL clock on signal line 216 is also fed to the multiplexer 320. It is to be appreciated that the delay generator

circuit 310 is merely exemplary and that other types of delay generator configurations and delay circuits may be used in lieu thereof. In one embodiment, each delay circuit generates the same time delay. However, this is not as restriction of the present invention. In another embodiment, the time delay from one delay circuit to another is different.

Figure 5 illustrates an exemplary timing diagram of the inputs and output signals of the multiplexer 320 of Figure 3. In this embodiment, "M" is chosen to be seven for illustration purposes only. However, in another embodiment, "M" may be any whole number greater than zero. Referring to Figure 5, the PLL clock is the output of the PLL 202 on signal line 216. In this embodiment, the period of the PLL clock is chosen to be $8 \cdot t$ for clarity and illustration purposes only. In this embodiment, delay clock signal D_1 is a delayed version of the PLL clock and is delayed by t . Each subsequent delay clock signal ($D_2, D_3, \dots, D_7$) is delayed by t from the previous delay clock signal ($D_1, D_2, \dots, D_6$).

The multiplexer 320 provides the TEST CLOCK by selecting, on an edge by edge basis, from delay clocks D_1 - D_7 and the PLL clock. Figure 5 shows an exemplary test clock, TEST_CLK1 provided on signal line 224 by the multiplexer 320. In this exemplary embodiment, the TEST_CLK1 signal has the same frequency as the PLL clock with the exception of a single shrunk clock cycle. Up to time 510, the TEST_CLK1 signal is the D_5 clock signal. Instead of having a falling edge at time 514 (which is the falling edge of the D_5 clock signal), the TEST_CLK1 signal is controlled to have a falling edge at time 512 (1/8th of a clock cycle earlier), which is the falling edge of the D_4 clock signal. Similarly, instead of having a rising edge at time 518 (which is the rising edge of the D_4 clock signal), the TEST_CLK1 signal is controlled to have a rising edge at time 516 (1/8th of a clock cycle earlier), which is the rising edge of the D_3 clock signal. Thereafter, the TEST_CLK1 signal is the D_3 clock signal. As

such, the shrunk clock cycle from time 510 to time 516 has a period of $6 \cdot t$, which is shrunk by 25% from the other clock cycles.

The present invention may also shrink and/or stretch a single clock phase rather than a whole clock cycle. The TEST_CLK2 signal exemplifies a clock signal having a single shrunk clock phase. Up to time 502, the TEST_CLK2 signal is the D₆ clock signal. Instead of having a rising edge at time 506 (which is the rising edge of the D₆ clock signal), the TEST_CLK2 signal is controlled to have a rising edge at time 504 (1/4th of a clock cycle earlier), which is the rising edge of the D₄ clock signal. Thereafter, the TEST_CLK2 signal is the D₄ clock signal. As such, the period of the shrunk clock phase from time 502 to time 504 is $2 \cdot t$. As can be seen in this example, the clock variator circuit of the present invention has shrunk a single clock phase by 50%.

The TEST_CLK3 signal has the same frequency as the PLL clock with the exception of a single shrunk clock cycle. In this case, the period of the shrunk clock is smaller than the shrunk clock of the TEST_CLK1 signal. Up to time 522, the TEST_CLK3 signal is the D₃ clock signal. Instead of having a falling edge at time 528 (which is the falling edge of the D₃ clock signal), the TEST_CLK3 signal is controlled to have a falling edge at time 524 (3/8th of a clock cycle earlier), which is the falling edge of the PLL clock. Instead of having a rising edge at time 530 (which is the rising edge of the PLL clock), the TEST_CLK3 signal is controlled to have a rising edge at time 526 (3/8th of a clock cycle earlier), which is the rising edge of the D₅ clock signal. Thereafter, the TEST_CLK3 signal is the D₅ clock signal. As such, the period of the shrunk clock cycle from time 522 to time 526 is $2 \cdot t$. As can be seen in this example, the clock variator circuit of the present invention has shrunk a single clock cycle by 75%.

In the exemplary embodiment shown where seven delay signals were generated, the granularity for a shrunk (or stretched) clock cycle is 1/4 of the

PLL clock. This may not be sufficient for speed path testing or other applications. To achieve finer granularity, the delay generator circuit 310 may generate more delayed versions of the PLL clock (e.g., by increasing "M"). However, in this embodiment, the granularity may be limited by the delay of the delay circuits, which may not be sufficient for all test purposes. Alternatively, the interpolating circuit 340 of Figure 3 may be used to provide finer granularity.

