

DESCRIPTION

SIGNAL PROCESSING APPARATUS AND SIGNAL PROCESSING METHOD,
ENCODER AND ENCODING METHOD, DECODER AND DECODING METHOD, AND
PROGRAM

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TECHNICAL FIELD

[0001]

The present invention relates to a signal processing
apparatus and a signal processing method, an encoder and an
encoding method, a decoder and a decoding method, and a program,
10 and more particularly to a signal processing apparatus and
a signal processing method, an encoder and an encoding method,
a decoder and a decoding method, and a program for reproducing
a music signal with improved sound quality by expansion of
15 a frequency band.

BACKGROUND ART

[0002]

Recently, music distribution services for distributing
20 music data via the internet have been increased. The music
distribution service distributes, as music data, encoded data
obtained by encoding a music signal. As an encoding method
of the music signal, an encoding method has been commonly used
in which the encoded data file size is suppressed to decrease
25 a bit rate so as to save time during download.

[0003]

Such an encoding method of the music signal is broadly
divided into an encoding method such as MP3 (MPEG (Moving
Picture Experts Group) Audio Layers 3) (International Standard
30 ISO/IEC 11172-3) and an encoding method such as HE-AAC (High
Efficiency MPEG4 AAC) (International Standard ISO/IEC

14496-3).

[0004]

The encoding method represented by MP3 cancels a signal component of a high frequency band (hereinafter, referred to as a high band) having about 15 kHz or more in music signal that is almost imperceptible to humans, and encodes the low frequency band (hereinafter, referred to as a low band) of the signal component of the remainder. Therefore, the encoding method is referred to as a high band cancelation encoding method. This kind of high band cancelation encoding method can suppress the file size of encoded data. However, since sound in a high band can be perceived slightly by human, if sound is produced and output from the decoded music signal obtained by decoding the encoded data, suffers a loss of sound quality whereby a sense of realism of an original sound is lost and a sound quality deterioration such a blur of sound occurs.

[0005]

Unlike this, the encoding method represented by HE-AAC extracts specific information from a signal component of the high band and encodes the information in conjunction with a signal component of the low band. The encoding method is referred to below as a high band characteristic encoding method. Since the high band characteristic encoding method encodes only characteristic information of the signal component of the high band as information on the signal component of the high band, deterioration of sound quality is suppressed and encoding efficiency can be improved.

[0006]

In decoding data encoded by the high band characteristic encoding method, the signal component of the low band and

characteristic information are decoded and the signal component of the high band is produced from a signal component of the low band and characteristic information after being decoded. Accordingly, a technology that expands a frequency band of the signal component of the high band by producing a signal component of the high band from signal component of the low band is referred to as a band expansion technology. [0007]

As an application example of a band expansion method, after decoding of data encoded by a high band cancelation encoding method, a post process is performed. In the post process, the high band signal component lost in the encoding is generated from the decoded low band signal component, thereby expanding the frequency band of the signal component of the low band (see Patent Document 1). The method of frequency band expansion of the related art is referred below to as a band expansion method of Patent Document 1. [0008]

In a band expansion method of the Patent Document 1, the apparatus estimates a power spectrum (hereinafter, suitably referred to as a frequency envelope of the high band) of the high band from the power spectrum of an input signal by setting the signal component of the low band after decoding as the input signal and produces the signal component of the high band having the frequency envelope of the high band from the signal component of the low band. [0009]

Fig. 1 illustrates an example of a power spectrum of the low band after the decoding as an input signal and a frequency envelope of an estimated high band. [0010]

In Fig. 1, the vertical axis illustrates a power as a logarithm and a horizontal axis illustrates a frequency.
[0011]

The apparatus determines the band in the low band of
5 the signal component of the high band (hereinafter, referred
to as an expansion start band) from a kind of an encoding system
on the input signal and information such as a sampling rate,
a bit rate and the like (hereinafter, referred to as side
information). Next, the apparatus divides the input signal
10 as signal component of the low band into a plurality of sub-band
signals. The apparatus obtains a plurality of sub-band
signals after division, that is, an average of respective
groups (hereinafter, referred to as a group power) in a time
direction of each power of a plurality of sub-band signals
15 of a low band side lower than the expansion start band is obtained
(hereinafter, simply referred to as a low band side). As
illustrated in Fig. 1, according to the apparatus, it is assumed
that the average of respective group powers of the signals
of a plurality of sub-bands of the low band side is a power
20 and a point making a frequency of a lower end of the expansion
start band be a frequency is a starting point. The apparatus
estimates a primary straight line of a predetermined slope
passing through the starting point as the frequency envelope
of the high band higher than the expansion start band
25 (hereinafter, simply referred to as a high band side). In
addition, a position in a power direction of the starting point
may be adjusted by a user. The apparatus produces each of
a plurality of signals of a sub-band of the high band side
from a plurality of signals of a sub-band of the low band side
30 to be an estimated frequency envelope of the high band side.
The apparatus adds a plurality of the produced signals of the

sub-band of the high band side to each other into the signal components of the high band and adds the signal components of the low band to each other to output the added signal components. Therefore, the music signal after expansion of the frequency band is close to the original music signal. However, it is possible to produce the music signal of a better quality.

[0012]

The band expansion method disclosed in the Patent Document 1 has an advantage that the frequency band can be expanded for the music signal after decoding of the encoded data with respect to various high band cancelation encoding methods and encoded data of various bit rates.

15 CITATION LIST

PATENT DOCUMENT

[0013]

Patent Document 1: Japanese Patent Application Laid-Open No. 2008-139844

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SUMMARY OF THE INVENTION

PROBLEMS TO BE SOLVED BY THE INVENTION

[0014]

Accordingly, the band expansion method disclosed in Patent Document 1 may be improved in that the estimated frequency envelope of a high band side is a primary straight line of a predetermined slope, that is, a shape of the frequency envelope is fixed.

[0015]

30 In other words, the power spectrum of the music signal has various shapes and the music signal has a lot of cases

where the frequency envelope of the high band side estimated by the band expansion method disclosed in Patent Document 1 deviates considerably.

[0016]

5 Fig. 2 illustrates an example of an original power spectrum of an attack music signal (attack music signal) having a rapid change in time as a drum is strongly hit once.

[0017]

10 In addition, Fig. 2 also illustrates the frequency envelope of the high band side estimated from the input signal by setting the signal component of the low band side of the attack relative music signal as an input signal by the band expansion method disclosed in the Patent Document 1.

[0018]

15 As illustrated in Fig. 2, the power spectrum of the original high band side of the attack music signal has a substantially flat shape.

[0019]

20 Unlike this, the estimated frequency envelope of the high band side has a predetermined negative slope and even if the frequency is adjusted to have the power close to the original power spectrum, difference between the power and the original power spectrum becomes large as the frequency becomes high.

25 [0020]

 Accordingly, in the band expansion method disclosed in Patent Document 1, the estimated frequency envelope of the high band side cannot reproduce the frequency envelope of the original high band side with high accuracy. Therefore, if
30 sound from the music signal after the expansion of the frequency band is produced and output, clarity of the sound in auditory

is lower than the original sound.

[0021]

In addition, in the high band characteristic encoding method such as HE-AAC and the like described above, the frequency envelope of the high band side is used as characteristic information of the encoded high band signal components. However, it needs to reproduce the frequency envelope of the original high band side with high accuracy in a decoding side.

10 [0022]

The present invention has been made in a consideration of such a circumstance and provides a music signal having a better sound quality by expanding a frequency band.

15 SOLUTIONS TO PROBLEMS

[0023]

A signal processing apparatus according to a first aspect of the present invention includes: a demultiplexing unit that demultiplexes input encoded data to at least low band encoded data and coefficient information; a low band decoding unit that decodes the low band encoded data to produce low band signals; a selection unit that selects a coefficient table which is obtained based on the coefficient information among a plurality of coefficient tables used for the production of high band signals and having coefficients for the respective sub-bands on a high band side; an extension and reduction unit that deletes the coefficients of some sub-bands to reduce the coefficient table or produces the coefficients of predetermined sub-bands based on the coefficients of some sub-bands to extend the coefficient table; a high band sub-band power calculation unit that calculates high band sub-band

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powers of high band sub-band signals of the respective sub-bands constituting the high band signals based on low band sub-band signals of the respective sub-bands constituting the lowband signals and the extended or reduced coefficient table;
5 and a high band signal production unit that produces the high band signals based on the high band sub-band powers and the low band sub-band signals.

[0024]

The extension and reduction unit may duplicate the
10 coefficients of a sub-band having a highest frequency which is included in the coefficient table and set the duplicated coefficients to coefficients of a sub-band having a higher frequency than the highest frequency to extend the coefficient table.

15 [0025]

The extension and reduction unit may delete the coefficients of a sub-band, which has a higher frequency than that of a sub-band having a highest frequency among sub-bands of the high band sub-band signals, from the coefficient table
20 to reduce the coefficient table.

[0026]

A signal processing method or a program according to the first aspect of the invention includes the steps of demultiplexing input encoded data to at least low band encoded
25 data and coefficient information; decoding the low band encoded data to produce low band signals; selecting a coefficient table which is obtained based on the coefficient information among a plurality of coefficient tables used for the production of high band signals and having coefficients for the respective
30 sub-bands on a high band side; deleting the coefficients of some sub-bands to reduce the coefficient table or generating

the coefficients of predetermined sub-bands based on the coefficients of some sub-bands to extend the coefficient table; calculating high band sub-band powers of high band sub-band signals of the respective sub-bands constituting the high band signals based on low band sub-band signals of the respective sub-bands constituting the low band signals and the extended or reduced coefficient table; and generating the high band signals based on the high band sub-band powers and the low band sub-band signals.

10 [0027]

According to the first aspect of the invention, input encoded data is demultiplexed to at least low band encoded data and coefficient information; the low band encoded data is decoded to produce low band signals; a coefficient table which is obtained based on the coefficient information is selected among a plurality of coefficient tables used for the production of high band signals and having coefficients for the respective sub-bands on a high band side; the coefficients of some sub-bands are deleted to reduce the coefficient table or the coefficients of predetermined sub-bands are produced based on the coefficients of some sub-bands to extend the coefficient table; high band sub-band powers of high band sub-band signals of the respective sub-bands constituting the high band signals are calculated based on low band sub-band signals of the respective sub-bands constituting the low band signals and the extended or reduced coefficient table; and the high band signals are produced based on the high band sub-band powers and the low band sub-band signals.

25 [0028]

30 A signal processing apparatus according to a second aspect of the present invention includes: a sub-band division

unit that produces low band sub-band signals of a plurality of sub-bands on a low band side of an input signal and high band sub-band signals of a plurality of sub-bands on a high band side of the input signal; an extension and reduction unit
5 that deletes the coefficients of some sub-bands to reduce a coefficient table or produces coefficients of predetermined sub-bands based on coefficients of some sub-bands to extend a coefficient table, the coefficient table having the coefficients for the respective sub-bands on the high band
10 side; a pseudo high band sub-band power calculation unit that calculates pseudo high band sub-band powers, which are estimated values of powers of the high band sub-band signals, for the respective sub-bands on the high band side based on the extended or reduced coefficient table and the low band
15 sub-band signals; a selection unit that compares high band sub-band powers of the high band sub-band signals and the pseudo high band sub-band powers to each other and selects one of a plurality of the coefficient tables; and a production unit that produces data containing coefficient information for
20 obtaining the selected coefficient table.

[0029]

The extension and reduction unit may duplicate the coefficients of a sub-band having a highest frequency which is included in the coefficient table and set the duplicated
25 coefficients to coefficients of a sub-band having a higher frequency than the highest frequency to extend the coefficient table.

[0030]

The extension and reduction unit may delete the
30 coefficients of a sub-band, which has a higher frequency than that of a sub-band having a highest frequency among sub-bands

of the high band sub-band signals, from the coefficient table to reduce the coefficient table.

[0031]

A signal processing method or a program according to the second aspect of the invention includes the steps of generating low band sub-band signals of a plurality of sub-bands on a low band side of an input signal and high band sub-band signals of a plurality of sub-bands on a high band side of the input signal; deleting the coefficients of some sub-bands to reduce a coefficient table or generating coefficients of predetermined sub-bands based on coefficients of some sub-bands to extend a coefficient table, the coefficient table having the coefficients for the respective sub-bands on the high band side; calculating pseudo high band sub-band powers, which are estimated values of powers of the high band sub-band signals, for the respective sub-bands on the high band side based on the extended or reduced coefficient table and the low band sub-band signals; comparing high band sub-band powers of the high band sub-band signals and the pseudo high band sub-band powers to each other and selecting one of a plurality of the coefficient tables; and generating data containing coefficient information for obtaining the selected coefficient table.

[0032]

According to the second aspect of the invention, low band sub-band signals of a plurality of sub-bands on a low band side of an input signal and high band sub-band signals of a plurality of sub-bands on a high band side of the input signal are produced; the coefficients of some sub-bands are deleted to reduce a coefficient table or coefficients of predetermined sub-bands are produced based on coefficients

of some sub-bands to extend a coefficient table, the coefficient table having the coefficients for the respective sub-bands on the high band side; pseudo high band sub-band powers, which are estimated values of powers of the high band sub-band signals, are calculated for the respective sub-bands on the high band side based on the extended or reduced coefficient table and the low band sub-band signals; high band sub-band powers of the high band sub-band signals and the pseudo high band sub-band powers are compared to each other and one of a plurality of the coefficient tables is selected; and data containing coefficient information for obtaining the selected coefficient table is produced.

[0033]

A decoder according to a third aspect of the present invention includes: a demultiplexing unit that demultiplexes input encoded data to at least low band encoded data and coefficient information; a low band decoding unit that decodes the low band encoded data to produce low band signals; a selection unit that selects a coefficient table which is obtained based on the coefficient information among a plurality of coefficient tables used for the production of high band signals and having coefficients for the respective sub-bands on a high band side; an extension and reduction unit that deletes the coefficients of some sub-bands to reduce the coefficient table or produces the coefficients of predetermined sub-bands based on the coefficients of some sub-bands to extend the coefficient table; a high band sub-band power calculation unit that calculates high band sub-band powers of high band sub-band signals of the respective sub-bands constituting the high band signals based on low band sub-band signals of the respective sub-bands constituting the low band signals and the extended

or reduced coefficient table; a high band signal production unit that produces the high band signals based on the high band sub-band powers and the low band sub-band signals; and a synthesis unit that synthesizes the low band signal and the high band signal with each other to produce an output signal.
[0034]

A decoding method according to the third aspect of the invention includes the steps of demultiplexing input encoded data to at least low band encoded data and coefficient information; decoding the low band encoded data to produce low band signals; selecting a coefficient table which is obtained based on the coefficient information among a plurality of coefficient tables used for the production of high band signals and having coefficients for the respective sub-bands on a high band side; deleting the coefficients of some sub-bands to reduce the coefficient table or generating the coefficients of predetermined sub-bands based on the coefficients of some sub-bands to extend the coefficient table; calculating high band sub-band powers of high band sub-band signals of the respective sub-bands constituting the high band signals based on low band sub-band signals of the respective sub-bands constituting the low band signals and the extended or reduced coefficient table; generating the high band signals based on the high band sub-band powers and the low band sub-band signals; and synthesizing the low band signal and the high band signal with each other to produce an output signal.
[0035]

According to the third aspect of the invention, input encoded data is demultiplexed to at least low band encoded data and coefficient information; the low band encoded data is decoded to produce low band signals; a coefficient table

which is obtained based on the coefficient information is selected among a plurality of coefficient tables used for the production of high band signals and having coefficients for the respective sub-bands on a high band side; the coefficients
5 of some sub-bands are deleted to reduce the coefficient table or the coefficients of predetermined sub-bands are produced based on the coefficients of some sub-bands to extend the coefficient table; high band sub-band powers of high band sub-band signals of the respective sub-bands constituting the
10 high band signals are calculated based on low band sub-band signals of the respective sub-bands constituting the low band signals and the extended or reduced coefficient table; the high band signals are produced based on the high band sub-band powers and the low band sub-band signals; and the low band
15 signal and the high band signal are synthesized with each other to produce an output signal.

[0036]

An encoder according to a fourth aspect of the present invention includes: a sub-band division unit that produces
20 low band sub-band signals of a plurality of sub-bands on a low band side of an input signal and high band sub-band signals of a plurality of sub-bands on a high band side of the input signal; an extension and reduction unit that deletes the coefficients of some sub-bands to reduce a coefficient table
25 or produces coefficients of predetermined sub-bands based on coefficients of some sub-bands to extend a coefficient table, the coefficient table having coefficients for the respective sub-bands on the high band side; a pseudo high band sub-band power calculation unit that calculates pseudo high band
30 sub-band powers, which are estimated values of powers of the high band sub-band signals, for the respective sub-bands on

the high band side based on the extended or reduced coefficient table and the low band sub-band signals; a selection unit that compares high band sub-band powers of the high band sub-band signals and the pseudo high band sub-band powers to each other and selects one of a plurality of the coefficient tables; a high band encoding unit that encodes coefficient information for obtaining the selected coefficient table to produce high band encoded data; a low band encoding unit that encodes low band signals of the input signal to produce low band encoded data; and a multiplexing unit that multiplexes the low band encoded data and the high band encoded data to produce an output code string.

[0037]

An encoding method according the fourth aspect of the invention includes the steps of generating low band sub-band signals of a plurality of sub-bands on a low band side of an input signal and high band sub-band signals of a plurality of sub-bands on a high band side of the input signal; deleting the coefficients of some sub-bands to reduce a coefficient table or generating coefficients of predetermined sub-bands based on coefficients of some sub-bands to extend a coefficient table, the coefficient table having coefficients for the respective sub-bands on the high band side; calculating pseudo high band sub-band powers, which are estimated values of powers of the high band sub-band signals, for the respective sub-bands on the high band side based on the extended or reduced coefficient table and the low band sub-band signals; comparing high band sub-band powers of the high band sub-band signals and the pseudo high band sub-band powers to each other and selecting one of a plurality of the coefficient tables; encoding coefficient information for obtaining the selected

coefficient table to produce high band encoded data; encoding low band signals of the input signal to produce low band encoded data; and multiplexing the low band encoded data and the high band encoded data to produce an output code string.

5 [0038]

According to the fourth aspect of the invention, low band sub-band signals of a plurality of sub-bands on a low band side of an input signal and high band sub-band signals of a plurality of sub-bands on a high band side of the input
10 signal are produced; the coefficients of some sub-bands are deleted to reduce a coefficient table or coefficients of predetermined sub-bands are produced based on coefficients of some sub-bands to extend a coefficient table, the coefficient table having coefficients for the respective
15 sub-bands on the high band side; pseudo high band sub-band powers, which are estimated values of powers of the high band sub-band signals, are calculated for the respective sub-bands on the high band side based on the extended or reduced coefficient table and the low band sub-band signals; high band
20 sub-band powers of the high band sub-band signals and the pseudo high band sub-band powers are compared to each other and one of a plurality of the coefficient tables is selected; coefficient information for obtaining the selected coefficient table is encoded to produce high band encoded data;
25 low band signals of the input signal are encoded to produce low band encoded data; and the low band encoded data and the high band encoded data are multiplexed to produce an output code string.

30 EFFECTS OF THE INVENTION

[0039]

According to the first embodiment to the fourth embodiment, it is possible to reproduce music signal with high sound quality by expansion of a frequency band.

5 BRIEF DESCRIPTION OF DRAWINGS

[0040]

Fig. 1 is a view an example of illustrating in an example of a power spectrum of a low band after decoding an input signal and a frequency envelope of a high band estimated.

10 Fig. 2 is a view illustrating an example of an original power spectrum of music signal of an attack according to rapid change in time.

Fig. 3 is a block diagram illustrating a functional configuration example of a frequency band expansion apparatus
15 in a first embodiment of the present invention.

Fig. 4 is a flowchart illustrating an example of a frequency band expansion process by a frequency band expansion apparatus in Fig. 3.

Fig. 5 is a view illustrating arrangement of a power
20 spectrum of signal input to a frequency band expansion apparatus in Fig. 3 and arrangement on a frequency axis of a band pass filter.

Fig. 6 is a view illustrating an example illustrating frequency characteristics of a vocal region and a power
25 spectrum of a high band estimated.

Fig. 7 is a view illustrating an example of a power spectrum of signal input to a frequency band expansion apparatus in Fig. 3.

Fig. 8 is a view illustrating an example of a power vector
30 after liftering of an input signal in Fig. 7.

Fig. 9 is a block diagram illustrating a functional

configuration example of a coefficient learning apparatus for performing learning of a coefficient used in a high band signal production circuit of a frequency band expansion apparatus in Fig. 3.

5 Fig. 10 is a flowchart describing an example of a coefficient learning process by a coefficient learning apparatus in Fig. 9.

10 Fig. 11 is a block diagram illustrating a functional configuration example of an encoder in a second embodiment of the present invention.

 Fig. 12 is a flowchart describing an example of an encoding process by an encoder in Fig. 11.

15 Fig. 13 is a block diagram illustrating a functional configuration example of a decoder in a second embodiment of the present invention.

 Fig. 14 is a flowchart describing an example of a decoding processing by a decoder in Fig. 13.

20 Fig. 15 is a block diagram illustrating a functional configuration example of a coefficient learning apparatus for performing learning of a representative vector used in a high band encoding circuit of an encoder in Fig. 11 and decoded high band sub-band power estimation coefficient used in a high band decoding circuit of decoder in Fig. 13.

25 Fig. 16 is a flowchart describing an example of a coefficient learning process by a coefficient learning apparatus in Fig. 15.

 Fig. 17 is a view illustrating an example of an encoded string to which an encoder in Fig. 11 is output.

30 Fig. 18 is a block diagram illustrating a functional configuration example of the encoder.

 Fig. 19 is a flowchart describing of encoding processing.

Fig. 20 is a block diagram illustrating a functional configuration example of a decoder.

Fig. 21 is a flowchart describing a decoding process.

Fig. 22 is a flowchart describing an encoding process.

5 Fig. 23 is a flowchart describing a decoding process.

Fig. 24 is a flowchart describing an encoding process.

Fig. 25 is a flowchart describing an encoding process.

Fig. 26 is a flowchart describing an encoding process.

Fig. 27 is a flowchart describing an encoding process.

10 Fig. 28 is a view illustrating a configuration example of a coefficient learning apparatus.

Fig. 29 is a flowchart describing a coefficient learning process.

Fig. 30 is a diagram illustrating a coefficient table.

15 Fig. 31 is a diagram illustrating the extension of a coefficient table.

Fig. 32 is a diagram illustrating the reduction of a coefficient table.