Figure 6A illustrates an exemplary embodiment of the interpolating circuit 340 of Figure 3. The interpolating circuit 340 provides two main functions. First, the interpolating circuit 340 interpolates a plurality of fine "tics" between each clock cycle. Second, the interpolating circuit 340 multiplexes out the clock on signal line 224. To that end, in one embodiment, the interpolating circuit 340 may be used in lieu of the delay generator circuit 310 and the multiplexer 320 to provide one or more varied clock phases. That is, the interpolating circuit 340 is another embodiment of the present invention with finer granularity. In another embodiment, the interpolating circuit 340 may be used in addition to the delay generator circuit 310 and the multiplexer 320. Referring to Figure 6A, the TEST CLOCK signal on signal line 325 is fed to the gates of P-channel device 604 and N-channel device 606. Devices 604 and 606 act as an inverter.

When the TEST CLOCK signal is low, device 604 turns on and device 606 turns off such that the signal on signal line 612 is pulled high. On the other hand, when the TEST CLOCK signal is high, the device 604 is turned off and the device 606 is turned on. The timing of the signal on signal line 612 depends on the state of the control lines DN_1 - DN_T on signal lines 610₁-610_T (where "T" is a positive whole number). For sake of clarity, "T" is chosen to be three. For example, when the TEST CLOCK signal goes high, if only DN_1 is high, device 608₁ turns on to pull the signal on signal line 612 low (a first "tic"). If DN_1 and

DN₂ are both high when device 606 turns on, then device 606 turns on harder and pulls the signal on signal line 612 low even faster (a second "tic"). If DN₁- DN₃ are all high when device 606 turns on, then the signal on signal line 612 is pulled low even faster (a third "tic"). In this manner, the timing between "tics" may be made even smaller than the smallest delay provided by the combination of the delay generator circuit 310 and the multiplexer 320. It is important to note that the time period between "tics" does not necessarily have to be equal and is a matter of design choice. For finer granularity, T may be increased.

The embodiment shown in Figure 6A is used to vary the delay between the output signal on signal line 224 and the input signal on signal line 325 when the input signal switches from a falling edge to a rising edge. However, to vary the delay between the output signal and the input signal when the input signal switches from a rising edge to a falling edge, the interpolator circuit is modified as shown in Figure 6B. The embodiments shown in Figures 6A and 6B may be combined in a single circuit to vary the delay between the output signal and the input signal when the input signal switched either from a falling edge to a rising edge or from a rising edge to a falling edge.

Figure 7 illustrates a timing diagram of an exemplary output clock signal of the interpolating circuit 340 of Figure 6A. Referring to Figure 7, the PLL clock signal and the TEST_CLK3 signal are shown for reference purposes. In Figure 5, the PLL clock was chosen to have a period of 8*t with "M" equal to seven with each delayed clock being delayed by t amount from the previous delayed clock. The TICS signal shows the available tics for the case where "T" is chosen to be equal three. For a high phase of the PLL clock, 12 tics are available. Since the delay generator circuit 310 and multiplexer 320 in Figures 3 and 4 provide a granularity of 1/4th of the PLL clock, there are 12/4 or three

tics interpolated for each 1/4th of a clock phase to yield a total granularity of 1/12th of the PLL clock.

An exemplary test clock, TEST_CLK4 is provided on signal line 224 by the clock variator circuit 218. Up to time 702, the TEST_CLK4 signal is the TEST_CLK3. Instead of having a falling edge at time 706 (which is the falling edge of the TEST_CLK3 signal), the TEST_CLK4 signal is controlled (e.g., by only turning on device 608₁) to have a falling edge at time 704 (one "tic" earlier). Similarly, instead of having a rising edge at time 710 (which is the rising edge of the TEST_CLK3 signal), the TEST_CLK4 signal is controlled to have a rising edge at time 708 (two "tics" earlier). As such, the shrunk clock cycle from time 702 to time 708 has a period of $2/3 \cdot t$, which is shrunk by 33% from the shrunk clock of the TEST_CLK3 signal.