20 Fig. 33 is a block diagram illustrating a functional configuration example of an encoder.

Fig. 34 is a flowchart describing an encoding process.

Fig. 35 is a block diagram illustrating a functional configuration example of a decoder.

Fig. 36 is a flowchart describing a decoding process.

25 Fig. 37 is a diagram illustrating the sharing of a coefficient table using blended learning.

Fig. 38 is a view illustrating a configuration example of a coefficient learning apparatus.

30 Fig. 39 is a flowchart describing a coefficient learning process.

Fig. 40 is a block diagram illustrating a configuration

example of hardware of a computer executing a process to which the present invention is applied by a program.

MODE FOR CARRYING OUT THE INVENTION

5 [0041]

An embodiment of the present invention will be described with reference to the drawings. In addition, the description thereof is performed in the following sequence.

1. First embodiment (when the present invention is
10 applied to a frequency band expansion apparatus)

2. Second embodiment (when the present invention is applied to an encoder and a decoder)

3. Third embodiment (when a coefficient index is included in high band encoded data)

15 4. Fourth embodiment (when a difference between coefficient index and a pseudo high band sub-band power is included in high band encoded data)

5. Fifth embodiment (when a coefficient index is selected using an estimation value).

20 6. Sixth embodiment (when a portion of a coefficient is common)

7. Seventh Embodiment (In Case Where Coefficient Table is Extended or Reduced)

8. Eighth Embodiment (In Case Where Learning is Performed
25 Using Broadband Instruction Signals Having Different Conditions)

[0042]

<1. First Embodiment>

In a first embodiment, a process that expands a frequency
30 band (hereinafter, referred to as a frequency band expansion process) is performed with respect to a signal component of

a low band after decoding obtained by decoding encoded data using a high cancelation encoding method.

[0043]

[Functional Configuration Example of Frequency Band Expansion
5 Apparatus]

Fig. 3 illustrates a functional configuration example of a frequency band expansion apparatus according to the present invention.

[0044]

10 A frequency band expansion apparatus 10 performs a frequency band expansion process with respect to the input signal by setting a signal component of the low band after decoding as the input signal and outputs the signal after the frequency band expansion process obtained by the result as
15 an output signal.

[0045]

The frequency band expansion apparatus 10 includes a low-pass filter 11, a delay circuit 12, a band pass filter 13, a characteristic amount calculation circuit 14, a high
20 band sub-band power estimation circuit 15, a high band signal production circuit 16, a high-pass filter 17 and a signal adder 18.

[0046]

The low-pass filter 11 filters an input signal by a
25 predetermined cut off frequency and supplies a low band signal component, which is a signal component of the low band as a signal after filtering to the delay circuit 12.

[0047]

Since the delay circuit 12 is synchronized when adding
30 the low band signal component from the low-pass filter 11 and a high band signal component which will be described later

to each other, it delays the low signal component only a certain time and the low signal component is supplied to the signal adder 18.

[0048]

5 The band pass filter 13 includes band pass filters 13-1 to 13-N having pass bands different from each other. The band pass filter 13- i ($1 \leq i \leq N$) passes a signal of a predetermined pass band of the input signal and supplies the passed signal as one of a plurality of sub-band signal to the characteristic amount calculation circuit 14 and the high band signal production circuit 16.

[0049]

15 The characteristic amount calculation circuit 14 calculates one or more characteristic amounts by using at least any one of a plurality of sub-band signals and the input signal from the band pass filter 13 and supplies the calculated characteristic amounts to the high band sub-band power estimation circuit 15. Herein, the characteristic amounts are information showing a feature of the input signal as a signal.

[0050]

25 The high band sub-band power estimation circuit 15 calculates an estimation value of a high band sub-band power which is a power of the high band sub-band signal for each high band sub-band based on one or more characteristic amounts from the characteristic amount calculation circuit 14 and supplies the calculated estimation value to the high band signal production circuit 16.

[0051]

30 The high band signal production circuit 16 produces the high band signal component which is a signal component of the

high band based on a plurality of sub-band signals from the band pass filter 13 and an estimation value of a plurality of high band sub-band powers from the high band sub-band power estimation circuit 15 and supplies the produced high signal component to the high-pass filter 17.

[0052]

The high-pass filter 17 filters the high band signal component from the high band signal production circuit 16 using a cut off frequency corresponding to the cut off frequency in the low-pass filter 11 and supplies the filtered high band signal component to a signal adder 18.

[0053]

The signal adder 18 adds the low band signal component from the delay circuit 12 and the high band signal component from the high-pass filter 17 and outputs the added components as an output signal.

[0054]

In addition, in a configuration in Fig. 3, in order to obtain a sub-band signal, the band pass filter 13 is applied but is not limited thereto. For example, the band division filter disclosed in Patent Document 1 may be applied.

[0055]

In addition, likewise, in a configuration in Fig. 3, the signal adder 18 is applied in order to synthesize a sub-band signal, but is not limited thereto. For example, a band synthetic filter disclosed in Patent Document 1 may be applied.

[0056]

[Frequency Band Expansion Process of Frequency Band Expansion Apparatus]

Next, referring to a flowchart in Fig. 4, the frequency band expansion process by the frequency band expansion

apparatus in Fig. 3 will be described.

[0057]

In step S1, the low-pass filter 11 filters the input signal by a predetermined cutoff frequency and supplies the
5 low band signal component as a signal after filtering to the delay circuit 12.

[0058]

The low-pass filter 11 can set an optional frequency as the cutoff frequency. However, in an embodiment of the
10 present invention, the low-pass filter can set to correspond a frequency of a low end of the expansion start band by setting a predetermined frequency as an expansion start band described blow. Therefore, the low-pass filter 11 supplies a low band
15 band than the expansion start band to the delay circuit 12 as a signal after filtering.

[0059]

In addition, the low-pass filter 11 can set the optimal frequency as the cutoff frequency in response to the encoding
20 parameter such as the high band cancelation encoding method or a bit rate and the like of the input signal. As the encoding parameter, for example, side information employed in the band expansion method disclosed in Patent Document 1 can be used.

[0060]

25 In step S2, the delay circuit 12 delays the low band signal component only a certain delay time from the low-pass filter 11 and supplies the delayed low band signal component to the signal adder 18.

[0061]

30 In step S3, the band pass filter 13 (band pass filters 13-1 to 13-N) divides the input signal into a plurality of

sub-band signals and supplies each of a plurality of sub-band signals after the division to the characteristic amount calculation circuit 14 and the high band signal production circuit 16. In addition, the process of division of the input
5 signal by the band pass filter 13 will be described below.
[0062]

In step S4, the characteristic amount calculation circuit 14 calculates one or more characteristic amounts by at least one of a plurality of sub-band signals from the band
10 pass filter 13 and the input signal and supplies the calculated characteristic amounts to the high band sub-band power estimation circuit 15. In addition, a process of the calculation for the characteristic amount by the characteristic amount calculation circuit 14 will be described
15 below in detail.
[0063]

In step S5, the high band sub-band power estimation circuit 15 calculates an estimation value of a plurality of high band sub-band powers based on one or more characteristic
20 amounts and supplies the calculated estimation value to the high band signal production circuit 16 from the characteristic amount calculation circuit 14. In addition, a process of a calculation of an estimation value of the high band sub-band power by the high band sub-band power estimation circuit 15
25 will be described below in detail.
[0064]

In step S6, the high band signal production circuit 16 produces a high band signal component based on a plurality of sub-band signals from the band pass filter 13 and an
30 estimation value of a plurality of high band sub-band powers from the high band sub-band power estimation circuit 15 and

supplies the produced high band signal component to the high-pass filter 17. In this case, the high band signal component is the signal component of the higher band than the expansion start band. In addition, a process on the production of the high band signal component by the high band signal production circuit 16 will be described below in detail.

[0065]

In step S7, the high-pass filter 17 removes the noise such as an alias component in the low band included in the high band signal component by filtering the high band signal component from the high band signal production circuit 16 and supplies the high band signal component to the signal adder 18.

[0066]

In step S8, a signal adder 18 adds the low band signal component from the delay circuit 12 and the high band signal component from the high-pass filter 17 to each other and outputs the added components as an output signal.

[0067]

According to the above-mentioned process, the frequency band can be expanded with respect to a signal component of the low band after decoding.

[0068]

Next, a description for each process of step S3 to S6 of the flowchart in Fig. 4 will be described.

[0069]

[Description of Process by Band Pass Filter]

First, a description of process by the band pass filter 13 in step S3 in a flowchart of Fig. 4 will be described.

[0070]

In addition, for convenience of the explanation, as

described below, it is assumed that the number N of the band pass filter 13 is $N = 4$.

[0071]

For example, it is assumed that one of 16 sub-bands
5 obtained by dividing Nyquist frequency of the input signal into 16 parts is an expansion start band and each of 4 sub-bands of the lower band than the expansion start band of 16 sub-bands is each pass band of the band pass filters 13-1 to 13-4.

[0072]

10 Fig. 5 illustrates arrangements on each axis of a frequency for each pass band of the band pass filters 13-1 to 13-4.

[0073]

As illustrated in Fig. 5, if it is assumed that an index
15 of the first sub-band from the high band of the frequency band (sub-band) of the lower band than the expansion start band is sb , an index of second sub-band is $sb-1$, and an index of I -th sub-band is $sb-(I-1)$, Each of band pass filters 13-1 to 13-4 assign each sub-band in which the index is sb to $sb-3$
20 among the sub-band of the low band lower than the expansion initial band as the pass band.

[0074]

In the present embodiment, each pass band of the band
25 pass filters 13-1 to 13-4 is 4 predetermined sub-bands of 16 sub-bands obtained by dividing the Nyquist frequency of the input signal into 16 parts but is not limited thereto and may be 4 predetermined sub-bands of 256 sub-band obtained by dividing the Nyquist frequency of the input signal into 256
30 parts. In addition, each bandwidth of the band pass filters 13-1 to 13-4 may be different from each other.

[0075]

[Description of Process by Characteristic Amount Calculation Circuit]

Next, a description of a process by the characteristic amount calculation circuit 14 in step S4 of the flowchart in Fig. 4 will be described.

[0076]

The characteristic amount calculation circuit 14 calculates one or more characteristic amounts used such that the high band sub-band power estimation circuit 15 calculates the estimation value of the high band sub-band power by using at least one of a plurality of sub-band signals from the band pass filter 13 and the input signal.

[0077]

In more detail, the characteristic amount calculation circuit 14 calculates as the characteristic amount, the power of the sub-band signal (sub-band power (hereinafter, referred to as a low band sub-band power)) for each sub-band from 4 sub-band signals of the band pass filter 13 and supplies the calculated power of the sub-band signal to the high band sub-band power estimation circuit 15.

[0078]

In other words, the characteristic amount calculation circuit 14 calculates the low band sub-band power $power(ib, J)$ in a predetermined time frame J from 4 sub-band signals $x(ib, n)$, which is supplied from the band pass filter 13 by using the following Equation (1). Herein, ib is an index of the sub-band, and n is expressed as index of discrete time. In addition, the number of a sample of one frame is expressed as F_{SIZE} and power is expressed as decibel.

[0079]

[Equation 1]

$$\text{power}(ib, J) = 10 \log_{10} \left\{ \left(\sum_{n=J*FSIZE}^{(J+1)*FSIZE-1} x(ib, n)^2 \right) / FSIZE \right\}$$

(sb-3 ≤ ib ≤ sb)

. . . (1)

[0080]

Accordingly, the low band sub-band power $\text{power}(ib, J)$
 5 obtained by the characteristic amount calculation circuit 14
 is supplied to the high band sub-band power estimation circuit
 15 as the characteristic amount.

[0081]

[Description of Process by High Band Sub-Band Power Estimation
 10 Circuit]

Next, a description of a process by the high band sub-band
 power estimation circuit 15 of step S5 of a flowchart in Fig.
 4 will be described.

[0082]

15 The high band sub-band power estimation circuit 15
 calculates an estimation value of the sub-band power (high
 band sub-band power) of the band (frequency expansion band)
 which is caused to be expanded following the sub-band
 (expansion start band) of which the index is sb+1, based on
 20 4 sub-band powers supplied from the characteristic amount
 calculation circuit 14.

[0083]

That is, if the high band sub-band power estimation
 circuit 15 considers the index of the sub-band of maximum band
 25 of the frequency expansion band to be eb, (eb-sb) sub-band
 power is estimated with respect to the sub-band in which the
 index is sb+1 to eb.

[0084]

In the frequency expansion band, the estimation value $\text{power}_{\text{est}}(\text{ib}, J)$ of sub-band power of which the index is ib is expressed by the following Equation (2) using 4 sub-band power
 5 $\text{power}(\text{ib}, j)$ supplied from the characteristic amount calculation circuit 14.

[0085]

[Equation 2]

$$\text{power}_{\text{est}}(\text{ib}, J) = \left(\sum_{kb=\text{sb}-3}^{\text{sb}} [A_{\text{ib}}(kb) \text{power}(kb, J)] \right) + B_{\text{ib}}$$

($J * \text{FSIZE} \leq n \leq (J+1) \text{FSIZE} - 1, \text{sb} + 1 \leq \text{ib} \leq \text{eb}$)

... (2)

10 [0086]

Herein, in Equation (2), coefficients $A_{\text{ib}}(kb)$, and B_{ib} are coefficients having value different for respective sub-band ib . Coefficients $A_{\text{ib}}(kb)$, B_{ib} are coefficients set suitably to obtain a suitable value with respect to various
 15 input signals. In addition, Coefficients $A_{\text{ib}}(kb)$, B_{ib} are also charged to an optimal value by changing the sub-band sb . A deduction of $A_{\text{ib}}(kb)$, B_{ib} will be described below.

[0087]

In Equation (2), the estimation value of the high band
 20 sub-band power is calculated by a primary linear combination using power of each of a plurality of sub-band signals from the band pass filter 13, but is not limited thereto, and for example, may be calculated using a linear combination of a plurality of the low band sub-band powers of frames before
 25 and after the time frame J , and may be calculated using a nonlinear function.

[0088]

As described above, the estimation value of the high band sub-band power calculated by the high band sub-band power estimation circuit 15 is supplied to the high band signal production circuit 16 will be described.

5 [0089]

[Description of Process by High Band Signal Production Circuit]

Next, a description will be made of process by the high band signal production circuit 16 in step S6 of a flowchart in Fig 4.

10 [0090]

The high band signal production circuit 16 calculates the low band sub-band power $power(ib, J)$ of each sub-band based on Equation (1) described above, from a plurality of sub-band signals supplied from the band pass filter 13. The high band signal production circuit 16 obtains a gain amount $G(ib, J)$ by Equation 3 described below, using a plurality of low band sub-band powers $power(ib, J)$ calculated, and an estimation value $power_{est}(ib, J)$ of the high band sub-band power calculated based on Equation (2) described above by the high band sub-band power estimation circuit 15.

[0091]

[Equation 3]

$$G(ib, J) = 10^{\{(power_{est}(ib, J) - power(sb_{map}(ib), J)) / 20\}}$$

$$(J * FSIZE \leq n \leq (J + 1) FSIZE - 1, sb + 1 \leq ib \leq eb)$$

... (3)

[0092]

25 Herein, in Equation (3), $sb_{map}(ib)$ shows the index of the sub-band of an original map of the case where the sub-band ib is considered as the sub-band of an original map and is expressed by the following Equation 4.

[0093]

[Equation 4]

$$sb_{map}(ib) = ib - 4 \text{INT} \left(\frac{ib - sb - 1}{4} + 1 \right)$$

$$(sb+1 \leq ib \leq eb)$$

. . . (4)

[0094]

5 In addition, in Equation (4), INT (a) is a function which cut down a decimal point of value a.

[0095]

Next, the high band signal production circuit 16 calculates the sub-band signal $x2(ib, n)$ after gain control by multiplying
 10 the gain amount $G(ib, J)$ obtained by Equation 3 by an output of the band pass filter 13 using the following Equation (5).

[0096]

[Equation 5]

$$x2(ib, n) = G(ib, J) \times (sb_{map}(ib), n)$$

$$(J \cdot FSIZE \leq n \leq (J+1) FSIZE - 1, sb+1 \leq ib \leq eb)$$

. . . (5)

15 [0097]

Further, the high band signal production circuit 16 calculates the sub-band signal $x3(ib, n)$ after the gain control which is cosine-transferred from the sub-band signal $x2(ib, n)$ after adjustment of gain by performing cosine transfer to
 20 a frequency corresponding a frequency of the upper end of the sub-band having index of sb from a frequency corresponding to a frequency of the lower end of the sub-band having the index of sb-3 by the following Equation (6).

[0098]

[Equation 6]

$$x3(ib, n) = x2(ib, n) * 2\cos(n) * [4(ib+1) \pi / 32]$$

$$(sb+1 \leq ib \leq eb)$$

... (6)

[0099]

In addition, in Equation (6), π shows a circular constant.

5 Equation (6) means that the sub-band signal $x2(ib, n)$ after the gain control is shifted to the frequency of each of 4 band part high band sides.

[0100]

10 Therefore, the high band signal production circuit 16 calculates the high band signal component $x_{high}(n)$ from the sub-band signal $x3(ib, n)$ after the gain control shifted to the high band side according to the following Equation 7.

[0101]

[Equation 7]

$$x_{high}(n) = \sum_{ib=sb+1}^{eb} x3(ib, n)$$

... (7)

15

[0102]

Accordingly, the high band signal component is produced by the high band signal production circuit 16 based on the 4 low band sub-band powers obtained based on the 4 sub-band
20 signals from the band pass filter 13 and an estimation value of the high band sub-band power from the high band sub-band power estimation circuit 15, and the produced high band signal component is supplied to the high-pass filter 17.

[0103]

25 According to process described above, since the low band

sub-band power calculated from a plurality of sub-band signals is set as the characteristic amount with respect to the input signal obtained after decoding of the encoded data by the high band cancelation encoding method, the estimation value of the high band sub-band power is calculated based on a coefficient set suitably thereto, and the high band signal component is produced adaptively from the estimation value of the low band sub-band power and the high band sub-band power, whereby it is possible to estimate the sub-band power of the frequency expansion band with high accuracy and to reproduce a music signal with a better sound quality.

[0104]

As described above, the characteristic amount calculation circuit 14 illustrates an example that calculates as the characteristic amount, only the low band sub-band power calculated from the plurality sub-band signal. However, in this case, the sub-band power of the frequency expansion band cannot be estimated with high accuracy by a kind of the input signal.

20 [0105]

Herein, the estimate of the sub-band power of the frequency expansion band in the high band sub-band power estimation circuit 15 can be performed with high accuracy because the characteristic amount calculation circuit 14 calculates a characteristic amount having a strong correlation with an output system of sub-band power of the frequency expansion band (a power spectrum shape of the high band).

[0106]

[Another Example of Characteristic Amount Calculated by Characteristic Amount Calculation Circuit]

30

Fig. 6 illustrates an example of the frequency

characteristic of a vocal region where most of vocal is occupied and the power spectrum of the high band obtained by estimating the high band sub-band power by calculating only the low band sub-band power as the characteristic amount.

5 [0107]

As illustrated in Fig. 6, in the frequency characteristic of the vocal region, there are many cases where the estimated power spectrum of the high band has a position higher than the power spectrum of the high band of an original signal. Since sense of incongruity of the singing voice of people is easily perceived by the people's ear, it is necessary to estimate the high band sub-band power with high accuracy in vocal region.

[0108]

15 In addition, as illustrated in Fig. 6, in the frequency characteristic of the vocal region, there are many cases that a larger concave is disposed from 4.9 kHz to 11.025 kHz.

[0109]

Herein, as described below, an example will be described which can apply a degree of the concave in 4.9 kHz to 11.025 kHz in the frequency area as a characteristic amount used in estimating the high band sub-band power of the vocal region. In addition, a characteristic amount showing a degree of the concave is referred to as a dip below.

25 [0110]

A calculation example of a dip in time frames J $\text{dip}(J)$ will be described below.

[0111]

Fast Fourier Transform (FFT) of 2048 points is performed with respect to signals of 2048 sample sections included in a range of a few frames before and after a time frame J of

the input signal, and coefficients on the frequency axis is calculated. The power spectrum is obtained by performing db conversion with respect to the absolute value of each of the calculated coefficients.

5 [0112]

Fig. 7 illustrates one example of the power spectrum obtained in above-mentioned method. Herein, in order to remove a fine component of the power spectrum, for example so as to remove component of 1.3 kHz or less, a liftering process is performed. If the liftering process is performed, it is possible to smooth the fine component of the spectrum peak by selecting each dimension of the power spectrum and performing a filtering process by applying the low-pass filter according to a time sequence.

15 [0113]

Fig. 8 illustrates an example of the power spectrum of the input signal after liftering. In the power spectrum following recovering illustrated in Fig. 8, difference between minimum value and maximum value included in a range corresponding to 4.9 kHz to 11.025 kHz is set as a dip dip(J).

[0114]

As described above, the characteristic amount having a strong correlation with the sub-band power of the frequency expansion band is calculated. In addition, a calculation example of a dip dip(J) is not limited to the above-mentioned method, and other method may be performed.

[0115]

Next, other example of calculation of a characteristic amount having a strong correlation with the sub-band power of the frequency expansion band will be described.

[0116]

[Still Another Example of Characteristic Amount Calculated
by Characteristic Amount Calculation Circuit]

In a frequency characteristic of an attack region, which
is, a region including an attack type music signal in any input
5 signal, there are many cases that the power spectrum of the
high band is substantially flat as described with reference
to Fig. 2. It is difficult for a method calculating as the
characteristic amount, only the low band sub-band power to
estimate the sub-band power of the almost flat frequency
10 expansion band seen from an attack region with high accuracy
in order to estimate the sub-band power of a frequency expansion
band without the characteristic amount indicating time
variation having a specific input signal including an attack
region.

15 [0117]

Herein, an example applying time variation of the low
band sub-band power will be described below as the
characteristic amount used for estimating the high band
sub-band power of the attack region.

20 [0118]

Time vibration power_d (J) of the low band sub-band power
in some time frames J, for example, is obtained from the
following Equation (8).

[0119]

25 [Equation 8]

$$\text{power}_d(J) = \frac{\sum_{ib=sb-3}^{sb} \sum_{n=J*FSIZE}^{(J+1)*FSIZE-1} (x(ib, n)^2)}{\sum_{ib=sb-3}^{sb} \sum_{n=(J-1)*FSIZE}^{J*FSIZE-1} (x(ib, n)^2)} \quad \dots (8)$$

[0120]

According to Equation 8, time variation power_d(J) of a low band sub-band power shows ratio between the sum of four low band sub-band powers in time frames J-1 and the sum of four low band sub-band powers in time frames (J-1) before one frame of the time frames J, and if this value become large, the time variation of power between frames is large, that is, a signal included in time frames J is regarded as having strong attack.

10 [0121]

In addition, if the power spectrum illustrated in Fig. 1, which is average statistically is compared with the power spectrum of the attack region (attack type music signal) illustrated in Fig. 2, the power spectrum in the attack region ascends toward the right in a middle band. Between the attack regions, there are many cases which show the frequency characteristics.

[0122]

Accordingly, an example which applies a slope in the middle band as the characteristic amount used for estimating the high band sub-band power between the attack regions will be described below.

[0123]

A slope slope (J) of a middle band in some time frames J, for example, is obtained from the following Equation (9).

[0124]

[Equation 9]

$$\text{slope}(J) = \frac{\sum_{ib=sb-3}^{sb} \sum_{n=J*FSIZE}^{(J+1)*FSIZE-1} \{W(ib)*x(ib,n)^2\}}{\sum_{ib=sb-3}^{sb} \sum_{n=J*FSIZE}^{(J+1)*FSIZE-1} (x(ib,n)^2)} \quad \dots (9)$$

[0125]

In the Equation (9), a coefficient $w(ib)$ is a weight factor adjusted to be weighted to the high band sub-band power.

5 According to the Equation (9), the slope (J) shows a ratio of the sum of four low band sub-band powers weighted to the high band and the sum of four low band sub-band powers. For example, if four low band sub-band powers are set as a power with respect to the sub-band of the middle band, the slope

10 (J) has a large value when the power spectrum in a middle band ascends to the right, and the power spectrum has small value when the power spectrum descends to the right.

[0126]

Since there are many cases that the slope of the middle

15 band considerably varies before and after the attack section, it may be assumed that the time variety $\text{slope}_d(J)$ of the slope expressed by the following Equation (10) is the characteristic amount used in estimating the high band sub-band power of the attack region.

20 [0127]

[Equation 10]

$$\text{slope}_d(J) = \text{slope}(J) / \text{slope}(J-1)$$

$$(J*FSIZE \leq n \leq (J+1)*FSIZE-1)$$

... (10)

[0128]

In addition, it may be assumed that time variety $\text{dip}_d(J)$ of the dip $\text{dip}(J)$ described above, which is expressed by the following Equation (11) is the characteristic amount used in estimating the high band sub-band power of the attack region.

5 [0129]

[Equation 11]

$$\begin{aligned} \text{dip}_d(J) &= \text{dip}(J) - \text{dip}(J-1) \\ (J * \text{FSIZE} \leq n \leq (J+1) \text{FSIZE} - 1) \end{aligned}$$

. . . (11)

[0130]

According to the above-mentioned method, since the characteristic amount having a strong correlation with the sub-band power of the frequency expansion band is calculated, if using this, the estimation for the sub-band power of the frequency expansion band in the high band sub-band power estimation circuit 15 can be performed with high accuracy.

15 [0131]

As described above, an example for calculating the characteristic amount having a strong correlation with the sub-band power of the frequency expansion band was described. However, an example for estimating the high band sub-band power will be described below using the characteristic amount calculated by the method described above.

[0132]

[Description of Process by High Band Sub-band Power Estimation Circuit]

25 Herein, an example for estimating the high band sub-band power using the dip described with reference to Fig. 8 and the low band sub-band power as the characteristic amount will be described.

[0133]

That is, in step S4 of the flowchart in Fig. 4, the characteristic amount calculation circuit 14 calculates as the characteristic amount, the low band sub-band power and the dip and supplies the calculated low band sub-band power and dip to the high band sub-band power estimation circuit 15 for each sub-band from four sub-band signals from the band pass filter 13.

[0134]

10 Therefore, in step S5, the high band sub-band power estimation circuit 15 calculates the estimation value of the high band sub-band power based on the four low band sub-band powers and the dip from the characteristic amount calculation circuit 14.

15 [0135]

Herein, in the sub-band power and the dip, since ranges of the obtained values (scales) are different from each other, the high band sub-band power estimation circuit 15, for example, performs the following conversion with respect to the dip value.

[0136]

The high band sub-band power estimation circuit 15 calculates the sub-band power of a maximum band of the four low band sub-band powers and a dip value with respect to a predetermined large amount of the input signal and obtains an average value and standard deviation respectively. Herein, it is assumed that the average value of sub-band power is $power_{ave}$, a standard deviation of the sub-band power is $power_{std}$, the average value of the dip is dip_{ave} , and the standard deviation of the dip is dip_{std} .

[0137]

The high band sub-band power estimation circuit 15 converts the value of the dip $\text{dip}(J)$ using the value as in the following Equation (12) and obtains the $\text{dip}_s \text{dip}(J)$ after conversion.

5 [0138]

[Equation 12]

$$\text{dip}_s(J) = \frac{\text{dip}(J) - \text{dip}_{\text{ave}}}{\text{dip}_{\text{std}}} \text{power}_{\text{std}} + \text{power}_{\text{ave}} \quad \dots (12)$$

[0139]

By performing conversion described in Equation (12),
 10 the high band sub-band power estimation circuit 15 can statistically convert the value of dip $\text{dip}(J)$ to an equal variable (dip) $\text{dip}_s(J)$ for the average and dispersion of the low band sub-band power and make a range of the value obtained from the dip approximately equal to a range of the value obtained
 15 from the sub-band power.

[0140]

In the frequency expansion band, the estimation value $\text{power}_{\text{est}}(\text{ib}, J)$ of the sub-band power in which index is ib , is expressed, according to Equation 13, by a linear combination
 20 of the four low band sub-band powers $\text{power}(\text{ib}, J)$ from the characteristic amount calculation circuit 14 and the $\text{dip}_s(J)$ shown in Equation (12).

[0141]

[Equation 13]

$$\text{power}_{\text{est}}(\text{ib}, J) = \left(\sum_{\text{kb}=\text{sb}-3}^{\text{sb}} \{C_{\text{ib}}(\text{kb}) \text{power}(\text{kb}, J)\} \right) + D_{\text{ib}} \text{dip}_s(J) + E_{\text{ib}}$$

(J*FSIZE ≤ n ≤ (J+1) FSIZE-1, sb+1 ≤ ib ≤ eb)

. . . (13)

[0142]

Herein, in Equation (13), coefficients $C_{\text{ib}}(\text{kb})$, D_{ib} , E_{ib} are coefficients having value different for each sub-band ib .
 5 The coefficients $C_{\text{ib}}(\text{kb})$, D_{ib} , and E_{ib} are coefficients set suitably in order to obtain a favorable value with respect to various input signals. In addition, the coefficient $C_{\text{ib}}(\text{kb})$, D_{ib} and E_{ib} are also changed to optimal values in order to change sub-band sb . Further, derivation of coefficient $C_{\text{ib}}(\text{kb})$, D_{ib} ,
 10 and E_{ib} will be described below.

[0143]

In Equation (13), the estimation value of the high band sub-band power is calculated by a linear combination, but is not limited thereto. For example, the estimation value may
 15 be calculated using a linear combination of a plurality characteristic amount of a few frames before and after the time frame J , and may be calculated using a non-linear function.

[0144]

According to the process described above, it may be
 20 possible to reproduce music signal having a better quality in that estimation accuracy of the high band sub-band power at the vocal region is improved compared with a case that it is assumed that only the low band sub-band power is the characteristic amount in estimation of the high band sub-band
 25 power using a value of a specific dip of vocal region as a characteristic amount, the power spectrum of the high band is produced by being estimated to be larger than that of the

high band power spectrum of the original signal and sense of incongruity can be easily perceived by the people's ear using a method setting only the low band sub-band as the characteristic amount.

5 [0145]

Therefore, if the number of divisions of sub-bands is 16, since frequency resolution is low with respect to the dip calculated as the characteristic amount by the method described above (a degree of the concave in a frequency characteristic of the vocal region), a degree of the concave cannot be expressed by only the low band sub-band power.

[0146]

Herein, the frequency resolution is improved and it may be possible to express the degree of the concave at only the low band sub-band power in that the number of the divisions of the sub-bands increases (for example, 256 divisions of 16 times), the number of the band divisions by the bandpass filter increases (for example, 64 of 16 times), and the number of the low band sub-band power calculated by the characteristic amount calculation circuit 14 increases (64 of 16 times).

[0147]

By only a low band sub-band power, it is assumed that it is possible to estimate the high band sub-band power with accuracy substantially equal to the estimation of the high band sub-band power used as the characteristic amount and the dip described above.

[0148]

However, a calculation amount increases by increasing the number of the divisions of the sub-bands, the number of the band divisions and the number of the low band sub-band powers. If it is assumed that the high band sub-band power

can be estimated with accuracy equal to any method, the method that estimates the high band sub-band power using the dip as the characteristic amount without increasing the number of divisions of the sub-bands is considered to be efficient in terms of the calculation amount.

[0149]

As described above, a method that estimates the high band sub-band power using the dip and the low band sub-band power was described, but as the characteristic amount used in estimating the high band sub-band power, one or more the characteristic amounts described above (a low band sub-band power, a dip, time variation of the low band sub-band power, slope, time variation of the slope, and time variation of the dip) without being limited to the combination. In this case, it is possible to improve accuracy in estimating the high band sub-band power.

[0150]

In addition, as described above, in the input signal, it may be possible to improve estimation accuracy of the section by using a specific parameter in which estimation of the high band sub-band power is difficult as the characteristic amount used in estimating the high band sub-band power. For example, time variety of the low band sub-band power, slope, time variety of slope and time variety of the dip are a specific parameter in the attack region, and can improve estimation accuracy of the high band sub-band power in the attack region by using the parameter thereof as the characteristic amount.

[0151]

In addition, even if estimation of the high band sub-band power is performed using the characteristic amount other than the low band sub-band power and the dip, that is, time variety

of the low band sub-band power, slope, time variety of the slope and time variety of the dip, the high band sub-band power can be estimated in the same manner as the method described above.

5 [0152]

In addition, each calculation method of the characteristic amount described in the specification is not limited to the method described above, and other method may be used.

10 [0153]

[Method for Obtaining Coefficients $C_{ib}(kb)$, D_{ib} , E_{ib}]

Next, a method for obtaining the coefficients $C_{ib}(kb)$, D_{ib} and E_{ib} will be described in Equation (13) described above.

[0154]

15. The method is applied in which coefficients is determined based on learning result, which performs learning using instruction signal having a predetermined broad band (hereinafter, referred to as a broadband instruction signal) such that as method for obtaining coefficients $C_{ib}(kb)$, D_{ib} and E_{ib} , coefficients $C_{ib}(kb)$, D_{ib} and E_{ib} become suitable values with respect to various input signals in estimating the sub-band power of the frequency expansion band.

[0155]

25 When learning of coefficient $C_{ib}(kb)$, D_{ib} and E_{ib} is performed, a coefficient learning apparatus including the band pass filter having the same pass band width as the band pass filters 13-1 to 13-4 described with reference to Fig. 5 is applied to the high band higher the expansion initial band. The coefficient learning apparatus performs learning when
30 broadband instruction is input.

[0156]

[Functional Configuration Example of Coefficient Learning Apparatus]

Fig. 9 illustrates a functional configuration example of a coefficient learning apparatus performing an instruction of coefficients $C_{1b}(kb)$, D_{1b} and E_{1b} .

[0157]

The signal component of the low band lower than the expansion initial band of a broadband instruction signal input to a coefficient learning apparatus 20 in Fig. 9 is a signal encoded in the same manner as an encoding method performed when the input signal having a limited band input to the frequency band expansion apparatus 10 in Fig. 3 is encoded.

[0158]

A coefficient learning apparatus 20 includes a band pass filter 21, a high band sub-band power calculation circuit 22, a characteristic amount calculation circuit 23, and a coefficient estimation circuit 24.

[0159]

The band pass filter 21 includes band pass filters 21-1 to 21-(K+N) having the pass bands different from each other. The band pass filter 21-i ($1 \leq i \leq K+N$) passes a signal of a predetermined pass band of the input signal and supplies the passed signal to the high band sub-band power calculation circuit 22 or the characteristic amount calculation circuit 23 as one of a plurality of sub-band signals. In addition, the band pass filters 21-1 to 21-K of the band pass filters 21-1 to 21-(K+N) pass a signal of the high band higher than the expansion start band.

[0160]

The high band sub-band power calculation circuit 22 calculates a high band sub-band power of each sub-band for

each constant time frame with respect to a plurality of sub-band signals of the high band, from the band pass filter 21 and supplies the calculated high band sub-band power to the coefficient estimation circuit 24.

5 [0161]

The characteristic amount calculation circuit 23 calculates the same characteristic amount as the characteristic amount calculated by the characteristic amount calculation circuit 14 of the frequency band expansion apparatus 10 in Fig. 3 for the same respective time frames as a constant time frames in which the high band sub-band power is calculated by the high band sub-band power calculation circuit 22. That is, the characteristic amount calculation circuit 23 calculates one or more characteristic amounts using at least one of a plurality of sub-band signals from the band pass filter 21, and the broadband instruction signal, and supplies the calculated characteristic amounts to the coefficient estimation circuit 24.

[0162]

20 The coefficient estimation circuit 24 estimates coefficient (coefficient data) used at the high band sub-band power estimation circuit 15 of the frequency band expansion apparatus 10 in Fig. 3 based on the high band sub-band power from the high band sub-band power calculation circuit 22 and the characteristic amount from the characteristic amount calculation circuit 23 for each constant time frame.

[0163]

[Coefficient Learning Process of Coefficient Learning Apparatus]

30 Next, referring to a flowchart in Fig. 10, coefficient learning process by a coefficient learning apparatus in Fig.

9 will be described.

[0164]

In step S11, the band pass filter 21 divides the input signal (expansionbandinstructionsignal) into (K+N) sub-band
5 signals. The bandpass filters 21-1 to 21-K supply a plurality of sub-band signals of the high band higher than the expansion initial band to the high band sub-band power calculation circuit 22. In addition, the band pass filters 21-(K+1) to 21-(K+N) supply a plurality of sub-band signals of the low
10 band lower than the expansion initial band to the characteristic amount calculation circuit 23.

[0165]

In step S12, the high band sub-band power calculation circuit 22 calculates the high band sub-band power power(ib,
15 J) of each sub-band for each constant time frame with respect to a plurality of the sub-band signals of the high band from the band pass filters 21 (band pass filter 21-1 to 21-K). The high band sub-band power power(ib, J) is obtained by the above mentioned Equation (1). The high band sub-band power
20 calculation circuit 22 supplies the calculated high band sub-band power to the coefficient estimation circuit 24.

[0166]

In step S13, the characteristic amount calculation circuit 23 calculates the characteristic amount for the same
25 each time frame as the constant time frame in which the high band sub-band power is calculated by the high band sub-band power calculation circuit 22.

[0167]

In addition, as described below, in the characteristic
30 amount calculation circuit 14 of the frequency band expansion apparatus 10 in Fig. 3, it is assumed that the four sub-band

powers and the dip of the low band are calculated as the characteristic amount and it will be described that the four sub-band powers and the dip of the low band calculated in the characteristic amount calculation circuit 23 of the

5 coefficient learning apparatus 20 similarly.

[0168]

That is, the characteristic amount calculation circuit 23 calculates four low band sub-band powers using four sub-band signals of the same respective four sub-band signals input

10 to the characteristic amount calculation circuit 14 of the frequency band expansion apparatus 10 from the bandpass filter 21 (band pass filter 21-(K+1) to 21-(K+4)). In addition, the characteristic amount calculation circuit 23 calculates the dip from the expansion band instruction signal and calculates

15 the dip $\text{dip}_s(J)$ based on the Equation (12) described above. Further, the characteristic amount calculation circuit 23 supplies the four low band sub-band powers and the dip $\text{dip}_s(J)$ as the characteristic amount to the coefficient estimation circuit 24.

20 [0169]

In step S14, the coefficient estimation circuit 24 performs estimation of coefficients $C_{1b}(kb)$, D_{1b} and E_{1b} based on a plurality of combinations of the (eb-sb) high band sub-band power of supplied to the same time frames from the high band

25 sub-band power calculation circuit 22 and the characteristic amount calculation circuit 23 and the characteristic amount (four low band sub-band powers and dip $\text{dip}_s(J)$). For example, the coefficient estimation circuit 24 determines the coefficients $C_{1b}(kb)$, D_{1b} and E_{1b} in Equation (13) by making

30 five characteristic amounts (four low band sub-band powers and dip $\text{dip}_s(J)$) be an explanatory variable with respect to

one of the sub-band of the high bands, and making the high band sub-band power $\text{power}(ib, J)$ be an explained variable and performing a regression analysis using a least-squares method.
[0170]

5 In addition, naturally the estimation method of coefficients $C_{ib}(kb)$, D_{ib} and E_{ib} is not limited to the above-mentioned method and various common parameter identification methods may be applied.
[0171]

10 According to the processes described above, since the learning of the coefficients used in estimating the high band sub-band power is set to be performed by using a predetermined expansion band instruction signal, there is possibility to obtain a preferred output result with respect to various input
15 signals input to the frequency band expansion apparatus 10 and thus it may be possible to reproduce a music signal having a better quality.
[0172]

 In addition, it is possible to calculate the coefficients
20 $A_{ib}(kb)$ and B_{ib} in the above-mentioned Equation (2) by the coefficient learning method.
[0173]

 As described above, the coefficient learning processes was described premising that each estimation value of the high
25 band sub-band power is calculated by the linear combination such as the four low band sub-band powers and the dip in the high band sub-band power estimation circuit 15 of the frequency band expansion apparatus 10.

 However, a method for estimating the high band sub-band
30 power in the high band sub-band power estimation circuit 15 is not limited to the example described above. For example,

since the characteristic amount calculation circuit 14 calculates one or more of the characteristic amounts other than the dip (time variation of a low band sub-band power, slope, time variation of the slope and time variation of the dip), the high band sub-band power may be calculated, the linear combination of a plurality of characteristic amount of a plurality of frames before and after time frames J may be used, or a non-linear function may be used. That is, in the coefficient learning process, the coefficient estimation circuit 24 may calculate (learn) the coefficient on the same condition as that regarding the characteristic amount, the time frames and the function used in a case where the high band sub-band power is calculated by the high band sub-band power estimation circuit 15 of the frequency band expansion apparatus 10.

[0174]

<2. Second Embodiment>

In a second embodiment, encoding processing and decoding processing in the high band characteristic encoding method by the encoder and the decoder are performed.

[0175]

[Functional Configuration Example of Encoder]

Fig. 11 illustrates a functional configuration example of the encoder to which the present invention is applied.

25 [0176]

An encoder 30 includes a 31, a low band encoding circuit 32, a sub-band division circuit 33, a characteristic amount calculation circuit 34, a pseudo high band sub-band power calculation circuit 35, a pseudo high band sub-band power difference calculation circuit 36, a high band encoding circuit 37, a multiplexing circuit 38 and a low band decoding circuit

39.

[0177]

The low-pass filter 31 filters an input signal using a predetermined cutoff frequency and supplies a signal of a low band lower than a cutoff frequency (hereinafter, referred to as a low band signal) as signal after filtering to the low band encoding circuit 32, a sub-band division circuit 33, and a characteristic amount calculation circuit 34.

[0178]

10 The low band encoding circuit 32 encodes a low band signal from the low-pass filter 31 and supplies low band encoded data obtained from the result to the multiplexing circuit 38 and the low band decoding circuit 39.

[0179]

15 The sub-band division circuit 33 equally divides the input signal and the low band signal from the low-pass filter 31 into a plurality of sub-band signals having a predetermined band width and supplies the divided signals to the characteristic amount calculation circuit 34 or the pseudo high band sub-band power difference calculation circuit 36. In particular, the sub-band division circuit 33 supplies a plurality of sub-band signals (hereinafter, referred to as a low band sub-band signal) obtained by inputting to the low band signal, to the characteristic amount calculation circuit 20 34. In addition, the sub-band division circuit 33 supplies the sub-band signal (hereinafter, referred to as a high band sub-band signal) of the high band higher than a cutoff frequency set by the low-pass filter 31 among a plurality of the sub-band signals obtained by inputting an input signal to the pseudo high band sub-band power difference calculation circuit 36.

30

[0180]

The characteristic amount calculation circuit 34 calculates one or more characteristic amounts using any one of a plurality of sub-band signals of the low band sub-band signal from the sub-band division circuit 33 and the low band signal from the low-pass filter 31 and supplies the calculated characteristic amounts to the pseudo high band sub-band power calculation circuit 35.

[0181]

The pseudo high band sub-band power calculation circuit 35 produces a pseudo high band sub-band power based on one or more characteristic amounts from the characteristic amount calculation circuit 34 and supplies the produced pseudo high band sub-band power to the pseudo high band sub-band power difference calculation circuit 36.

15 [0182]

The pseudo high band sub-band power difference calculation circuit 36 calculates a pseudo high band sub-band power difference described below based on the high band sub-band signal from the sub-band division circuit 33 and the pseudo high band sub-band power from the pseudo high band sub-band power calculation circuit 35 and supplies the calculated pseudo high band sub-band power difference to the high band encoding circuit 37.

[0183]

The high band encoding circuit 37 encodes the pseudo high band sub-band power difference from the pseudo high band sub-band power difference calculation circuit 36 and supplies the high band encoded data obtained from the result to the multiplexing circuit 38.

30 [0184]

The multiplexing circuit 38 multiplies the low band

encoded data from the low band encoding circuit 32 and the high band encoded data from the high band encoding circuit 37 and outputs as an output code string.

[0185]

5 The low band decoding circuit 39 suitably decodes the low band encoded data from the low band encoding circuit 32 and supplies decoded data obtained from the result to the sub-band division circuit 33 and the characteristic amount calculation circuit 34.

10 [0186]

[Encoding Processing of Encoder]

Next, referring to a flowchart in Fig. 12, the encoding processing by the encoder 30 in Fig. 11 will be described.

[0187]

15 In step S111, the low-pass filter 31 filters the input signal using a predetermined cutoff frequency and supplies the low band signal as the signal after filtering to the low band encoding circuit 32, the sub-band division circuit 33 and the characteristic amount calculation circuit 34.

20 [0188]

In step S112, the low band encoding circuit 32 encodes the low band signal from the low-pass filter 31 and supplies low band encoded data obtained from the result to the multiplexing circuit 38.