Figure 8 illustrates an exemplary embodiment of a control circuit 800 for controlling the control lines 330 of the multiplexer 320 of Figure 3. Referring to Figures 3 and 8, the control circuit 800 includes a register file 810 containing Y entries (where "Y" is a positive whole number), where each entry in the register file 810 selects a specific clock (delayed or the original clock) to be provided on signal line 325 by controlling the control lines 330. The control circuit 800 further includes a clock pointer 820 that can be changed to point to any entry in the register file 810 on a phase by phase basis. Each time the clock pointer 820 changes to a new entry, the output clock is changes to a different clock to change the frequency of at least one phase.

Referring to Figures 5 and 8, an example is provided for sake of clarity. Initially, entries R0 and R1 are loaded to select the D₄ and D₃ clocks respectively. During execution, the clock pointer is set to point to entry R0 which causes the output clock to be the D₄ clock. At a point where a shrunk clock is desired (e.g., time 512), the clock pointer is incremented to point to entry R1 and the next output clock edge is generated from D₃ (time 516), which

is earlier than the next clock edge of D_4 . Thereafter, the output clock is D_3 (unless the pointer is changed again). The control circuit 800 presupposes a mechanism to load one or more entries in the register file and a state machine to increment the clock pointer 820 at the desired time.

It is be appreciated to one skilled in the art that although the exemplary embodiments of the present invention has been described within a microprocessor, the present invention may be implemented in various types of integrated circuits such as, but not limited or restricted to, a microcontroller, an application specific integrated circuit ("ASIC"), a graphics chip, a sequential circuit, and the like.

In addition to providing speed path debugging, the present invention may be used to optimize the clock usage of an integrated circuit. In this embodiment, the clock variator circuit provides the CORE CLOCK for both the normal mode and the test mode. During normal mode, the CORE CLOCK may be varied for certain instructions (e.g., by controlling the clock variator circuit through microcode). For example, if an instruction exercises a functional units (e.g., circuit) that fails with the CORE CLOCK, the CORE CLOCK can be stretched for the clock cycles for that instruction. Conversely, if an instruction exercises a functional unit that operates at a higher frequency than the CORE CLOCK, the CORE CLOCK can be shrunk for the clock cycles for that instruction. This technique allows a CORE CLOCK to be intelligently varied to optimize the speed of the integrated circuit. A further advantage is that the divider circuit 220 and the multiplexer 230 are not needed.

Because a varied clock can feedback to the PLL and cause it to lose its lock, a separate feedback loop is provided for when the clock is varied. During normal PLL operation, the actual clock spine output is fed back into the PLL to help maintain the phase lock. Introducing a large enough shrink and/or stretch into this network could cause the PLL to jitter or even lose lock. To

prevent this, during clock shrinking and/or clock stretching, a separate loop is fed back to the PLL. While the clock spine sees a varied clock, this separate loop is shielded from the varied clock. In this way, the PLL maintains its lock.

As discussed above, the clock variator circuit of the present invention provides the ability to change the frequency domain of a microprocessor (or any other synchronous IC design) during the normal runtime of the chip. This is a powerful design-for-testability feature that can be used, for example, to "shrink" the period of a single clock, thereby stressing any functional units.

While certain exemplary embodiments have been described and shown in the accompanying drawings, it is to be understood that such embodiments are merely illustrative of and not restrictive on the broad invention, and that this invention not be limited to the specific constructions and arrangements shown and described, since various other modifications may occur to those ordinarily skilled in the art.

CLAIMS

What is claimed is:

1. A circuit to vary an input clock frequency on a phase by phase basis, comprising:
a delay generator to generate at least two delayed clocks from the input clock; and
a select circuit coupled to receive the at least two delayed clocks and provide an output clock from one of the at least two delayed clocks, said select circuit to switch the output clock from the one of the at least two delayed clocks to the other of the at least two delayed clocks on a first edge.
2. The circuit of claim 1 wherein the select circuit to provide an output clock from a first delayed clock, said select circuit to switch the output clock from the first delayed clock to a second delayed clock on a first edge, the second delayed clock being delayed less than the first delayed clock with respect to the input clock.
3. The circuit of claim 1 wherein the first edge is a falling edge.
4. The circuit of claim 1 wherein the first edge is a rising edge.
5. The circuit of claim 1 wherein the select circuit to switch the output clock from the one of the at least two delayed clocks to the other of the at least two delayed clocks on a first edge to provide a shrunk clock phase.
6. The circuit of claim 1 wherein the select circuit to switch the output clock from the one of the at least two delayed clocks to the other of the at least two delayed clocks on a first edge to provide a stretched clock phase.