25 [0189]

In addition, for encoding of the low band signal in step S112, a suitable encoding method should be selected according to an encoding efficiency and a obtained circuit scale, and the present invention does not depend on the encoding method.

30 [0190]

In step S113, the sub-band division circuit 33 equally

divides the input signal and the low band signal to a plurality of sub-band signals having a predetermined bandwidth. The sub-band division circuit 33 supplies the low band sub-band signal obtained by inputting the low band signal to the
5 characteristic amount calculation circuit 34. In addition, the sub-band division circuit 33 supplies the high band sub-band signal of a band higher than a frequency of the band limit, which is set by the low-pass filter 31 of a plurality of sub-band signals obtained by inputting the input signal
10 to the pseudo high band sub-band power difference calculation circuit 36.

[0191]

In a step S114, the characteristic amount calculation circuit 34 calculates one or more characteristic amounts using
15 at least any one of a plurality of sub-band signals of the low band sub-band signal from sub-band division circuit 33 and a low band signal from the low-pass filter 31 and supplies the calculated characteristic amounts to the pseudo high band sub-band power calculation circuit 35. In addition, the
20 characteristic amount calculation circuit 34 in Fig. 11 has basically the same configuration and function as those of the characteristic amount calculation circuit 14 in Fig. 3. Since a process in step S114 is substantially identical with that of step S4 of a flowchart in Fig. 4, the description thereof
25 is omitted.

[0192]

In step S115, the pseudo high band sub-band power calculation circuit 35 produces a pseudo high band sub-band power based on one or more characteristic amounts from the
30 characteristic amount calculation circuit 34 and supplies the produced pseudo high band sub-band power to the pseudo high

band sub-band power difference calculation circuit 36. In addition, the pseudo high band sub-band power calculation circuit 35 in Fig. 11 has basically the same configuration and function as those of the high band sub-band power estimation circuit 15 in Fig. 3. Therefore, since a process in step S115 is substantially identical with that of step S5 of a flowchart in Fig. 4, the description thereof is omitted.

[0193]

In step S116, a pseudo high band sub-band power difference calculation circuit 36 calculates the pseudo high band sub-band power difference based on the high band sub-band signal from the sub-band division circuit 33 and the pseudo high band sub-band power from the pseudo high band sub-band power calculation circuit 35 and supplies the calculated pseudo high band sub-band power difference to the high band encoding circuit 37.

[0194]

Specifically, the pseudo high band sub-band power difference calculation circuit 36 calculates the (high band) sub-band power $\text{power}(\text{ib}, J)$ in a constant time frames J with respect to the high band sub-band signal from the sub-band division circuit 33. In addition, in an embodiment of the present invention, all the sub-band of the low band sub-band signal and the sub-band of the high band sub-band signal distinguishes using the index ib . The calculation method of the sub-band power can apply to the same method as first embodiment, that is, the method used by Equation (1) thereto.

[0195]

Next, the pseudo high band sub-band power difference calculation circuit 36 calculates a difference value (pseudo high band sub-band power difference) $\text{power}_{\text{diff}}(\text{ib}, J)$ between

the high band sub-band power $\text{power}(\text{ib}, J)$ and the pseudo high band sub-band power $\text{power}_{\text{lh}}(\text{ib}, J)$ from the pseudo high band sub-band power calculation circuit 35 in a time frame J . The pseudo high band sub-band power difference $\text{power}_{\text{diff}}(\text{ib}, J)$ is obtained by the following Equation (14).

[0196]

[Equation 14]

$$\text{power}_{\text{diff}}(\text{ib}, J) = \text{power}(\text{ib}, J) - \text{power}_{\text{lh}}(\text{ib}, J)$$

$$(J * \text{FSIZE} \leq n \leq (J+1) \text{FSIZE} - 1, \text{sb} + 1 \leq \text{ib} \leq \text{eb})$$

. . . (14)

[0197]

10 In Equation (14), an index $\text{sb} + 1$ shows an index of the sub-band of the lowest band in the high band sub-band signal. In addition, an index eb shows an index of the sub-band of the highest band encoded in the high band sub-band signal.

[0198]

15 As described above, the pseudo high band sub-band power difference calculated by the pseudo high band sub-band power difference calculation circuit 36 is supplied to the high band encoding circuit 37.

[0199]

20 In step S117, the high band encoding circuit 37 encodes the pseudo high band sub-band power difference from the pseudo high band sub-band power difference calculation circuit 36 and supplies high band encoded data obtained from the result to the multiplexing circuit 38.

25 [0200]

Specifically, the high band encoding circuit 37 determines that on obtained by making the pseudo high band sub-band power difference from the pseudo high band sub-band

power difference calculation circuit 36 be a
 vector (hereinafter, referred to as a pseudo high band sub-band
 power difference vector) belongs to which cluster among a
 plurality of clusters in a characteristic space of the
 5 predetermined pseudo high band power sub-band difference.
 Herein, the pseudo high band sub-band power difference vector
 in a time frame J has, as a element of the vector, a value
 of a pseudo high band sub-band power difference $\text{power}_{\text{diff}}(\text{ib}, J)$
 for each index ib , and shows the vector of an $(\text{eb}-\text{sb})$ dimension.
 10 In addition, the characteristic space of the pseudo high band
 sub-band power difference is set as a space of the $(\text{eb}-\text{sb})$
 dimension in the same way.

[0201]

Therefore, the high band encoding circuit 37 measures
 15 a distance between a plurality of each representative vector
 of a plurality of predetermined clusters and the pseudo high
 band sub-band power difference vector in a characteristic space
 of the pseudo high band sub-band power difference, obtains
 index of the cluster having the shortest distance (hereinafter,
 20 referred to as a pseudo high band sub-band power difference
 ID) and supplies the obtained index as the high band encoded
 data to the multiplexing circuit 38.

[0202]

In step S118, the multiplexing circuit 38 multiples low
 25 band encoded data output from the low band encoding circuit
 32 and high band encoded data output from the high band encoding
 circuit 37 and outputs an output code string.

[0203]

Therefore, as an encoder in the high band characteristic
 30 encoding method, Japanese Patent Application Laid-Open No.
 2007-17908 discloses a technology producing the pseudo high

band sub-band signal from the low band sub-band signal,
comparing the pseudo high band sub-band signal and power of
the high band sub-band signal with each other for each sub-band,
calculating a gain of power for each sub-band to match the
5 power of the pseudo high band sub-band signal to the power
of the high band sub-band signal, and causing the calculated
gain to be included in the code string as information of the
high band characteristic.

[0204]

10 According to the process described above, only the pseudo
high band sub-band power difference ID may be included in the
output code string as information for estimating the high band
sub-band power in decoding. That is, for example, if the number
of the predetermined clusters is 64, as information for
15 restoring the high band signal in a decoder, 6 bit information
may be added to the code string per a time frame and an amount
of information included in the code string can be reduced to
improve decoding efficiency compared with a method disclosed
in Japanese Patent Application Laid-Open No. 2007-17908, and
20 it is possible to reproduce a music signal having a better
sound quality.

[0205]

In addition, in the processes described above, the low
band decoding circuit 39 may input the low band signal obtained
25 by decoding the low band encoded data from the low band encoding
circuit 32 to the sub-band division circuit 33 and the
characteristic amount calculation circuit 34 if there is a
margin in the characteristic amount. In the decoding
processing by the decoder, the characteristic amount is
30 calculated from the low band signal decoding the low band
encoded data and the power of the high band sub-band is estimated

based on the characteristic amount. Therefore, even in the encoding processing, if the pseudo high band sub-band power difference ID which is calculated based on the characteristic amount calculated from the decoded low band signal is included
5 in the code string, in the decoding processing by the decoder, the high band sub-band power having a better accuracy can be estimated. Therefore, it is possible to reproduce a music signal having a better sound quality.

[0206]

10 [Functional Configuration Example of Decoder]

Next, referring to Fig. 13, a functional configuration example of a decoder corresponding to the encoder 30 in Fig. 11 will be described.

[0207]

15 A decoder 40 includes a demultiplexing circuit 41, a low band decoding circuit 42, a sub-band division circuit 43, a characteristic amount calculation circuit 44, and a high band decoding circuit 45, a decoded high band sub-band power calculation circuit 46, a decoded high band signal production
20 circuit 47, and a synthesis circuit 48.

[0208]

The demultiplexing circuit 41 demultiplexes the input code string into the high band encoded data and the low band encoded data and supplies the low band encoded data to the
25 low band decoding circuit 42 and supplies the high band encoded data to the high band decoding circuit 45.

[0209]

The low band decoding circuit 42 performs decoding of the low band encoded data from the demultiplexing circuit 41.
30 The low band decoding circuit 42 supplies a signal of a low band obtained from the result of the decoding (hereinafter,

referred to as a decoded low band signal) to the sub-band division circuit 43, the characteristic amount calculation circuit 44 and the synthesis circuit 48.

[0210]

5 The sub-band division circuit 43 equally divides a decoded low band signal from the low band decoding circuit 42 into a plurality of sub-band signals having a predetermined bandwidth and supplies the sub-band signal (decoded low band sub-band signal) to the characteristic amount calculation
10 circuit 44 and the decoded high band signal production circuit 47.

[0211]

 The characteristic amount calculation circuit 44 calculates one or more characteristic amounts using any one
15 of a plurality of sub-band signals of decoded low band sub-band signals from the sub-band division circuit 43, and a decoded low band signal from a low band decoding circuit 42, and supplies the calculated characteristic amounts to the decoded high band sub-band power calculation circuit 46.

20 [0212]

 The high band decoding circuit 45 decodes high band encoded data from the demultiplexing circuit 41 and supplies a coefficient (hereinafter, referred to as a decoded high band sub-band power estimation coefficient) for estimating a high
25 band sub-band power using a pseudo high band sub-band power difference ID obtained from the result, which is prepared for each predetermined ID (index), to the decoded high band sub-band power calculation circuit 46.

[0213]

30 The decoded high band sub-band power calculation circuit 46 calculates the decoded high band sub-band power based on

one or more characteristic amounts from the characteristic amount calculation circuit 44 and the decoded high band sub-band power estimation coefficient from the high band decoding circuit 45 and supplies the calculated decoded high band sub-band power to the decoded high band signal production circuit 47.

[0214]

The decoded high band signal production circuit 47 produces a decoded high band signal based on a decoded low band sub-band signal from the sub-band division circuit 43 and the decoded high band sub-band power from the decoded high band sub-band power calculation circuit 46 and supplies the produced signal and power to the synthesis circuit 48.

[0215]

The synthesis circuit 48 synthesizes a decoded low band signal from the low band decoding circuit 42 and the decoded high band signal from the decoded high band signal production circuit 47 and outputs the synthesized signals as an output signal.

[0216]

[Decoding Process of Decoder]

Next, a decoding process using the decoder in Fig. 13 will be described with reference to a flowchart in Fig. 14.

[0217]

In step S131, the demultiplexing circuit 41 demultiplexes an input code string into the high band encoded data and the low band encoded data, supplies the low band encoded data to the low band decoding circuit 42 and supplies the high band encoded data to the high band decoding circuit 45.

[0218]

In step S132, the low band decoding circuit 42 decodes

the low band encoded data from the demultiplexing circuit 41 and supplies the decoded low band signal obtained from the result to the sub-band division circuit 43, the characteristic amount calculation circuit 44 and the synthesis circuit 48.

5 [0219]

In step S133, the sub-band division circuit 43 equally divides the decoded low band signal from the low band decoding circuit 42 into a plurality of sub-band signals having a predetermined bandwidth and supplies the obtained decoded low band sub-band signal to the characteristic amount calculation circuit 44 and the decoded high band signal production circuit 47.

[0220]

In step S134, the characteristic amount calculation circuit 44 calculates one or more characteristic amount from any one of a plurality of the sub-band signals of the decoded low band sub-band signals from the sub-band division circuit 43 and the decoded low band signal from the low band decoding circuit 42 and supplies the signals to the decoded high band sub-band power calculation circuit 46. In addition, the characteristic amount calculation circuit 44 in Fig. 13 basically has the same configuration and function as the characteristic amount calculation circuit 14 in Fig. 3 and the process in step S134 has the same process in step S4 of a flowchart in Fig. 4. Therefore, the description thereof is omitted.

[0221]

In step S135, the high band decoding circuit 45 decodes the high band encoded data from the demultiplexing circuit 41 and supplies the decoded high band sub-band power estimation coefficient prepared for each predetermined ID (index) using

the pseudo high band sub-band power difference ID obtained from the result to the decoded high band sub-band power calculation circuit 46.

[0222]

5 In step S136, the decoded high band sub-band power calculation circuit 46 calculates the decoded high band sub-band power based on one or more characteristic amount from the characteristic amount calculation circuit 44 and the decoded high band sub-band power estimation coefficient from
10 the high band decoding circuit 45 and supplies the power to the decoded high band signal production circuit 47. In addition, since the decoding high band, decoding high bands sub-band calculation circuit 46 in Fig. 13 has the same configuration and a function as those of the high band sub-band
15 power estimation circuit 15 in Fig. 3 and process in step S136 has the same process in step S5 of a flowchart in Fig. 4, the detailed description is omitted.

[0223]

20 In step S137, the decoded high band signal production circuit 47 outputs a decoded high band signal based on a decoded low band sub-band signal from the sub-band division circuit 43 and a decoded high band sub-band power from the decoded high band sub-band power calculation circuit 46. In addition, since the decoded high band signal production circuit 47 in
25 Fig. 13 basically has the same configuration and function as the high band signal production circuit 16 in Fig. 3 and the process in step S137 has the same process as step S6 of the flowchart in Fig. 4, the detailed description thereof is omitted.

30 [0224]

 In step S138, the synthesis circuit 48 synthesizes a

decoded low band signal from the low band decoding circuit 42 and a decoded high band signal from the decoded high band signal production circuit 47 and outputs synthesized signal as an output signal.

5 [0225]

According to the process described above, it is possible to improve estimation accuracy of the high band sub-band power and thus it is possible to reproduce music signals having a good quality in decoding by using the high band sub-band power estimation coefficient in decoding in response to the difference characteristic between the pseudo high band sub-band power calculated in advance in encoding and an actual high band sub-band power.

[0226]

15 In addition, according to the process, since information for producing the high band signal included in the code string has only a pseudo high band sub-band power difference ID, it is possible to effectively perform the decoding processing.

[0227]

20 As described above, although the encoding process and decoding processing according to the present invention are described, hereinafter, a method of calculates each representative vector of a plurality of clusters in a specific space of a predetermined pseudo high band sub-band power difference in the high band encoding circuit 37 of the encoder 30 in Fig. 11 and a decoded high band sub-band power estimation coefficient output by the high band decoding circuit 45 of the decoder 40 in Fig. 13 will be described.

[0228]

30 [Calculation Method of Calculating Representative Vector of A plurality of Clusters in Specific Space of Pseudo High Band

Sub-Band Power Difference And Decoding High Band Sub-Band Power Estimation Coefficient Corresponding to Each Cluster]

As a way for obtaining the representative vector of a plurality of clusters and the decoded high band sub-band power estimation coefficient of each cluster, it is necessary to prepare the coefficient so as to estimate the high band sub-band power in a high accuracy in decoding in response to a pseudo high band sub-band power difference vector calculated in encoding. Therefore, learning is performed by a broadband instruction signal in advance and the method of determining the learning is applied based on the learning result.

[0229]

[Functional Configuration Example of Coefficient Learning Apparatus]

Fig. 15 illustrates a functional configuration example of a coefficient learning apparatus performing learning of a representative vector of a plurality of cluster and a decoded high band sub-band power estimation coefficient of each cluster.

[0230]

It is preferable that a signal component of the broadband instruction signal input to the coefficient learning apparatus 50 in Fig. 15 and of a cutoff frequency or less set by a low-pass filter 31 of the encoder 30 is a decoded low band signal in which the input signal to the encoder 30 passes through the low-pass filter 31, that is encoded by the low band encoding circuit 32 and that is decoded by the low band decoding circuit 42 of the decoder 40.

[0231]

A coefficient learning apparatus 50 includes a low-pass filter 51, a sub-band division circuit 52, a characteristic

amount calculation circuit 53, a pseudo high band sub-band
power calculation circuit 54, a pseudo high band sub-band power
difference calculation circuit 55, a pseudo high band sub-band
power difference clustering circuit 56 and a coefficient
5 estimation circuit 57.

[0232]

In addition, since each of the low-pass filter 51, the
sub-band division circuit 52, the characteristic amount
calculation circuit 53 and the pseudo high band sub-band power
10 calculation circuit 54 in the coefficient learning apparatus
50 in Fig. 15 basically has the same configuration and function
as each of the low-pass filter 31, the sub-band division circuit
33, the characteristic amount calculation circuit 34 and the
pseudo high band sub-band power calculation circuit 35 in the
15 encoder 30 in Fig. 11, the description thereof is suitably
omitted.

[0233]

In other word, although the pseudo high band sub-band
power difference calculation circuit 55 provides the same
20 configuration and function as the pseudo high band sub-band
power difference calculation circuit 36 in Fig. 11, the
calculated pseudo high band sub-band power difference is
supplied to the pseudo high band sub-band power difference
clustering circuit 56 and the high band sub-band power
25 calculated when calculating the pseudo high band sub-band power
difference is supplied to the coefficient estimation circuit
57.

[0234]

The pseudo high band sub-band power difference
30 clustering circuit 56 clusters a pseudo high band sub-band
power difference vector obtained from a pseudo high band

sub-band power difference from the pseudo high band sub-band power difference calculation circuit 55 and calculates the representative vector at each cluster.

[0235]

5 The coefficient estimation circuit 57 calculates the high band sub-band power estimation coefficient for each cluster clustered by the pseudo high band sub-band power difference clustering circuit 56 based on a high band sub-band power from the pseudo high band sub-band power difference calculation circuit 55 and one or more characteristic amount from the characteristic amount calculation circuit 53.

[0236]

[Coefficient Learning Process of Coefficient Learning Apparatus]

15 Next, a coefficient learning process by the coefficient learning apparatus 50 in Fig. 15 will be described with reference to a flowchart in Fig. 16.

[0237]

20 In addition, the process of step S151 to S155 of a flowchart in Fig. 16 is identical with those of step S111, S113 to S116 of a flowchart in Fig. 12 except that signal input to the coefficient learning apparatus 50 is a broadband instruction signal, and thus the description thereof is omitted.

25 [0238]

 That is, in step S156, the pseudo high band sub-band power difference clustering circuit 56 clusters a plurality of pseudo high band sub-band power difference vectors (a lot of time frames) obtained from a pseudo high band sub-band power difference from the pseudo high band sub-band power difference calculation circuit 55 to 64 clusters and calculates the

representative vector for each cluster. As an example of a clustering method, for example, clustering by k-means method can be applied. The pseudo high band sub-band power difference clustering circuit 56 sets a center vector of each cluster obtained from the result performing clustering by k-means method to the representative vector of each cluster. In addition, a method of the clustering or the number of cluster is not limited thereto, but may apply other method.

[0239]

10 In addition, the pseudo high band sub-band power difference clustering circuit 56 measures distance between 64 representative vectors and the pseudo high band sub-band power difference vector obtained from the pseudo high band sub-band power difference calculation circuit 55 in the time frames 15 J and determines index CID(J) of the cluster included in the representative vector that has is the shortest distance. In addition, the index CID(J) takes an integer value of 1 to the number of the clusters (for example, 64). Therefore, the 20 pseudo high band sub-band power difference clustering circuit 56 outputs the representative vector and supplies the index CID(J) to the coefficient estimation circuit 57.

[0240]

In step S157, the coefficient estimation circuit 57 25 calculates a decoded high band sub-band power estimation coefficient at each cluster every set having the same index CID (J) (included in the same cluster) in a plurality of combinations of a number (eb-sb) of the high band sub-band power and the characteristic amount supplied to the same time 30 frames from the pseudo high band sub-band power difference calculation circuit 55 and the characteristic amount

calculation circuit 53. A method for calculating the coefficient by the coefficient estimation circuit 57 is identical with the method by the coefficient estimation circuit 24 of the coefficient learning apparatus 20 in Fig. 9. However,
5 the other method may be used.

[0241]

According to the processing described above, by using a predetermined broadband instruction signal, since a learning for the each representative vector of a plurality of clusters
10 in the specific space of the pseudo high band sub-band power difference predetermined in the high band encoding circuit 37 of the encoder 30 in Fig. 11 and a learning for the decoded high band sub-band power estimation coefficient output by the high band decoding circuit 45 of the decoder 40 in Fig. 13
15 is performed, it is possible to obtain the desired output result with respect to various input signals input to the encoder 30 and various input code string input to the decoder 40 and it is possible to reproduce a music signal having the high quality.

20 [0242]

In addition, with respect to encoding and decoding of the signal, the coefficient data for calculating the high band sub-band power in the pseudo high band sub-band power calculation circuit 35 of encoder 30 and the decoded high band
25 sub-band power calculation circuit 46 of the decoder 40 can be processed as follows. That is, it is possible to record the coefficient in the front position of code string by using the different coefficient data by the kind of the input signal.

[0243]

30 For example, it is possible to achieve an encoding efficiency improvement by changing the coefficient data by

a signal such as speech and jazz.

[0244]

Fig. 17 illustrates the code string obtained from the above method.

5 [0245]

The code string A in Fig. 17 encodes the speech and an optimal coefficient data α in the speech is recorded in a header.

[0246]

10 In this contrast, since the code string B in Fig. 17 encodes jazz, the optimal coefficient data β in the jazz is recorded in the header.

[0247]

The plurality of coefficient data described above can be easily learned by the same kind of the music signal in advance and the encoder 30 may select the coefficient data from genre information recorded in the header of the input signal. In addition, the genre is determined by performing a waveform analysis of the signal and the coefficient data may be selected. That is, a genre analysis method of signal is not limited in particular.

20

[0248]

When calculation time allows, the encoder 30 is equipped with the learning apparatus described above and thus the process is performed by using the coefficient dedicated to the signal and as illustrated in the code string C in Fig. 17, finally, it is also possible to record the coefficient in the header.

25

[0249]

An advantage using the method will be described as follow.