7. The circuit of claim 1 wherein the select circuit to subsequently switch the output clock from the other of the at least two delayed clocks to the input clock on a second edge to provide a shrunk clock cycle.
8. The circuit of claim 7 wherein the first edge is a falling edge and the second edge is a rising edge.
9. The circuit of claim 7 wherein the first edge is a rising edge and the second edge is a falling edge.
10. The circuit of claim 1 wherein the delay generator includes:
 - a first delay circuit coupled to receive the input clock and provide a first delayed clock; and
 - a second delay circuit coupled to receive the first delayed clock and provide a second delayed clock.
11. The circuit of claim 1 further comprising an interpolating circuit coupled to receive the output clock, interpolate one of a plurality of ticks within the output clock, and provide a modified output clock.
12. The circuit of claim 1 wherein the select circuit is a multiplexer.
13. A method of varying a clock phase frequency of an input clock, comprising:
 - generating at least two delayed clocks from the input clock;
 - providing an output clock from one of the at least two delayed clocks;
 - and
 - switching the output clock from the one of the at least two delayed clocks to the other of the at least two delayed clocks on a first edge.



14. The method of claim 13 wherein switching the output clock comprises switching the output clock from the first delayed clock to a second delayed clock on the first edge where the second delayed clock is delayed less than the first delayed clock with respect to the input clock.

15. The method of claim 13 further comprising subsequently switching the output clock from the other of the at least two delayed clocks to the input clock on a second edge to provide a shrunk clock cycle.

16. The method of claim 15 wherein the first edge is a falling edge and the second edge is a rising edge.

17. The method of claim 15 wherein the first edge is a rising edge and the second edge is a falling edge.

18. The method of claim 13 further comprising interpolating one of a plurality of ticks in the output clock to provide finer granularity.

19. A circuit, comprising:
a delay generator to generate at least one delayed clock from an input clock; and
a select element coupled to receive the at least one delayed clock and the input clock, to provide an output clock from one of the at least one delayed clock and the input clock, and to switch the output clock from the one of the at least one delayed clock and the input clock to the other of the at least one delayed clock and the input clock on a first edge.



20. The circuit of claim 19 wherein the select circuit to provide an output clock from a first delayed clock, said select circuit to switch the output clock from the first delayed clock to a second delayed clock on a first edge, the second delayed clock being delayed less than the first delayed clock with respect to the input clock.

21. The circuit of claim 19 wherein the select circuit to switch the output clock from one of the at least one delayed clock and the input clock to the other of the at least one delayed clock and the input clock on a first edge to provide a shrunk clock phase.

22. The circuit of claim 19 wherein the select circuit to switch the output clock from one of the at least one delayed clock and the input clock to the other of the at least one delayed clock and the input clock on a first edge to provide a stretched clock phase.

23. A circuit to vary an input clock frequency for one or more phases, comprising:

a circuit coupled to receive the input clock and provide an output clock;
at least one control terminal; and

one or more devices coupled to the circuit and to the at least one control terminal, wherein the delay between the output clock and the input clock being dependent on the number of the one or more devices turned on when the input clock switches from a first clock edge to a second clock edge.

24. The circuit of claim 23 wherein the first clock edge is a low clock edge and the second clock edge is a high clock edge.



25. The circuit of claim 23 wherein the first clock edge is a high clock edge and the second clock edge is a low clock edge.

26. The circuit of claim 23 wherein the circuit comprises a first device of a first conductivity type having a gate to receive the input clock, a drain to provide the output clock, and a source.

27. The circuit of claim 26 wherein the one or more devices are devices of the first conductivity type each having a gate coupled to a respective control terminal, a drain coupled to the source of the first device, and a source coupled to a circuit ground.

28. The circuit of claim 26 wherein the one or more devices are devices of the first conductivity type each having a gate coupled to a respective control terminal, a drain coupled to the source of the first device, and a source coupled to a power supply terminal having a voltage thereon.