30 [0250]

A shape of the high band sub-band power includes a

plurality of similar positions in one input signal. By using characteristic of a plurality of input signals, and by performing the learning of the coefficient for estimating of the high band sub-band power every the input signal, separately, 5 redundancy due to in the similar position of the high band sub-band power is reduced, thereby improving encoding efficiency. In addition, it is possible to perform estimation of the high band sub-band power with higher accuracy than the learning of the coefficient for estimating the high band 10 sub-band power using a plurality of signals statistically. [0251]

Further, as described above, the coefficient data learned from the input signal in decoding can take the form to be inserted once into every several frames. 15 [0252]

<3. Third Embodiment>

[Functional Configuration Example of Encoder]

In addition, although it was described that the pseudo high band sub-band power difference ID is output from the 20 encoder 30 to the decoder 40 as the high band encoded data, the coefficient index for obtaining the decoded high band sub-band power estimation coefficient may be set as the high band encoded data. [0253]

25 In this case, the encoder 30, for example, is configured as illustrated in Fig. 18. In addition, in Fig. 18, parts corresponding to parts in Fig. 11 has the same numeral reference and the description thereof is suitably omitted. [0254]

30 The encoder 30 in Fig. 18 is the same except that the encoder 30 in Fig. 11 and the low band decoding circuit 39

are not provided and the remainder is the same.

[0255]

In the encoder 30 in Fig. 18, the characteristic amount calculation circuit 34 calculates the low band sub-band power as the characteristic amount by using the low band sub-band signal supplied from the sub-band division circuit 33 and is supplied to the pseudo high band sub-band power calculation circuit 35.

[0256]

In addition, in the pseudo high band sub-band power calculation circuit 35, a plurality of decoded high band sub-band power estimation coefficients obtained by the predetermined regression analysis is corresponded to a coefficient index specifying the decoded high band sub-band power estimation coefficient to be recorded.

[0257]

Specifically, sets of a coefficient $A_{ib}(kb)$ and a coefficient B_{ib} for each sub-band used in operation of Equation (2) described above are prepared in advance as the decoded high band sub-band power estimation coefficient. For example, the coefficient $A_{ib}(kb)$ and the coefficient B_{ib} are calculated by an regression analysis using a least-squares method by setting the low band sub-band power to an explanation variable and the high band sub-band power to an explained variable in advance. In the regression analysis, an input signal including the low band sub-band signal and the high band sub-band signal is used as the broadband instruction signal.

[0258]

The pseudo high band sub-band power calculation circuit 35 calculates the pseudo high band sub-band power of each sub-band of the high band side by using the decoded high band

sub-band power estimation coefficient and the characteristic amount from the characteristic amount calculation circuit 34 for each of a decoded high band sub-band power estimation coefficient recorded and supplies the sub-band power to the
 5 pseudo high band sub-band power difference calculation circuit 36.

[0259]

The pseudo high band sub-band power difference calculation circuit 36 compares the high band sub-band power
 10 obtained from the high band sub-band signal supplied from the sub-band division circuit 33 with the pseudo high band sub-band power from the pseudo high band sub-band power calculation circuit 35.

[0260]

15 In addition, the pseudo high band sub-band power difference calculation circuit 36 supplies the coefficient index of the decoded high band sub-band power estimation coefficient, in which the pseudo high band sub-band power closed to the highest pseudo high band sub-band power is
 20 obtained among the result of the comparison and a plurality of decoded high band sub-band power estimation coefficient to the high band encoding circuit 37. That is, the coefficient index of decoded high band sub-band power estimation coefficient from which the high band signal of the input signal
 25 to be reproduced in decoding that is the decoded high band signal closest to a true value is obtained.

[0261]

[Encoding Process of Encoder]

Next, referring to a flow chart in Fig. 19, an encoding
 30 process performing by the encoder 30 in Fig. 18 will be described. In addition, processing of step S181 to step S183 are identical

with those of step S111 to S113 in Fig 12. Therefore, the description thereof is omitted.

[0262]

In step S184, the characteristic amount calculation circuit 34 calculates characteristic amount by using the low band sub-band signal from the sub-band division circuit 33 and supplies the characteristic amount to the pseudo high band sub-band power calculation circuit 35.

[0263]

Specially, the characteristic amount calculation circuit 34 calculates as a characteristic amount, the low band sub-band power $\text{power}(ib, J)$ of the frames J (where, $0 \leq J$) with respect to each sub-band ib (where, $sb-3 \leq ib \leq sb$) in a low band side by performing operation of Equation (1) described above. That is, the low band sub-band power $\text{power}(ib, J)$ calculates by digitizing a square mean value of the sample value of each sample of the low band sub-band signal constituting the frames J .

[0264]

In step S185, the pseudo high band sub-band power calculation circuit 35 calculates the pseudo high band sub-band power based on the characteristic amount supplied from the characteristic amount calculation circuit 34 and supplies the pseudo high band sub-band power to the pseudo high band sub-band power difference calculation circuit 36.

[0265]

For example, the pseudo high band sub-band power calculation circuit 35 calculates the pseudo high band sub-band power $\text{power}_{\text{est}}(ib, J)$, which performs above-mentioned Equation (2) by using the coefficient $A_{ib}(kb)$ and the coefficient B_{ib} recorded as the decoded high band sub-band power coefficient

in advance and the pseudo high band sub-band power $power_{est}(ib, J)$ which performs the operation the above-mentioned Equation (2) by using the low band sub-band power $power(kb, J)$ (where, $sb-s \leq kb \leq sb$).

5 [0266]

That is, coefficient $A_{ib}(kb)$ for each sub-band multiplies the low band sub-band power $power(kb, J)$ of each sub-band of the low band side supplied as the characteristic amount and the coefficient B_{ib} is added to the sum of the low band sub-band power by which the coefficient is multiplied and then becomes the pseudo high band sub-band power $power_{est}(ib, J)$. This pseudo high band sub-band power is calculated for each sub-band of the high band side in which the index is $sb+1$ to eb

10 [0267]

15 In addition, the pseudo high band sub-band power calculation circuit 35 performs the calculation of the pseudo high band sub-band power for each decoded high band sub-band power estimation coefficient recorded in advance. For example, it is assumed that the coefficient index allows 1 to K (where, $2 \leq K$) number of decoding high band sub-band estimation coefficient to be prepared in advance. In this case, the pseudo high band sub-band power of each sub-band is calculated for each of the K decoded high band sub-band power estimation coefficients.

20 [0268]

In step S186, the pseudo high band sub-band power difference calculation circuit 36 calculates the pseudo high band sub-band power difference based on a high band sub-band signal from the sub-band division circuit 33, and the pseudo high band sub-band power from the pseudo high band sub-band power calculation circuit 35.

25 [0269]

30

[0269]

Specifically, the pseudo high band sub-band power difference calculation circuit 36 does not perform the same operation as the Equation (1) described above and calculates the high band sub-band power $\text{power}(\text{ib}, J)$ in the frames J with respect to high band sub-band signal from the sub-band division circuit 33. In addition, in the embodiment, the whole of the sub-band of the low band sub-band signal and the high band sub-band signal is distinguished by using index ib .

10 [0270]

Next, the pseudo high band sub-band power difference calculation circuit 36 performs the same operation as the Equation (14) described above and calculates the difference between the high band sub-band power $\text{power}(\text{ib}, J)$ in the frames J and the pseudo high band sub-band power $\text{power}_{\text{est}}(\text{ib}, J)$. In this case, the pseudo high band sub-band power difference $\text{power}_{\text{diff}}(\text{ib}, J)$ is obtained for each decoded high band sub-band power estimation coefficient with respect to each sub-band of the high band side which index is $\text{sb}+1$ to eb .

20 [0271]

In step S187, the pseudo high band sub-band power difference calculation circuit 36 calculates the following Equation (15) for each decoded high band sub-band power estimation coefficient and calculates a sum of squares of the pseudo high band sub-band power difference.

[0272]

[Equation 15]

$$E(J, \text{id}) = \sum_{\text{ib}=\text{sb}+1}^{\text{eb}} \{\text{power}_{\text{diff}}(\text{ib}, J, \text{id})\}^2 \quad \dots (15)$$

[0273]

In addition, in Equation (15), the square sum for a difference $E(J, id)$ is obtained with respect to the decoded high band sub-band power estimation coefficient in which the coefficient index is id and the frames J . In addition, in
 5 Equation (15), $power_{diff}(ib, J, id)$ is obtained with respect to the decoded high band sub-band power estimation coefficient in which the coefficient index is id decoded high band sub-band power and shows the pseudo high band sub-band power difference ($power_{diff}(ib, J)$) of the pseudo high band sub-band power
 10 difference $power_{diff}(ib, J)$ of the frames J of the sub-band which the index is ib . The square sum of a difference $E(J, id)$ is calculated with respect to the number of K of each decoded high band sub-band power estimation coefficient.

[0274]

15 The square sum for a difference $E(J, id)$ obtained above shows a similar degree of the high band sub-band power calculated from the actual high band signal and the pseudo high band sub-band power calculated using the decoded high band sub-band power estimation coefficient, which the
 20 coefficient index is id .

[0275]

That is, the error of the estimation value is shown with respect to the true value of the high band sub-band power. Therefore, the smaller the square sum for the difference $E(J, id)$, the more the decoded high band signal closed by the actual
 25 high band signal is obtained by the operation using the decoded high band sub-band power estimation coefficient. That is, the decoded high band sub-band power estimation coefficient in which the square sum for the difference $E(J, id)$ is minimum
 30 is an estimation coefficient most suitable for the frequency band expansion process performed in decoding the output code

string.

[0276]

The pseudo high band sub-band power difference calculation circuit 36 selects the square sum for difference having a minimum value among the K square sums for difference E (J, id) and supplies the coefficient index showing the decoded high band sub-band power estimation coefficient corresponding to the square sum for difference to the high band encoding circuit 37.

10 [0277]

In step S188, the high band encoding circuit 37 encodes the coefficient index supplied from the pseudo high band sub-band power difference calculation circuit 36 and supplies obtained high band encoded data to the multiplexing circuit 38.

15 38.

[0278]

For example, step S188, an entropy encoding and the like is performed with respect to the coefficient index. Therefore, information amount of the high band encoded data output to the decoder 40 can be compressed. In addition, if high band encoded data is information that an optimal decoded high band sub-band power estimation coefficient is obtained, any information is preferable; for example, the index may be the high band encoded data as it is.

25 [0279]

In step S189, the multiplexing circuit 38 multiplexes the low band encoded data supplied from the low band encoding circuit 32 and the high band encoded data supplied from the high band encoding circuit 37 and outputs the output code string and the encoding process is completed.

30 and the encoding process is completed.

[0280]

As described above, the decoded high band sub-band power estimation coefficient mostly suitable to process can be obtained by outputting the high band encoded data obtained by encoding the coefficient index as the output code string in decoder 40 receiving an input of the output code string, together with the low frequency encoded data. Therefore, it is possible to obtain signal having higher quality.

[0281]

[Functional Configuration Example of Decoder]

10 In addition, the output code string output from the encoder 30 in Fig. 18 is input as the input code string and for example, the decoder 40 for decoding is configuration illustrated in Fig. 20. In addition, in Fig. 20, the parts corresponding to the case Fig. 13 use the same symbol and the description is omitted.

[0282]

20 The decoder 40 in Fig. 20 is identical with the decoder 40 in Fig. 13 in that the demultiplexing circuit 41 to the synthesis circuit 48 is configured, but is different from the decoder 40 in Fig. 13 in that the decoded low band signal from the low band decoding circuit 42 is supplied to the characteristic amount calculation circuit 44.

[0283]

25 In the decoder 40 in Fig. 20, the high band decoding circuit 45 records the decoded high band sub-band power estimation coefficient identical with the decoded high band sub-band power estimation coefficient in which the pseudo high band sub-band power calculation circuit 35 in Fig. 18 is recorded in advance. That is, the set of the coefficient $A_{1b}(kb)$ and coefficient B_{1b} as the decoded high band sub-band power estimation coefficient by the regression analysis is

recorded to correspond to the coefficient index.

[0284]

The high band decoding circuit 45 decodes the high band encoded data supplied from the demultiplexing circuit 41 and supplies the decoded high band sub-band power estimation coefficient indicated by the coefficient index obtained from the result to the decoded high band sub-band power calculation circuit 46.

[0285]

10 [Decoding Process of Decoder]

Next, the decoding process performs by decoder 40 in Fig. 20 will be described with reference to a flowchart in Fig. 21.

[0286]

15 The decoding process starts if the output code string output from the encoder 30 is provided as the input code string to the decoder 40. In addition, since the processes of step S211 to step S213 is identical with those of step S131 to step S133 in Fig. 14, the description is omitted.

20 [0287]

In step S214, the characteristic amount calculation circuit 44 calculates the characteristic amount by using the decoded low band sub-band signal from the sub-band division circuit 43 and supplies it decoded high band sub-band power calculation circuit 46. In detail, the characteristic amount calculation circuit 44 calculates the characteristic amount of the low band sub-band power $power(ib, J)$ of the frames J (but, $0 \leq J$) by performing operation of the Equation (1) described above with respect to the each sub-band ib of the low band side.

[0288]

In step S215, the high band decoding circuit 45 performs decoding of the high band encoded data supplied from the demultiplexing circuit 41 and supplies the decoded high band sub-band power estimation coefficient indicated by the
 5 coefficient index obtained from the result to the decoded high band sub-band power calculation circuit 46. That is, the decoded high band sub-band power estimation coefficient is output, which is indicated by the coefficient index obtained by the decoding in a plurality of decoded high band sub-band
 10 power estimation coefficient recorded to the high band decoding circuit 45 in advance.

[0289]

In step S216, the decoded high band sub-band power calculation circuit 46 calculates the decoded high band
 15 sub-band power based on the characteristic amount supplied from the characteristic amount calculation circuit 44 and the decoded high band sub-band power estimation coefficient supplied from the high band decoding circuit 45 and supplies it to the decoded high band signal production circuit 47.

20 [0290]

That, the decoded high band sub-band power calculation circuit 46 performs operation the Equation (2) described above using the coefficient $A_{1b}(kb)$ as the decoded high band sub-band power estimation coefficient and the low band sub-band power
 25 power(kb, J) and the coefficient B_{1b} (where, $sb-3 \leq kb \leq sb$) as characteristic amount and calculates the decoded high band sub-band power. Therefore, the decoded high band sub-band power is obtained with respect to each sub-band of the high band side, which the index is $sb+1$ to eb .

30 [0291]

In step S217, the decoded high band signal production

circuit 47 produces the decoded high band signal based on the decoded low band sub-band signal supplied from the sub-band division circuit 43 and the decoded high band sub-band power supplied from the decoded high band sub-band power calculation circuit 46.

[0292]

In detail, the decoded high band signal production circuit 47 performs operation of the above-mentioned Equation (1) using the decoded low band sub-band signal and calculates the low band sub-band power with respect to each sub-band of the low band side. In addition, the decoded high band signal production circuit 47 calculates the gain amount $G(ib, J)$ for each sub-band of the high band side by performing operation of the Equation (3) described above using the low band sub-band power and the decoded high band sub-band power obtained.

[0293]

Further, the decoded high band signal production circuit 47 produces the high band sub-band signal $x3(ib, n)$ by performing the operation of the Equations (5) and (6) described above using the gain amount $G(ib, J)$ and the decoded low band sub-band signal with respect to each sub-band of the high band side.

[0294]

That is, the decoded high band signal production circuit 47 performs an amplitude modulation of the decoded high band sub-band signal $x(ib, n)$ in response to the ratio of the low band sub-band power to the decoded high band sub-band power and thus performs frequency-modulation the decoded low band sub-band signal ($x2(ib, n)$) obtained. Therefore, the signal of the frequency component of the sub-band of the low band side is converted to signal of the frequency component of the

sub-band of the high band side and the high band sub-band signal $x3(ib, n)$ is obtained.

[0295]

As described above, the processes for obtaining the high
5 band sub-band signal of each sub-band is a process described
below in more detail.

[0296]

The four sub-bands being a line in the frequency area
is referred to as the band block and the frequency band is
10 divided so that one band block (hereinafter, referred to as
a low band block) is configured from four sub-bands in which
the index existed in the low side is sb to $sb-3$. In this case,
for example, the band including the sub-band in which the index
of the high band side includes $sb+1$ to $sb+4$ is one band block.
15 In addition, the high band side, that is, a band block including
sub-band in which the index is $sb+1$ or more is particularly
referred to as the high band block.

[0297]

In addition, attention is paid to one sub-band
20 constituting the high band block and the high band sub-band
signal of the sub-band (hereinafter, referred to as attention
sub-band) is produced. First, the decoded high band signal
production circuit 47 specifies the sub-band of the low band
block that has the same position relation to the position of
25 the attention sub-band in the high band block.

[0298]

For example, if the index of the attention sub-band is
 $sb+1$, the sub-band of the low band block having the same position
relation with the attention sub-band is set as the sub-band
30 that the index is $sb-3$ since the attention sub-band is a band
that the frequency is the lowest in the high band blocks.

[0299]

As described above, the sub-band, if the sub-band of the low band block sub-band having the same position relationship of the attention sub-band is specific, the low
5 band sub-band power and the decoded low band sub-band signal and the decoded high band sub-band power is used and the high band sub-band signal of the attention sub-band is produced.

[0300]

That is, the decoded high band sub-band power and the
10 low band sub-band power are substituted for Equation (3), so that the gain amount according to the rate of the power thereof is calculated. In addition, the calculated gain amount is multiplied by the decoded low band sub-band signal, the decoded low band sub-band signal multiplied by the gain amount is set
15 as the frequency modulation by the operation of the Equation (6) to be set as the high band sub-band signal of the attention sub-band.

[0301]

In the processes, the high band sub-band signal of the
20 each sub-band of the high band side is obtained. In addition, the decoded high band signal production circuit 47 performs the Equation (7) described above to obtain sum of the each high band sub-band signal and to produce the decoded high band signal. The decoded high band signal production circuit 47
25 supplies the obtained decoded high band signal to the synthesis circuit 48 and the process precedes from step S217 to the step S218 and then the decoding process is terminated.

[0302]

In step S218, the synthesis circuit 48 synthesizes the
30 decoded low band signal from the low band decoding circuit 42 and the decoded high band signal from the decoded high band

signal production circuit 47 and outputs as the output signal.
[0303]

As described above, since decoder 40 obtained the coefficient index from the high band encoded data obtained from the demultiplexing of the input code string and calculates the decoded high band sub-band power by the decoded high band sub-band power estimation coefficient indicated by using the decoded high band sub-band power estimation coefficient indicated by the coefficient index, it is possible to improve the estimation accuracy of the high band sub-band power. Therefore, it is possible to produce the music signal having high quality.

[0304]

<4. Fourth Embodiment>

15 [Encoding Processes of Encoder]

First, in as described above, the case that only the coefficient index is included in the high band encoded data is described. However, the other information may be included.
[0305]

20 For example, if the coefficient index is included in the high band encoded data, the decoding high band sub-band power estimation coefficient that the decoded high band sub-band power closest to the high band sub-band power of the actual high band signal is notified of the decoder 40 side.

25 [0306]

Therefore, the actual high band sub-band power (true value) and the decoded high band sub-band power (estimation value) obtained from the decoder 40 produces difference substantially equal to the pseudo high band sub-band power difference $\text{power}_{\text{diff}}(\text{ib}, J)$ calculated from the pseudo high band sub-band power difference calculation circuit 36.

[0307]

Herein, if the coefficient index and the pseudo high band sub-band power difference of the sub-band is included in the high band encoded data, the error of the decoded high band sub-band power regarding the actual high band sub-band power is approximately known in the decoder 40 side. If so, it is possible to improve the estimation accuracy of the high band sub-band power using the difference.

[0308]

10 The encoding process and the decoding process in a case where the pseudo high band sub-band power difference is included in the high band encoded data will be described with reference with a flow chart of Figs. 22 and 23.

[0309]

15 First, the encoding process performed by encoder 30 in Fig. 18 will be described with reference to the flowchart in Fig. 22. In addition, the processes of step S241 to step S246 is identical with those of step S181 to step S186 in Fig. 19. Therefore, the description thereof is omitted.

20 [0310]

In step S247, the pseudo high band sub-band power difference calculation circuit 36 performs operation of the Equation (15) described above to calculate sum $E(J, id)$ of squares for difference for each decoded high band sub-band power estimation coefficient.

[0311]

In addition, the pseudo high band sub-band power difference calculation circuit 36 selects sum of squares for difference where the sum of squares for difference is set as a minimum in the sum of squares for difference among sum $E(J, id)$ of squares for difference and supplies the coefficient

index indicating the decoded high band sub-band power estimation coefficient corresponding to the sum of square for difference to the high band encoding circuit 37.

[0312]

5 In addition, the pseudo high band sub-band power difference calculation circuit 36 supplies the pseudo high band sub-band power difference $\text{power}_{\text{diff}}(\text{ib}, J)$ of the each sub-band obtained with respect to the decoded high band sub-band power estimation coefficient corresponding to
10 selected sum of squares of residual error to the high band encoding circuit 37.

[0313]

 In step S248, the high band encoding circuit 37 encodes the coefficient index and the pseudo high band sub-band power
15 difference supplied from the pseudo high band sub-band power difference calculation circuit 36 and supplies the high band encoded data obtained from the result to the multiplexing circuit 38.

[0314]

20 Therefore, the pseudo high band sub-band power difference of the each sub-band power of the high band side where the index is $\text{sb}+1$ to eb , that is, the estimation difference of the high band sub-band power is supplied as the high band encoded data to the decoder 40.

25 [0315]

 If the high band encoded data is obtained, after this, encoding process of step S249 is performed to terminate encoding process. However, the process of step S249 is identical with the process of step S189 in Fig. 19. Therefore,
30 the description is omitted.

[0316]

As described above, if the pseudo high band sub-band power difference is included in the high band encoded data, it is possible to improve estimation accuracy of the high band sub-band power and to obtain music signal having good quality
5 in the decoder 40.

[0317]

[Decoding Processing of Decoder]

Next, a decoding process performed by the decoder 40 in Fig. 20 will be described with reference to a flowchart
10 in Fig. 23. In addition, the process of step S271 to step S274 is identical with those of step S211 to step S214 in Fig. 21. Therefore, the description thereof is omitted.

[0318]

In step S275, the high band decoding circuit 45 performs
15 the decoding of the high band encoded data supplied from the demultiplexing circuit 41. In addition, the high band decoding circuit 45 supplies the decoded high band sub-band power estimation coefficient indicated by the coefficient index obtained by the decoding and the pseudo high band sub-band
20 power difference of each sub-band obtained by the decoding to the decoded high band sub-band power calculation circuit 46.

[0319]

In a step S276, the decoded high band sub-band power
25 calculation circuit 46 calculates the decoded high band sub-band power based on the characteristic amount supplied from the characteristic amount calculation circuit 44 and the decoded high band sub-band power estimation coefficient 216 supplied from the high band decoding circuit 45. In addition,
30 step S276 has the same process as step S216 in Fig. 21.

[0320]

In step S277, the decoded high band sub-band power calculation circuit 46 adds the pseudo high band sub-band power difference supplied from the high band decoding circuit 45 to the decoded high band sub-band power and supplies the added
5 result as an ultimate decoded high band sub-band power to decoded high band signal production circuit 47.

That is, the pseudo high band sub-band power difference of the same sub-band is added to the decoding high band sub-band power of the each calculated sub-band.

10 [0321]

In addition, after that, processes of step S278 and step S279 is performed and the decoding process is terminated. However, their processes are identical with step S217 and step S218 in Fig. 21. Therefore, the description will be omitted.

15 [0322]

By doing the above, the decoder 40 obtains the coefficient index and the pseudo high band sub-band power from the high band encoded data obtained by the demultiplexing of the input code string. In addition, decoder 40 calculates the decode
20 high band sub-band power using the decoded high band sub-band power estimation coefficient indicated by the coefficient index and the pseudo high band sub-band power difference. Therefore, it is possible to improve accuracy of the high band sub-band power and to reproduce music signal having high sound
25 quality.

[0323]

In addition, the difference of the estimation value of the high band sub-band power producing between encoder 30 and decoder 40, that is, the difference (hereinafter, referred
30 to as an difference estimation between device) between the pseudo high band sub-band power and decoded high band sub-band

power may be considered.

[0324]

In this case, for example, the pseudo high band sub-band power difference serving as the high band encoded data is corrected by the difference estimation between devices and the estimation difference between devices is included in the high band encoded data, the pseudo high band sub-band power difference is corrected by the estimation difference between apparatus in decoder 40 side. In addition, the estimation difference between apparatus may be recorded in decoder 40 side in advance and the decoder 40 may make correction by adding the estimation difference between devices to the pseudo high band sub-band power difference. Therefore, it is possible to obtain the decoded high band signal closed to the actual high band signal.

[0325]

<5. Fifth Embodiment>

In addition, in the encoder 30 in Fig. 18, it is described that the pseudo high band sub-band power difference calculation circuit 36 selects the optimal index from a plurality of coefficient indices using the square sum $E(J, id)$ of for a difference. However, the circuit may select the coefficient index using the index different from the square sum for a difference.

25 [0326]

For example, as an index selecting a coefficient index, mean square value, maximum value and an average value of a residual error of the high band sub-band power and the pseudo high band sub-band power may be used. In this case, the encoder 30 in Fig. 18 performs encoding process illustrated in a flowchart in Fig. 24.

[0327]

An encoding process using the encoder 30 will be described with reference to a flowchart in Fig. 24. In addition, processes of step S301 to step S305 are identical with those of step S181 to step S185 in Fig. 19. Therefore, the description will be omitted. If the processes of step S301 to step S305 are performed, the pseudo high band sub-band power of each sub-band is calculated for each K number of decoded high band sub-band power estimation coefficient.

10 [0328]

In step S306, the pseudo high band sub-band power difference calculation circuit 36 calculates an estimation value $Res(id, J)$ using a current frame J to be processed for each K number of decoded high band sub-band power estimation coefficient.

15

[0329]

In detail, the pseudo high band sub-band power difference calculation circuit 36 calculates the high band sub-band power $power(ib, J)$ in frames J by performing the same operation as the Equation (1) described above using the high band sub-band signal of each sub-band supplied from the sub-band division circuit 33. In addition, in an embodiment of the present invention, it is possible to discriminate all of the sub-band of the low band sub-band signal and the high band sub-band using index ib .

25

[0330]

If the high band sub-band power $power(ib, J)$ is obtained, the pseudo high band sub-band power difference calculation circuit 36 calculates the following Equation (16) and calculates the residual square mean square value $Res_{std}(id, J)$.

30

[0331]

[Equation 16]

$$\text{Res}_{\text{std}}(\text{id}, J) = \sum_{\text{ib}=\text{sb}+1}^{\text{eb}} \{\text{power}(\text{ib}, J) - \text{power}_{\text{est}}(\text{ib}, \text{id}, J)\}^2 \quad \cdot \cdot \cdot (16)$$

[0332]

That is, the difference between the high band sub-band
 5 power $\text{power}(\text{ib}, J)$ and the pseudo high band sub-band power
 $\text{power}_{\text{est}}(\text{ib}, \text{id}, J)$ is obtained with respect to each sub-band
 on the high band side where the index $\text{sb}+1$ to eb and square
 sum for the difference becomes the residual square mean value
 $\text{Res}_{\text{std}}(\text{id}, J)$. In addition, the pseudo high band sub-band
 10 power $\text{power}_{\text{rest}}(\text{ibh}, \text{id}, J)$ indicates the pseudo high band
 sub-band power of the frames J of the sub-band where the index
 is ib , which is obtained with respect to the decoded high band
 sub-band power estimation coefficient where index is ib .

[0333]

15 Continuously, the pseudo high band sub-band power
 difference calculation circuit 36 calculates the following
 Equation (17) and calculates the residual maximum value
 $\text{Res}_{\text{max}}(\text{id}, J)$.

[0334]

20 [Equation 17]

$$\text{Res}_{\text{max}}(\text{id}, J) = \max_{\text{ib}} \{|\text{power}(\text{ib}, J) - \text{power}_{\text{est}}(\text{ib}, \text{id}, J)|\} \quad \cdot \cdot \cdot (17)$$

[0335]

In addition, in an Equation (17),
 $\max_{\text{ib}} \{|\text{power}(\text{ib}, J) - \text{power}_{\text{est}}(\text{ib}, \text{id}, J)|\}$ indicates a maximum
 25 value among absolute value of the difference between the high
 band sub-band power $\text{power}(\text{ib}, J)$ of each sub-band where the
 index is $\text{sb}+1$ to eb and the pseudo high band sub-band power

power_{est}(ib,id,J). Therefore, a maximum value of the absolute value of the difference between the high band sub-band power power(ib,J) in the frames J and the pseudo high band sub-band power power_{est}(ib,id,J) is set as the residual difference maximum value Res_{max}(id,J).

[0336]

In addition, the pseudo high band sub-band power difference calculation circuit 36 calculates the following Equation (18) and calculates the residual average value Res_{ave}(id,J).

[0337]

[Equation 18]

$$\text{Res}_{\text{ave}}(\text{id}, J) = \left| \left(\sum_{\text{ib}=\text{sb}+1}^{\text{eb}} \{ \text{power}(\text{ib}, J) - \text{power}_{\text{est}}(\text{ib}, \text{id}, J) \} \right) / (\text{eb} - \text{sb}) \right| \quad \cdot \cdot \cdot (18)$$

[0338]

That is, for each sub-band on the high band side in which the index is sb+1 to eb, the difference between the high band sub-band power power(ib,J) of the frames J and the pseudo high band sub-band power power_{est}(ib,id,J) is obtained and the sum of the difference is obtained. In addition, the absolute value of a value obtained by dividing the sum of the obtained difference by the number of the sub-bands (eb - sb) of the high band side is set as the residual average value Res_{ave}(id,J). The residual average value Res_{ave}(id,J) indicates a size of the average value of the estimation error of each sub-band that a symbol is considered.

[0339]

In addition, if the residual square mean Res_{std}(id,J), the residual difference maximum value Res_{max}(id,J), and the

residual average value $\text{Res}_{\text{ave}}(\text{id}, J)$ are obtained, the pseudo high band sub-band power difference calculation circuit 36 calculates the following Equation (19) and calculates an ultimate estimation value $\text{Res}(\text{id}, J)$.

5 [0340]

[Equation 19]

$$\text{Res}(\text{id}, J) = \text{Res}_{\text{std}}(\text{id}, J) + W_{\text{max}} \times \text{Res}_{\text{max}}(\text{id}, J) + W_{\text{ave}} \times \text{Res}_{\text{ave}}(\text{id}, J) \quad \cdot \cdot \cdot (19)$$

[0341]

That is, the residual square average value $\text{Res}_{\text{std}}(\text{id}, J)$,
 10 the residual maximum value $\text{Res}_{\text{max}}(\text{id}, J)$ and the residual average value $\text{Res}_{\text{ave}}(\text{id}, J)$ are added with weight and set as an ultimate estimation value $\text{Res}(\text{id}, J)$. In addition, in the Equation (19), W_{max} and W_{ave} is a predetermined weight and for example, $W_{\text{max}}=0.5$, $W_{\text{ave}}=0.5$.

15 [0342]

The pseudo high band sub-band power difference calculation circuit 36 performs the above process and calculates the estimation value $\text{Res}(\text{id}, J)$ for each of the K numbers of the decoded high band sub-band power estimation
 20 coefficient, that is, the K number of the coefficient index id.

[0343]

In step S307, the pseudo high band sub-band power difference calculation circuit 36 selects the coefficient
 25 index id based on the estimation value Res for each of the obtained (id, J) coefficient index id.

[0344]

The estimation value $\text{Res}(\text{id}, J)$ obtained from the process described above shows a similarity degree between the high

band sub-band power calculated from the actual high band signal and the pseudo high band sub-band power calculated using the decoded high band sub-band power estimation coefficient which is the coefficient index id . That is, a size of the estimation error of the high band component is indicated.

[0345]

Accordingly, as the evaluation $Res(id, J)$ become low, the decoded high band signal closer to the actual high band signal is obtained by an operation using the decoded high band sub-band power estimation coefficient. Therefore, the pseudo high band sub-band power difference calculation circuit 36 selects the estimation value which is set as a minimum value among the K numbers of the estimation value $Res(id, J)$ and supplies the coefficient index indicating the decoded high band sub-band power estimation coefficient corresponding to the estimation value to the high band encoding circuit 37.

[0346]

If the coefficient index is output to the high band encoding circuit 37, after that, the processes of step S308 and step S309 are performed, the encoding process is terminated. However, since the processes are identical with step S188 in Fig. 19 and step S189, the description thereof will be omitted.

[0347]

As described above, in the encoder 30, the estimation value $Res(id, J)$ calculated by using the residual square average value $Res_{std}(id, J)$, the residual maximum value $Res_{max}(id, J)$ and the residual average value $Res_{ave}(id, J)$ is used, and the coefficient index of the an optimal decoded high band sub-band power estimation coefficient is selected.

[0348]

If the estimation value $Res(id, J)$ is used, since an

estimation accuracy of the high band sub-band power is able to be evaluated using the more estimation standard compared with the case using the square sums for difference, it is possible to select more suitable decoded high band sub-band power estimation coefficient. Therefore, when using, the decoder 40 receiving the input of the output code string, it is possible to obtain the decoded high band sub-band power estimation coefficient, which is mostly suitable to the frequencybandexpansionprocessandsignalhavinghighersound quality.

[0349]

<Modification Example 1>

In addition, if the encoding process described above is performed for each frame of the input signal, There may be a case where the coefficient index different in each consecutive frame is selected in a stationary region that the time variation of the high band sub-band power of each sub-band of the high band side of the input signal is small.

[0350]

That is, since the high band sub-band power of each frame has almost identical values in consecutive frames constituting the standard region of the input signal, the same coefficient index should be continuously selected in their frame. However, the coefficient index selected for each frame in a section of the consecutive frames is changed and thus the high band component of the voice reproduced in the decoder 40 side may be no longer stationary. If so, incongruity in auditory occurs in the reproduced sound.

[0351]

Accordingly, if the coefficient index is selected in the encoder 30, the estimation result of the high band component

in the previous frame in time may be considered. In this case, encoder 30 in Fig. 18 performs the encoding process illustrated in the flowchart in Fig. 25.

[0352]

5 As described below, an encoding process by the encoder 30 will be described with reference to the flowchart in Fig. 25. In addition, the processes of step S331 to step S336 are identical with those of step S301 to step S306 in Fig. 24. Therefore, the description thereof will be omitted.

10 [0353]

The pseudo high band sub-band power difference calculation circuit 36 calculates the estimation value $\text{ResP}(\text{id}, J)$ using a past frame and a current frame in step S337.

[0354]

15 Specifically, the pseudo high band sub-band power difference calculation circuit 36 records the pseudo high band sub-band power of each sub-band obtained by the decoded high band sub-band power estimation coefficient of the coefficient index selected finally with respect to frames $J-1$ earlier than frame J to be processed by one in time. Herein, the finally selected coefficient index is referred to as a coefficient index output to the decoder 40 by encoding using the high band encoding circuit 37.

20

[0355]

25 As described below, in particular, the coefficient index id selected in frame $(J-1)$ is set to as $\text{id}_{\text{selected}}(J-1)$. In addition, the pseudo high band sub-band power of the sub-band that the index obtained by using the decoded high band sub-band power estimation coefficient of the coefficient index

30 $\text{id}_{\text{selected}}(J-1)$ is ib (where, $\text{sb}+1 \leq \text{ib} \leq \text{eb}$) is continuously explained as $\text{power}_{\text{est}}(\text{ib}, \text{id}_{\text{selected}}(J-1), J-1)$.

[0356]

The pseudo high band sub-band power difference calculation circuit 36 calculates firstly following Equation (20) and then the estimation residual square mean value

5 ResP_{std}(id,J).

[0357]

[Equation 20]

$$\text{ResP}_{\text{std}}(\text{id}, J) = \sum_{\text{ib}=\text{sb}+1}^{\text{eb}} \{ \text{power}_{\text{est}}(\text{ib}, \text{id}_{\text{selected}}(J-1), J-1) - \text{power}_{\text{est}}(\text{ib}, \text{id}, J) \}^2 \quad \cdot \cdot \cdot (20)$$

[0358]

10 That is, the difference between the pseudo high band sub-band power $\text{power}_{\text{est}}(\text{ib}, \text{id}_{\text{selected}}(J-1), J-1)$ of the frame J-1 and the pseudo high band sub-band power $\text{power}_{\text{est}}(\text{ib}, \text{id}, J)$ of the frame J is obtained with respect to each sub-band of the high band side where the index is sb+1 to eb. In addition,
 15 the square sum for difference thereof is set as estimation error difference square average value ResP_{std}(id,J). In addition, the pseudo high band sub-band power $-(\text{power}_{\text{est}}(\text{ib}, \text{id}, J))$ shows the pseudo high band sub-band power of the frames (J) of the sub-band which the index is ib which
 20 is obtained with respect to the decoded high band sub-band power estimation coefficient where the coefficient index is id.

[0359]

Since this estimation residual square value
 25 ResP_{std}(id,J) is the of square sum for the difference of pseudo high band sub-band power between frames that is continuous in time, the smaller the estimation residual square mean ResP_{std}(id,J) is, the smaller the time variation of the

estimation value of the high band component is.

[0360]

Continuously, the pseudo high band sub-band power difference calculation circuit 36 calculates the following
 5 Equation (21) and calculates the estimation residual maximum value $\text{ResP}_{\max}(\text{id}, \text{J})$.

[0361]

[Equation 21]

$$\text{ResP}_{\max}(\text{id}, \text{J}) = \max_{\text{ib}} \{ |\text{power}_{\text{est}}(\text{ib}, \text{id}_{\text{selected}}(\text{J}-1), \text{J}-1) - \text{power}_{\text{est}}(\text{ib}, \text{id}, \text{J})| \} \quad \cdot \cdot \cdot (21)$$

10 [0362]

In addition, in the Equation (21),
 $\max_{\text{ib}} \{ |\text{power}_{\text{est}}(\text{ib}, \text{id}_{\text{selected}}(\text{J}-1), \text{J}-1) - \text{power}_{\text{est}}(\text{ib}, \text{id}, \text{J})| \}$
 indicates the maximum absolute value of the difference between
 the pseudo high band sub-band power
 15 $\text{power}_{\text{est}}(\text{ib}, \text{id}_{\text{selected}}(\text{J}-1), \text{J}-1)$ of each sub-band in which the
 index is sb+1 to eb and the pseudo high band sub-band power
 $\text{power}_{\text{est}}(\text{ib}, \text{id}, \text{J})$. Therefore, the maximum value of the
 absolute value of the difference between frames which is
 continuous in time is set as the estimation residual error
 20 difference maximum value $\text{ResP}_{\max}(\text{id}, \text{J})$.

[0363]

The smaller the estimation residual error maximum value
 $\text{ResP}_{\max}(\text{id}, \text{J})$ is, the closer estimation result of the high
 band component between the consecutive frames is closed.

25 [0364]

If the estimation residual maximum value $\text{ResP}_{\max}(\text{id}, \text{J})$
 is obtained, next, the pseudo high band sub-band power
 difference calculation circuit 36 calculates the following

Equation (22) and calculates the estimation residual average value $\text{ResP}_{\text{ave}}(\text{id}, J)$.

[0365]

[Equation 22]

$$\text{ResP}_{\text{ave}}(\text{id}, J) = \left| \left(\sum_{\text{ib}=\text{sb}+1}^{\text{eb}} \{ \text{power}_{\text{est}}(\text{ib}, \text{id}_{\text{selected}}(J-1), J-1) - \text{power}_{\text{est}}(\text{ib}, \text{id}, J) \} \right) / (\text{eb} - \text{sb}) \right| \quad \cdot \cdot \cdot (22)$$

5

[0366]

That is, the difference between the pseudo high band sub-band power $\text{power}_{\text{est}}(\text{ib}, \text{id}_{\text{selected}}(J-1), J-1)$ of the frame (J-1) and the pseudo high band sub-band power $\text{power}_{\text{est}}(\text{ib}, \text{id}, J)$ of the frame J is obtained with respect to each sub-band of the high band side when the index is sb+1 to eb. In addition, the absolute value of the value obtained by dividing the sum of the difference of each sub-band by the number of the sub-bands (eb - sb) of the high band side is set as the estimation residual average $\text{ResP}_{\text{ave}}(\text{id}, J)$. The estimation residual error average value $\text{ResP}_{\text{ave}}(\text{id}, J)$ shows the size of the average value of the difference of the estimation value of the sub-band between the frames where the symbol is considered.

15

[0367]

In addition, if the estimation residual square mean value $\text{ResP}_{\text{std}}(\text{id}, J)$, the estimation residual error maximum value $\text{ResP}_{\text{max}}(\text{id}, J)$ and the estimation residual average value $\text{ResP}_{\text{ave}}(\text{id}, J)$ are obtained, the pseudo high band sub-band power difference calculation circuit 36 calculates the following Equation (23) and calculates the average value $\text{ResP}(\text{id}, J)$.

25

[0368]

[Equation 23]

$$\begin{aligned} \text{ResP}(\text{id}, \text{J}) = & \text{ResP}_{\text{std}}(\text{id}, \text{J}) + W_{\text{max}} \times \text{ResP}_{\text{max}}(\text{id}, \text{J}) \\ & + W_{\text{ave}} \times \text{ResP}_{\text{ave}}(\text{id}, \text{J}) \quad \cdot \cdot \cdot (23) \end{aligned}$$

[0369]

That is, the estimation residual square value
 5 $\text{ResP}_{\text{std}}(\text{id}, \text{J})$, the estimation residual error maximum value
 $\text{ResP}_{\text{max}}(\text{id}, \text{J})$ and the estimation residual average value
 $\text{ResP}_{\text{ave}}(\text{id}, \text{J})$ are added with weight and set as the estimation
 value $\text{ResP}(\text{id}, \text{J})$. In addition, in Equation (23), W_{max} and
 W_{ave} are a predetermined weight, for example,
 10 $W_{\text{max}}=0.5$, $W_{\text{ave}}=0.5$.

[0370]

Therefore, if the evaluation value $\text{ResP}(\text{id}, \text{J})$ using the
 past frame and the current value is calculated, the process
 proceeds from the step S337 to S338.

15 [0371]

In step S338, the pseudo high band sub-band power
 difference calculation circuit 36 calculates the Equation (24)
 and calculates the ultimate estimation value $\text{Res}_{\text{all}}(\text{id}, \text{J})$.

[0372]

20 [Equation 24]

$$\text{Res}_{\text{all}}(\text{id}, \text{J}) = \text{Res}(\text{id}, \text{J}) + W_p(\text{J}) \times \text{ResP}(\text{id}, \text{J}) \quad \cdot \cdot \cdot (24)$$

[0373]

That is, the obtained estimation value $\text{Res}(\text{id}, \text{J})$ and
 the estimation value $\text{ResP}(\text{id}, \text{J})$ are added with weight. In
 25 addition, in the Equation (24), $W_p(\text{J})$, for example, is a weight
 defined by the following Equation (25).

[0374]

[Equation 25]

$$W_p(J) = \begin{cases} \frac{-\text{power}_r(J)}{50} + 1 & (0 \leq \text{power}_r(J) \leq 50) \\ 0 & (\text{otherwise}) \end{cases} \quad \dots (25)$$

[0375]

In addition, $\text{power}_r(J)$ in the Equation (25) is a value
5 defined by the following Equation (26).

[0376]

[Equation 26]

$$\text{power}_r(J) = \sqrt{\left(\sum_{ib=sb+1}^{eb} [\text{power}(ib, J) - \text{power}(ib, J-1)]^2 \right) / (eb - sb)} \quad \dots (26)$$

[0377]

10 This $\text{power}_r(J)$ shows the average of the difference
between the high band sub-band powers of frames (J-1) and frames
J. In addition, according to the Equation (25), when $\text{power}_r(J)$
is a value of the predetermined range in the vicinity of 0,
the smaller the $\text{power}_r(J)$, $W_p(J)$ is closer to 1 and when
15 $\text{power}_r(J)$ is larger than a predetermined range value, it is
set as 0.

[0378]

Herein, when $\text{power}_r(J)$ is a value of a predetermined
range in the vicinity of 0, the average of the difference of
20 the high band sub-band power between the consecutive frames
becomes small to a degree. That is, the time variation of
the high band component of the input signal is small and the
current frames of the input signal become steady region.

[0379]

As the high band component of the input signal is steady, the weight $W_p(J)$ becomes a value is close to 1, whereas as the high band component is not steady, the weight ($W_p(J)$) becomes a value close to 0. Therefore, in the estimation value

5 $Res_{all}(id, J)$ shown in Equation (24), as the time variety of the high band component of the input signal becomes small, the coefficient of determination of the estimation value $ResP(id, J)$ considering the comparison result and the estimation result of the high band component as the evaluation standards

10 in the previous frames become larger.

[0380]

Therefore, in a steady region of the input signal, the decoded high band sub-band power estimation coefficient obtained in the vicinity of the estimation result of the high

15 band component in previous frames is selected and in the decoder 40 side, it is possible to more naturally reproduce the sound having high quality. Whereas in a non-steady region of the input signal, a term of estimation value $ResP(id, J)$ in the estimation value $Res_{all}(id, J)$ is set as 0 and the decoded high

20 band signal closed to the actual high band signal is obtained.

[0381]

The pseudo high band sub-band power difference calculation circuit 36 calculates the estimation value $Res_{all}(id, J)$ for each of the K number of the decoded high band

25 sub-band power evaluation coefficient by performing the above-mentioned processes.

[0382]

In step S339, the pseudo high band sub-band power difference calculation circuit 36 selects the coefficient

30 index id based on the estimation value $Res_{all}(id, J)$ for each obtained decoded high band sub-band power estimation

coefficient.

[0383]

The estimation value $\text{Res}_{a11}(\text{id}, J)$ obtained from the process described above linearly combines the estimation value
 5 $\text{Res}(\text{id}, J)$ and the estimation value $\text{ResP}(\text{id}, J)$ using weight. As described above, the smaller the estimation value $\text{Res}(\text{id}, J)$, a decoded high band signal closer to an actual high band signal can be obtained. In addition, the smaller the estimation value $\text{ResP}(\text{id}, J)$, a decoded high band signal closer to the decoded
 10 high band signal of the previous frame can be obtained.

[0384]

Therefore, the smaller the estimation value $\text{Res}_{a11}(\text{id}, J)$, a more suitable decoded high band signal is obtained. Therefore, the pseudo high band sub-band power difference
 15 calculation circuit 36 selects the estimation value having a minimum value in the K number of the estimation $\text{Res}_{a11}(\text{id}, J)$ and supplies the coefficient index indicating the decoded high band sub-band power estimation coefficient corresponding to this estimation value to the high band encoding circuit 37.

20 [0385]

If the coefficient index is selected, after that, the processes of step S340 and step S341 are performed to complete the encoding process. However, since these processes are the same as the processes of step S308 and step S309 in Fig. 24,
 25 the description thereof will be omitted.

[0386]

As described above, in the encoder 30, the estimation value $\text{Res}_{a11}(\text{id}, J)$ obtained by linearly combining the estimation value $\text{Res}(\text{id}, J)$ and the estimation value $\text{ResP}(\text{id}, J)$
 30 is used, so that the coefficient index of the optimal decoded high band sub-band power estimation coefficient is selected.

[0387]

If the estimation value $\text{Res}_{\text{all}}(\text{id}, \text{J})$ is used, as the case uses the estimation value $\text{Res}(\text{id}, \text{J})$, it is possible to select a more suitable decoded high band sub-band power estimation coefficient by more many estimation standards. However, if
5 the estimation value $\text{Res}_{\text{all}}(\text{id}, \text{J})$ is used, it is possible to control the time variation in the steady region of the high band component of signal to be reproduced in the decoder 40 and it is possible to obtain a signal having high quality.

10 [0388]

<Modification Example 2>

By the way, in the frequency band expansion process, if the sound having high quality is desired to be obtained, the sub-band of the lower band side is also important in term
15 of the audibility. That is, among sub-bands on the high band side as the estimation accuracy of the sub-band close to the low band side become larger, it is possible to reproduce sound having high quality.

[0389]

20 Herein, when the estimation value with respect to each decoded high band sub-band power estimation coefficient is calculated, a weight may be placed on the sub-band of the low band side. In this case, the encoder 30 in Fig. 18 performs the encoding process shown in the flowchart in Fig. 26.

25 [0390]

Hereinafter, the encoding process by the encoder 30 will be described with reference to the flowchart in Fig. 26. In addition, the processes of steps S371 to step S375 are identical with those of step S331 to step S335 in Fig. 25. Therefore,
30 the description thereof will be omitted.

[0391]

In step S376, the pseudo high band sub-band power difference calculation circuit 36 calculates estimation value $Res_{W_{band}}(id, J)$ using the current frame J to be processed for each of the K number of decoded high band sub-band power estimation coefficient.

[0392]

Specially, the pseudo high band sub-band power difference calculation circuit 36 calculates high band sub-band power $power(ib, J)$ in the frames J performing the same operation as the above-mentioned Equation (1) using the high band sub-band signal of each sub-band supplied from the sub-band division circuit 33.

[0393]

If the high band sub-band power $power(ib, J)$ is obtained, the pseudo high band sub-band power difference calculation circuit 36 calculates the following Equation 27 and calculates the residual square average value $Res_{std W_{band}}(id, J)$.

[0394]

[Equation 27]

$$Res_{std W_{band}}(ib, J) = \sum_{ib=sb+1}^{eb} \{W_{band}(ib) \times [power(ib, J) - power_{est}(ib, id, J)]\}^2 \quad \dots (27)$$

[0395]

That is, the difference between the high band sub-band power $power(ib, J)$ of the frames (J) and the pseudo high band sub-band power ($power_{est}(ib, id, J)$) is obtained and the difference is multiplied by the weight $W_{band}(ib)$ for each sub-band, for each sub-band on the high band side where the index is $sb+1$ to eb . In addition, the sum of square for difference by which the weight $W_{band}(ib)$ is multiplied is set

as the residual error square average value $\text{Res}_{\text{std}W_{\text{band}}}(id, J)$.
[0396]

Herein, the weight $W_{\text{band}}(ib)$ (where, $sb+1 \leq ib \leq eb$ is defined by the following Equation 28. For example, the value
5 of the weight $W_{\text{band}}(ib)$ becomes as large as the sub-band of the low band side.

[0397]

[Equation 28]

$$W_{\text{band}}(ib) = \frac{-3 \times ib}{7} + 4 \quad \cdot \cdot \cdot (28)$$

10 [0398]

Next, the pseudo high band sub-band power difference calculation circuit 36 calculates the residual maximum value $\text{Res}_{\text{max}W_{\text{band}}}(id, J)$. Specifically, the maximum value of the absolute value of the values multiplying the difference between
15 the high band sub-band power $\text{power}(ib, J)$ of each sub-band where the index is $sb+1$ to eb and the pseudo high band sub-band power $\text{power}_{\text{est}}(ib, id, J)$ by the weight $W_{\text{band}}(ib)$ is set as the residual error difference maximum value $\text{Res}_{\text{max}W_{\text{band}}}(id, J)$.

[0399]

20 In addition, the pseudo high band sub-band power difference calculation circuit 36 calculates the residual error average value $\text{Res}_{\text{ave}W_{\text{band}}}(id, J)$.

[0400]

Specially, in each sub-band where the index is $sb + 1$
25 to eb , the difference between the high band sub-band power $\text{power}(ib, J)$ and the pseudo high band sub-band power $\text{power}_{\text{est}}(ib, id, J)$ is obtained and thus weight $W_{\text{band}}(ib)$ is multiplied so that the sum total of the difference by which the weight $W_{\text{band}}(ib)$ is multiplied, is obtained. In addition,
30 the absolute value of the value obtained by dividing the

obtained sum total of the difference into the sub-band number (eb - sb) of the high band side is set as the residual error average value $\text{Res}_{\text{ave}W_{\text{band}}}(\text{id}, J)$.

[0401]

5 In addition, the pseudo high band sub-band power difference calculation circuit 36 calculates the evaluation value $\text{Res}W_{\text{band}}(\text{id}, J)$. That is, the sum of the residual squares mean value $\text{Res}_{\text{std}W_{\text{band}}}(\text{id}, J)$, the residual error maximum value $\text{Res}_{\text{max}W_{\text{band}}}(\text{id}, J)$ that the weight (W_{max}) is multiplied, and the
10 residual error average value $\text{Res}_{\text{ave}W_{\text{band}}}(\text{id}, J)$ by which the weight (W_{ave}) is multiplied, is set as the average value $\text{Res}W_{\text{band}}(\text{id}, J)$.

[0402]

 In step S377, the pseudo high band sub-band power
15 difference calculation circuit 36 calculates the average value $\text{Res}PW_{\text{band}}(\text{id}, J)$ using the past frames and the current frames.

[0403]

 Specially, the pseudo high band sub-band power difference calculation circuit 36 records the pseudo high band
20 sub-band power of each sub-band obtained by using the decoded high band sub-band power estimation coefficient of the coefficient index selected finally with respect to the frames J-1 before one frame earlier than the frame (J) to be processed in time.

25 [0404]

 The pseudo high band sub-band power difference calculation circuit 36 first calculates the estimation
residual error average value $\text{Res}P_{\text{std}W_{\text{band}}}(\text{id}, J)$. That is, for each sub-band on the high band side in which the index
30 is sb+1 to eb, the weight $W_{\text{band}}(\text{id})$ is multiplied by obtaining the difference between the pseudo high band sub-band power

power_{est}(ib,id_{selected}(J-1),J-1) and the pseudo high band sub-band power power_{est}(ib,id,J). In addition, the squared sum of the difference from which the weigh W_{band}(ib) is calculated, is set as the estimation error difference average value ResP_{std}W_{band}(id,J).
[0405]

The pseudo high band sub-band power difference calculation circuit 36 continuously calculates the estimation residual error maximum value ResP_{max}W_{band}(id,J). Specially, the maximum value of the absolute value obtained by multiplying the difference between the pseudo high band sub-band power power_{est}(ib,id_{selected}(J-1),J-1) of each sub-band in which the index is sb+1 to eb and the pseudo high band sub-band power -power_{est}(ib,id,J) by the weight W_{band}(ib) is set as the estimation residual error maximum value ResP_{max}W_{band}(id,J).
[0406]

Next, the pseudo high band sub-band power difference calculation circuit 36 calculates the estimation residual error average value ResP_{ave}W_{band}(id,J). Specially, the difference between the pseudo high band sub-band power_{est}(ib,id_{selected}(J-1),J-1) and the pseudo high band sub-band power power_{est}(ib,id,J) is obtained for each sub-band where the index is sb+1 to eb and the weight W_{band}(ib) is multiplied. In addition, the sum total of the difference by which the weight W_{band}(ib) is multiplied is the absolute value of the values obtained by being divided into the number (eb-sb) of the sub-bands of the high band side. However, it is set as the estimation residual error average value ResP_{ave}W_{band}(id,J).

[0407]

Further, the pseudo high band sub-band power difference

calculation circuit 36 obtains the sum of the estimation residual error square average value $ResP_{std} W_{band}(id, J)$ of the estimation residual error maximum value $ResP_{max} W_{band}(id, J)$ by which the weight W_{max} is multiplied and the estimation residual error average value $ResP_{ave} W_{band}(id, J)$ by which the weight W_{ave} is multiplied and the sum is set as the estimation value $ResPW_{band}(id, J)$.

[0408]

In step S378, the pseudo high band sub-band power difference calculation circuit 36 adds the evaluation value $ResW_{band}(id, J)$ to the estimation value $ResPW_{band}(id, J)$ by which the weight $W_p(J)$ of the Equation (25) is multiplied to calculate the final estimation value $Res_{all} W_{band}(id, J)$. This estimation value $Res_{all} W_{band}(id, J)$ is calculated for each of the K number decoded high band sub-band power estimation coefficient.

[0409]

In addition, after that, the processes of step S379 to step S381 are performed to terminate the encoding process. However, since their processes are identical to those of with step S339 to step S341 in Fig. 25, the description thereof is omitted. In addition, the estimation value $Res_{all} W_{band}(id, J)$ is selected to be a minimum in the K number of coefficient index in step S379.

25 [0410]

As described above, in order to place the weight in the sub-band of the low band side, it is possible to obtain sound having further high quality in the decoder 40 side by providing the weight for each of the sub-band.

30 [0411]

In addition, as described above, the selection of the

number of the decoded high band sub-band power estimation coefficient has been described as being performed based on the estimation value $\text{Res}_{\text{allW}_{\text{band}}}(\text{id}, \text{J})$. However, the decoded high band sub-band power evaluation coefficient may be selected
 5 based on the estimation value $\text{ResW}_{\text{band}}(\text{id}, \text{J})$.

[0412]

<Modification Example 3>

In addition, since the auditory of person has a property that properly perceives a larger frequency band of the
 10 amplitude (power), the estimation value with respect to each decoded high band sub-band power estimation coefficient may be calculated so that the weight may be placed on the sub-band having a larger power.

[0413]

15 In this case, the encoder 30 in Fig. 18 performs an encoding process illustrated in a flowchart in Fig. 27. The encoding process by the encoder 30 will be described below with reference to the flowchart in Fig. 27. In addition, since the processes of step S401 to step S405 are identical with
 20 those of step S331 to step S335 in Fig. 25, the description thereof will be omitted.

[0414]

In step S406, the pseudo high band sub-band power difference calculation circuit 36 calculates the estimation
 25 value $\text{ResW}_{\text{power}}(\text{id}, \text{J})$ using the current frame J to be processed for the K number of decoded high band sub-band power estimation coefficient.

[0415]

Specifically, the pseudo high band sub-band power
 30 difference calculation circuit 36 calculates the high band sub-band power $\text{power}(\text{id}, \text{J})$ in the frames J by performing the

same operation as the Equation (1) described above by using a high band sub-band signal of each sub-band supplied from the sub-band division circuit 33.

[0416]

5 If the high band sub-band power $\text{power}(\text{ib}, J)$ is obtained, the pseudo high band sub-band power difference calculation circuit 36 calculates the following Equation (29) and calculates the residual error squares average value $\text{Res}_{\text{std}} W_{\text{power}}(\text{id}, J)$.

10 [0417]

[Equation 29]

$$\begin{aligned} \text{Res}_{\text{std}} W_{\text{power}}(\text{id}, J) = & \sum_{\text{ib}=\text{sb}+1}^{\text{eb}} \{W_{\text{power}}(\text{power}(\text{ib}, J)) \\ & \times [\text{power}(\text{ib}, J) - \text{power}_{\text{est}}(\text{ib}, \text{id}, J)]\}^2 \\ & \dots (29) \end{aligned}$$

[0418]

That is, the difference between the high band sub-band power $\text{power}_{\text{est}}(\text{ib}, J)$ and the pseudo high band sub-band power $\text{power}_s(\text{ib}, \text{id}, J)$ is obtained and the weight $W_{\text{power}}(\text{power}(\text{ib}, J))$ for each of the sub-bands is multiplied by the difference thereof with respect to each band of the high band side in which the index is $\text{sb}+1$ to eb . In addition, the square sum of the difference by which the weight $W_{\text{power}}(\text{power}(\text{ib}, J))$ is multiplied by set as the residual error squares average value $\text{Res}_{\text{std}} W_{\text{power}}(\text{id}, J)$.

[0419]

Herein, the weight $W_{\text{power}}(\text{power}(\text{ib}, J))$ (where, $\text{sb}+1 \leq \text{ib} \leq \text{eb}$), for example, is defined as the following Equation (30). As the high band sub-band power $\text{power}(\text{ib}, J)$ of the sub-band becomes large, the value of weight $W_{\text{power}}(\text{power}(\text{ib}, J))$ becomes larger.

[0420]

[Equation 30]

$$W_{\text{power}}(\text{power}(\text{ib}, J)) = \frac{3 \times \text{power}(\text{ib}, J)}{80} + \frac{35}{8} \quad \cdot \cdot \cdot (30)$$

[0421]

5 Next, the pseudo high band sub-band power difference calculation circuit 36 calculates the residual error maximum value $\text{Res}_{\text{max}} W_{\text{power}}(\text{id}, J)$. Specially, the maximum value of the absolute value multiplying the difference between the high band sub-band power $\text{power}(\text{ib}, J)$ of the each sub-band that the index is sb+1 to eb and the pseudo high band sub-band power $\text{power}_{\text{est}}(\text{ib}, \text{id}, J)$ by the weight $W_{\text{power}}(\text{power}(\text{ib}, J))$ is set as the residual error maximum value $\text{Res}_{\text{max}} W_{\text{power}}(\text{id}, J)$.

[0422]

15 In addition, the pseudo high band sub-band power difference calculation circuit 36 calculates the residual error average value $\text{Res}_{\text{ave}} W_{\text{power}}(\text{id}, J)$.

[0423]

20 Specially, in each sub-band where the index is sb+1 to eb, the difference between the high band sub-band power $\text{power}(\text{ib}, J)$ and the pseudo high band sub-band power $\text{power}_{\text{est}}(\text{ib}, \text{id}, J)$ is obtained and the weight by which $W_{\text{power}}(\text{power}(\text{ib}, J))$ is multiplied and the sum total of the difference that the weight $W_{\text{power}}(\text{power}(\text{ib}, J))$ is multiplied is obtained. In addition, the absolute value of the values obtained by dividing the sum total of the obtained difference into the number of the high band sub-band and eb-sb) is set as the residual error average $\text{Res}_{\text{ave}} W_{\text{power}}(\text{id}, J)$.

[0424]

30 Further, the pseudo high band sub-band power difference calculation circuit 36 calculates the estimation value

Res $W_{power}(id, J)$. That is, the sum of residual squares average value Res $_{std}W_{power}(id, J)$, the residual error difference value Res $_{max}W_{power}(id, J)$ by which the weight (W_{max}) is multiplied and the residual error average value Res $_{ave}W_{power}(id, J)$ by which
 5 the weight (W_{ave}) is multiplied, is set as the estimation value Res $W_{power}(id, J)$.

[0425]

In step S407, the pseudo high band sub-band power difference calculation circuit 36 calculates the estimation
 10 value Res $PW_{power}(id, J)$ using the past frame and the current frames.

[0426]

Specifically, the pseudo high band sub-band power difference calculation circuit 36 records the pseudo high band
 15 sub-band power of each sub-band obtained by using the decoded high band sub-band power estimation coefficient of the coefficient index selected finally with respect to the frames($J-1$) before one frame earlier than the frame J to be processed in time.

20 [0427]

The pseudo high band sub-band power difference calculation circuit 36 first calculates the estimation residual square average value Res $P_{std}W_{power}(id, J)$. That is, the difference between the pseudo high band sub-band power
 25 power $_{est}(ib, idJ)$ and the pseudo high band sub-band power (power $_{est}(ib, id_{selected}(J-1), J-1)$ is obtained to multiply the weight $W_{power}(power(ib, J)$, with respect to each sub-band the high-band side in which the index is $sb+1$ and eb . The square sum of the difference that the weight $W_{power}(power(ib, J)$ is
 30 multiplied is set as the estimation residual square average value Res $P_{std}W_{power}(id, J)$.

[0428]

Next, the pseudo high band sub-band power difference calculation circuit 36 calculates the estimation residual error maximum value $\text{ResP}_{\text{max}}W_{\text{power}}(\text{id}, J)$. Specifically, the absolute value of the maximum value of the values multiplying the difference between the pseudo high band sub-band power $\text{power}_{\text{est}}(\text{ib}, \text{id}_{\text{selected}}(J-1), J-1)$ of each sub-band in which the index is sb+1 to as eb and the pseudo high band sub-band power $\text{power}_{\text{est}}(\text{ib}, \text{id}, J)$ by the weight $W_{\text{power}}(\text{power}(\text{ib}, J))$ is set as the estimation residual error maximum value $\text{ResP}_{\text{max}}W_{\text{power}}(\text{id}, J)$.

[0429]

Next, the pseudo high band sub-band power difference calculation circuit 36 calculates the estimation residual error average value $\text{ResP}_{\text{ave}}W_{\text{power}}(\text{id}, J)$. Specifically, the difference between the pseudo high band sub-band power $\text{power}_{\text{est}}(\text{ib}, \text{id}_{\text{selected}}(J-1), J-1)$ and the pseudo high band sub-band power $\text{power}_{\text{est}}(\text{ib}, \text{id}, J)$ is obtained with respect to each sub-band in which the index is sb+1 to eb and the weight $W_{\text{power}}(\text{power}(\text{ib}, J))$ is multiplied. In addition, the absolute values of the values obtained by dividing the sum total of the multiplied difference of the weight $W_{\text{power}}(\text{power}(\text{ib}, J))$ into the number (eb-sb) of the sub-band of high band side is set as the estimation residual error average value $\text{ResP}_{\text{ave}}W_{\text{power}}(\text{id}, J)$.

[0430]

Further, the pseudo high band sub-band power difference calculation circuit 36 obtains the sum of the estimation residual squares mean value $\text{ResP}_{\text{std}}W_{\text{power}}(\text{id}, J)$, the estimation residual error maximum value $\text{ResP}_{\text{max}}W_{\text{power}}(\text{id}, J)$ by which the weight (W_{max}) is multiplied and the estimation residual

error average value $\text{ResP}_{\text{aveWpower}}(\text{id}, \text{J})$ that the weight (W_{ave}) is multiplied is obtained and the sum is set as the estimation value $\text{ResPW}_{\text{power}}(\text{id}, \text{J})$.

[0431]

5 In step S408, the pseudo high band sub-band power difference calculation circuit 36 adds the estimation value $\text{ResW}_{\text{power}}(\text{id}, \text{J})$ to the estimation value $\text{ResPW}_{\text{power}}(\text{id}, \text{J})$ by which the weight $W_p(\text{J})$ of the Equation (25) is multiplied to calculate the final estimation value $\text{Res}_{\text{allWpower}}(\text{id}, \text{J})$.
 10 The estimation value $\text{Res}_{\text{allWpower}}(\text{id}, \text{J})$ is calculated from each K number of the decoded high band sub-band power estimation coefficient.

[0432]

15 In addition, after that, the processes of step S409 to step S411 are performed to terminate the encoding process. However, since these processes are identical with those of step S339 to step S341 in Fig. 25, the description thereof is omitted. In addition, in step S409, the coefficient index in which the estimation value $\text{Res}_{\text{allWpower}}(\text{id}, \text{J})$ is set as
 20 a minimum is selected among the K number of the coefficient index.

[0433]

25 As described above, in order for weight to be placed on the sub-band having a large sub-band, it is possible to obtain sound having high quality by providing the weight for each sub-band in the decoder 40 side.

[0434]

30 In addition, as described above, the selection of the decoded high band sub-band power estimation coefficient has been described as being performed based on the estimation value $\text{Res}_{\text{allWpower}}(\text{id}, \text{J})$. However, the decoded high band sub-band

power estimation coefficient may be selected based on the estimation value $\text{ResW}_{\text{power}}(\text{id}, \text{J})$.

[0435]

<6. Sixth embodiment>

5 [Configuration of Coefficient Learning Apparatus]

By the way, a set of a coefficient $A_{\text{lb}}(\text{kb})$ as the decoded high band sub-band power estimation coefficient and a coefficient B_{lb} is recorded in a decoder 40 in Fig. 20 to correspond to the coefficient index. For example, if the
10 decoded high band sub-band power estimation coefficient of 128 coefficient index is recorded in decoder 40, a large area is needed as the recording area such as memory for recording the decoded high band sub-band power estimation coefficient thereof.

15 [0436]

Herein, a portion of a number of the decoded high band sub-band power estimation coefficient is set as common coefficient and the recording area necessary to record the decoded high band sub-band power estimation coefficient may
20 be made smaller. In this case, the coefficient learning apparatus obtained by learning the decoded high band sub-band power estimation coefficient, for example, is configured as illustrated in Fig. 28.

[0437]

25 The coefficient learning apparatus 81 includes a sub-band division circuit 91, a high band sub-band power calculation circuit 92, a characteristic amount calculation circuit 93 and a coefficient estimation circuit 94.

[0438]

30 A plurality of composition data using learning is provided in a plurality of the coefficient learning apparatus

81 as a broadband instruction signal. The broadband instruction signal is a signal including a plurality of sub-band component of the high band and a plurality of the sub-band components of the low band.

5 [0439]

The sub-band division circuit 91 includes the band pass filter and the like, divides the supplied broadband instruction signal into a plurality of the sub-band signals and supplies to the signals the high band sub-band power calculation circuit 10 92 and the characteristic amount calculation circuit 93. Specifically, the high band sub-band signal of each sub-band of the high band side in which the index is $sb+1$ to eb is supplied to the high band sub-band power calculation circuit 92 and the low band sub-band signal of each sub-band of the low band 15 in which the index is $sb-3$ to sb is supplied to the characteristic amount calculation circuit 93.

[0440]

The high band sub-band power calculation circuit 92 calculates the high band sub-band power of each high band 20 sub-band signal supplied from the sub-band division circuit 91 and supplies it to the coefficient estimation circuit 94. The characteristic amount calculation circuit 93 calculates the low band sub-band power as the characteristic amount, the low band sub-band power based on each low band sub-band signal 25 supplied from the sub-band division circuit 91 and supplies it to the coefficient estimation circuit 94.

[0441]

The coefficient estimation circuit 94 produces the decoded high band sub-band power estimation coefficient by 30 performing a regression analysis using the high band sub-band power from the high band sub-band power calculation circuit

92 and the characteristic amount from the characteristic amount calculation circuit 93 and outputs to decoder 40.

[0442]

[Description of Coefficient Learning Process]

5 Next, a coefficient learning process performed by a coefficient learning apparatus 81 will be described with reference to a flowchart in Fig. 29.

[0443]

10 In step S431, the sub-band division circuit 91 divides each of a plurality of the supplied broadband instruction signal into a plurality of sub-band signals. In addition, the sub-band division circuit 91 supplies a high band sub-band signal of the sub-band that the index is sb+1 to eb to the high band sub-band power calculation circuit 92 and supplies
15 the low band sub-band signal of the sub-band that the index is sb-3 to sb to the characteristic amount calculation circuit 93.

[0444]

20 In step S432, the high band sub-band power calculation circuit 92 calculates the high band sub-band power by performing the same operation as the Equation (1) described above with respect to each high band sub-band signal supplied from the sub-band division circuit 91 and supplies it to the coefficient estimation circuit 94.

25 [0445]

30 In step S433, the characteristic amount calculation circuit 93 calculates the low band sub-band power as the characteristic amount by performing the operation of the Equation (1) described above with respect each low band sub-band signal supplied from the sub-band division circuit 91 and supplies to it the coefficient estimation circuit 94.

[0446]

Accordingly, the high band sub-band power and the low band sub-band power are supplied to the coefficient estimation circuit 94 with respect to each frame of a plurality of the
5 broadband instruction signal.

[0447]

In step S434, the coefficient estimation circuit 94 calculates a coefficient $A_{ib}(kb)$ and a coefficient B_{ib} by performing the regression of analysis using least-squares
10 method for each of the sub-band ib (where, $sb+1 \leq ib \leq eb$) of the high band in which the index is $sb+1$ to eb .

[0448]

In the regression analysis, it is assumed that the low band sub-band power supplied from the characteristic amount
15 calculation circuit 93 is an explanatory variable and the high band sub-band power supplied from the high band sub-band power calculation circuit 92 is an explained variable. In addition, the regression analysis is performed by using the low band sub-band power and the high band sub-band power of the whole
20 frames constituting the whole broadband instruction signal supplied to the coefficient learning apparatus 81.

[0449]

In step S435, the coefficient estimation circuit 94 obtains the residual vector of each frame of the broadband
25 instruction signal using a coefficient $A_{ib}(kb)$ and a coefficient (B_{ib}) for each of obtained sub-band ib .

[0450]

For example, the coefficient estimation circuit 94 obtains the residual error by subtracting the sum of total
30 of the lower band sub-band power $power(kb, J)$ (where, $sb-3 \leq kb \leq sb$) that is acquired by the coefficient $A_{ib}A_{ib}(kb)$

thereto coefficient B_{ib} multiplied from the high band power
 ((power(ib, J) for each of the sub-band ib (where, $sb+1 \leq ib \leq eb$) of
 the frame J and. In addition, vector including the residual
 error of each sub-band ib of the frame J is set as the residual
 5 vector.

[0451]

In addition, the residual vector is calculated with
 respect to the frame constituting the broadband instruction
 signal supplied to the coefficient learning apparatus 81.

10 [0452]

In step S436, the coefficient estimation circuit 94
 normalizes the residual vector obtained with respect to each
 frame. For example, the coefficient estimation circuit 94
 normalizes, for each sub-band ib, the residual vector by
 15 obtaining variance of the residual of the sub-band ib of the
 residual vector of the whole frame and dividing a residual
 error of the sub-band ib in each residual vector into the square
 root of the variance.

[0453]

20 In step S437, the coefficient estimation circuit 94
 clusters the residual vector of the whole normalized frame
 by the k-means method or the like.

[0454]

For example, the average frequency envelope of the whole
 25 frame obtained when performing the estimation of the high band
 sub-band power using the coefficient $A_{ib}(kb)$ and the
 coefficient B_{ib} is referred to as an average frequency envelope
 SA. In addition, it is assumed that a predetermined frequency
 envelope having larger power than the average frequency
 30 envelope SA is frequency envelope SH and a predetermined
 frequency envelope having smaller power than the average

frequency envelope SA is frequency envelope SL.

[0455]

In this case, each residual vector of the coefficient in which the frequency envelope close to the average frequency envelop SA, the frequency envelop SH and the frequency envelop SL is obtained, performs clustering of the residual vector so as to be included in a cluster CA, a cluster CH, and a cluster CL. That is, the residual vector of each frame performs clustering so as to be included in any one of cluster CA, a cluster CH or a cluster CL.

[0456]

In the frequency band expansion process for estimating the high band component based on a correlation of the low band component and the high band component, in terms of this, if the residual vector is calculated using the coefficient A_{ib} (kb) and the coefficient B_{ib} obtained from the regression analysis, the residual error increases as much as large as the sub-band of the high band side. Therefore, the residual vector is clustered without changing, the weight is placed in as much as sub-band of the high band side to perform process.

[0457]

In this contrast, in the coefficient learning apparatus 81, variance of the residual error of each sub-band is apparently equal by normalizing the residual vector as the variance of the residual error of the sub-band and clustering can be performed by providing the equal weight to each sub-band.

[0458]

In step S438, the coefficient estimation circuit 94 selects as a cluster to be processed of any one of the cluster CA, the cluster CH and the cluster CL.

[0459]

In step S439, the coefficient estimation circuit 94 calculates $A_{ib}(kb)$ and the coefficient B_{ib} of each sub-band ib (where, $sb+1 \leq ib \leq eb$) by the regression analysis using the frames of the residual vector included in the cluster selected
 5 as the cluster to be processed.

[0460]

That is, if the frame of the residual vector included in the cluster to be processed is referred to as the frame to be processed, the low band sub-band power and the high band
 10 sub-band power of the whole frame to be processed is set as the exploratory variable and the explained variable and the regression analysis used the least-squares method is performed. Accordingly, the coefficient $A_{ib}(kb)$ and the coefficient B_{ib} is obtained for each sub-band ib .

15 [0461]

In step S440, the coefficient estimation circuit 94 obtains the residual vector using the coefficient $A_{ib}(kb)$ and the coefficient B_{ib} obtained by the process of step S439 with respect the whole frame to be processed. In addition, in step
 20 S440, the same process as the step S435 is performed and thus the residual vector of each frame to be processed is obtained.

[0462]

In step S441, the coefficient estimation circuit 94 normalizes the residual vector of each frame to be processed
 25 obtained by process of step S440 by performing the same process as step S436. That is, normalization of the residual vector is performed by dividing the residual error by the variance for each the sub-band.

[0463]

30 In step S442, the coefficient estimation circuit 94 clusters the residual vector of the whole normalized frame

to be processed using k-means method or the like. The number of this cluster number is defined as following. For example, in the coefficient learning apparatus 81, when decoded high band sub-band power estimation coefficients of 128 coefficient indices are produced, 128 is multiplied by the frame number to be processed and the number obtained by dividing the whole frame number is set as the cluster number. Herein, the whole frame number is referred to as sum of the whole frame of the broadband instruction signal supplied to the coefficient learning apparatus 81.

[0464]

In step S443, the coefficient estimation circuit 94 obtains a center of gravity vector of each cluster obtained by process of step S442.

15 [0465]

For example, the cluster obtained by the clustering of the step S442 corresponds to the coefficient index and in the coefficient learning apparatus 81, the coefficient index is assigned for each cluster to obtain the decoded high band sub-band power estimation coefficient of the each coefficient index.

[0466]

Specifically, in step S438, it is assumed that the cluster CA is selected as a cluster to be processed and F clusters are obtained by clustering in step S442. When one cluster CF of F clusters is focused, the decoded high band sub-band power estimation coefficient of a coefficient index of the cluster CF is set as the coefficient $A_{ib}(kb)$ in which the coefficient $A_{ib}(kb)$ obtained with respect to the cluster CA in step S439 is a linear correlative term. In addition, the sum of the vector performing a reverse process (reverse

normalization) of a normalization performed at step S441 with respect to center of gravity vector of the cluster CF obtained from step S443 and the coefficient B_{ib} obtained at step S439 is set as the coefficient B_{ib} which is a constant term of the
5 decoded high band sub-band power estimation coefficient. The reverse normalization is set as the process multiplying the same value (root square for each sub-band) as when being normalized with respect to each element of center of gravity vector of the cluster CF when the normalization, for example,
10 performed at step S441 divides the residual error into the root square of the variance for each sub-band.

[0467]

That is, the set of the coefficient $A_{ib}(kb)$ obtained at step S439 and the coefficient B_{ib} obtained as described
15 is set as the decoded high band sub-band power estimation coefficient of the coefficient index of the cluster CF. Accordingly, each of the F clusters obtained by clustering commonly has the coefficient $A_{ib}(kb)$ obtained with respect to the cluster CA as the linear correlation term of the decoded
20 high band sub-band power estimation coefficient.

[0468]

In step S444, the coefficient learning apparatus 81 determines whether the whole cluster of the cluster CA, the cluster CH and the cluster CL is processed as a cluster to
25 be processed. In addition, in step S444, if it is determined that the whole cluster is not processed, the process returns to step S438 and the process described is repeated. That is, the next cluster is selected to be processed and the decoded highband sub-band power estimation coefficient is calculated.

30 [0469]

In this contrast, in step S444, if it is determined that

the whole cluster is processed, since a predetermined number of the decoded high band sub-band power to be obtained is calculated, the process proceeds to step S445.

[0470]

5 In step S445, the coefficient estimation circuit 94 outputs and the obtained coefficient index and the decoded high band sub-band power estimation coefficient to decoder 40 and thus the coefficient learning process is terminated.

[0471]

10 For example, in the decoded high band sub-band power estimation coefficients output to decoder 40, there are several same coefficients $A_{ib}(kb)$ as linear correlation term. Herein, the coefficient learning apparatus 81 corresponds to the linear correlation term index (pointer) which is information that
15 specifies the coefficient $A_{ib}(kb)$ to the coefficient $A_{ib}(kb)$ common to thereof and corresponds the coefficient B_{ib} which is the linear correlation index and the constant term to the coefficient index.

[0472]

20 In addition, the coefficient learning apparatus 81 supplies the corresponding linear correlation term index (pointer) and a coefficient $A_{ib}(kb)$, and the corresponding coefficient index and the linear correlation index (pointer) and the coefficient B_{ib} to the decoder 40 and records them in
25 a memory in the high band decoding circuit 45 of the decoder 40. Like this, when a plurality of the decoded high band sub-band power estimation coefficients are recorded, if the linear correlation term index (pointer) is stored in the recording area for each decoded high band sub-band power
30 estimation coefficient with respect to the common linear correlation term, it is possible to reduce the recording area

remarkably.

[0473]

In this case, since the linear correlation term index and to the coefficient $A_{ib}(kb)$ are recorded in the memory in the high band decoding circuit 45 to correspond to each other, the linear correlation term index and the coefficient B_{ib} are obtained from the coefficient index and thus it is possible to obtain the coefficient $A_{ib}(kb)$ from the linear correlation term index.

10 [0474]

In addition, according to a result of analysis by the applicant, even though the linear correlation term of a plurality of the decoded high band sub-band power estimation coefficients is communized in a three-pattern degree, it has known that deterioration of sound quality of audibility of sound subjected to the frequency band expansion process does not almost occur. Therefore, it is possible for the coefficient learning apparatus 81 to decrease the recording area required in recording the decoded high band sub-band power estimation coefficient without deteriorating sound quality of sound after the frequency band expansion process.

20

[0475]

As described above, the coefficient learning apparatus 81 produces the decoded high band sub-band power estimation coefficient of each coefficient index from the supplied broadband instruction signal, and output the produced coefficient.

25

[0476]

In addition, in the coefficient learning process in Fig. 29, the description is made that the residual vector is normalized. However, the normalization of the residual

30

vector may not be performed in one or both of step S436 and step S441.

[0477]

In addition, the normalization of the residual vector is performed and thus communization of the linear correlation term of the decoded high band sub-band power estimation coefficient may not be performed. In this case, the normalization process is performed in step S436 and then the normalized residual vector is clustered in the same number of clusters as that of the decoded high band sub-band power estimation coefficient to be obtained. In addition, the frames of the residual error included in each cluster are used to perform the regression analysis for each cluster and the decoded high band sub-band power estimation coefficient of each cluster is produced.

[0478]

<7. Seventh Embodiment>

[Regarding Sharing of Coefficient Table]

Incidentally, in the above description, it has been described that, in order to obtain the high band sub-band signals of the sub-band ib on the high band side in which the index is ib (wherein, $sb+1 \leq ib \leq eb$), the coefficients $A_{ib}(sb-3)$ to $A_{ib}(sb)$ and the coefficient B_{ib} as the decoding high band sub-band power estimation coefficients are used.

[0479]

Since the high band components includes $(eb-sb)$ sub-bands of the sub-bands $sb+1$ to eb , a coefficient set illustrated in, for example, Fig. 30 is necessary in order to obtain a decoded high band signal including the high band sub-band signals of the respective sub-bands.

[0480]

That is, the coefficients $A_{sb+1}(sb-3)$ to $A_{sb+1}(sb)$ in the uppermost row of Fig. 30 are coefficients which are multiplied by the respective low band sub-band powers of the sub-bands sb-3 to sb on the low band side in order to obtain the decoding high band sub-band power of the sub-band sb+1. In addition, the coefficient B_{sb+1} in the uppermost row of the drawing is a constant term of a linear combination of the low band sub-band powers for obtaining the decoding high band sub-band power of the sub-band sb+1.

10 [0481]

Similarly, the coefficients $A_{eb}(sb-3)$ to $A_{eb}(sb)$ in the lowermost row of the drawing are coefficients which are multiplied by the respective low band sub-band powers of the sub-bands sb-3 to sb on the low band side in order to obtain the decoding high band sub-band power of the sub-band eb. In addition, the coefficient B_{eb} in the lowermost row of the drawing is a constant term of a linear combination of the low band sub-band powers for obtaining the decoding high band sub-band power of the sub-band eb.

20 [0482]

In this way, in the encoder 30 and the decoder 40, $5 \times (eb-sb)$ coefficient sets are recorded in advance as the decoding high band sub-band power estimation coefficients which are specified by one coefficient index. Hereinafter, these $5 \times (eb-sb)$ coefficient sets as the decoding high band sub-band power estimation coefficients will be referred to as the coefficient tables.

[0483]

For example, when it is attempted to obtain the decoded high band signal including more than $(eb-sb)$ sub-bands, the coefficient table illustrated in Fig. 30 lacks the coefficients

and thus the decoded high band signals are not obtained appropriately. Conversely, when it is attempted to obtain the decoded high band signals including less than (eb-sb) sub-bands, the coefficient table illustrated in Fig. 30 have
5 many redundant coefficients.

[0484]

Therefore, in the encoder 30 and the decoder 40, many coefficient tables should be recorded in advance to correspond to the number of sub-bands constituting the decoded high band
10 signals and thus there is a case where the size of a recoding area where coefficient tables are recorded increases.

[0485]

Therefore, by recording a coefficient table for obtaining the decoded high band signals of a predetermined
15 number of sub-bands and extending or reducing the coefficient table, the decoded high band signals having different numbers of sub-bands may be handled.

[0486]

Specifically, for example, it is assumed that a
20 coefficient table of a case where Index $eb=sb+8$ is recorded in the encoder 30 and the decoder 40. In this case, when the respective coefficients constituting the coefficient table are used, the decoded high band signal having 8 sub-bands can be obtained.

25 [0487]

Here, for example, as illustrated on the left side of Fig. 31, when it is attempted to obtain the decoded high band signal including 10 sub-bands of the sub-bands $sb+1$ to $sb+10$, the coefficient table which is recorded in the encoder 30 and
30 the decoder 40 lacks coefficients. That is, the coefficients $A_{1b}(kb)$ and B_{1b} of the sub-bands $sb+9$ and $sb+10$ are lacking.

[0488]

Therefore, when the coefficient table is extended as illustrated on the right side of the drawing, by using the coefficient table of the case where there are 8 sub-bands on the high band side, the decoded high band signal including 10 sub-bands can be appropriately obtained. Here, in the drawing, the horizontal axis represents the frequency and the vertical axis represents the power. In addition, the respective frequency components of an input signal are illustrated on the left side of the drawing, and lines in the vertical direction indicate the boundary positions of the respective sub-bands on the high band side.

[0489]

In an example of Fig. 31, the coefficients $A_{sb+8}(sb-3)$ to $A_{sb+8}(sb)$ and the coefficient B_{sb+8} of the sub-band $sb+8$ as the decoding high band sub-band power estimation coefficients are used as the coefficients of the sub-bands $sb+9$ and $sb+10$ without any change.

[0490]

That is, in the coefficient table, the coefficients $A_{sb+8}(sb-3)$ to $A_{sb+8}(sb)$ and the coefficient B_{sb+8} of the sub-band $sb+8$ are duplicated and used as the coefficients $A_{sb+9}(sb-3)$ to $A_{sb+9}(sb)$ and the coefficient B_{sb+9} of the sub-band $sb+9$ without any change. Similarly, in the coefficient table, the coefficients $A_{sb+8}(sb-3)$ to $A_{sb+8}(sb)$ and the coefficient B_{sb+8} of the sub-band $sb+8$ are duplicated and used as the coefficients $A_{sb+10}(sb-3)$ to $A_{sb+10}(sb)$ and the coefficient B_{sb+10} of the sub-band $sb+10$ without any change.

[0491]

In this way, when a coefficient table is extended, the coefficients $A_{ib}(kb)$ and B_{ib} of a sub-band having the highest

frequency in the coefficient table are used for lacking coefficients of a sub-band without any change.

[0492]

In addition, even when the estimation accuracy of components of a sub-band having a high frequency of high band components such as the sub-bands $sb+9$ and $sb+10$ deteriorates to some degree, there is no deterioration in audibility at the time of the reproduction of an output signal including the decoded high band signals and the decoding low band signals.

10 [0493]

In addition, the extension of the coefficient table is not limited to the example of duplicating the coefficients $A_{ib}(kb)$ and B_{ib} of the sub-band having the highest frequency and setting the duplicated coefficients to coefficients of other sub-bands. The coefficients of some sub-bands of the coefficient table may be duplicated and set to coefficients of the sub-bands which are to be extended (which are lacking). In addition, the coefficients to be duplicated are not limited to those of one sub-band. The coefficients of plural sub-bands may be duplicated and respectively set to coefficients of plural sub-bands to be extended. Furthermore, the coefficients of sub-bands to be extended may be calculated based on the coefficients of some sub-bands.

20 [0494]

25 On the other hand, for example, it is assumed that a coefficient table of a case where $Index\ eb=sb+8$ is recorded in the encoder 30 and the decoder 40 and a decoded high band signal including 6 sub-bands is produced as illustrated on, for example, on the left side of Fig. 32. Here, in the drawing, 30 the horizontal axis represents the frequency and the vertical axis represents the power. In addition, the respective

frequency components of an input signal are illustrated on the left side of the drawing, and lines in the vertical direction indicate the boundary positions of the respective sub-bands on the high band side.

5 [0495]

In this case, a coefficient table in which there are 6 sub-bands on the high band side is not recorded in the encoder 30 and the decoder 40. Therefore, when the coefficient table is reduced as illustrated on the right side of the drawing, 10 the decoded high band signal including 6 sub-bands can be obtained using the coefficient table in which there are 8 sub-bands on the high band side.

[0496]

In the example of Fig. 32, from the coefficient table 15 as the decoding high band sub-band power estimation coefficients, the coefficients $A_{sb+7}(sb-3)$ to $A_{sb+7}(sb)$ and the coefficient B_{sb+7} of the sub-band $sb+7$ and the coefficients $A_{sb+8}(sb-3)$ to $A_{sb+8}(sb)$ and the coefficient B_{sb+8} of the sub-band $sb+8$ are deleted. In addition, a new coefficient table having 20 the coefficients of six sub-bands of the sub-bands $sb+1$ to $sb+6$, from which the coefficients of the sub-bands $sb+7$ and $sb+8$ are deleted, is used as the decoding high band sub-band power estimation coefficients to produce a decoded high band signal.

25 [0497]

In this way, when a coefficient table is reduced, the coefficients $A_{lb}(kb)$ and B_{lb} of unnecessary sub-bands in the coefficient table, that is, sub-bands which are not used for the production of a decoded high band signals are deleted and 30 thus the reduced coefficient table is obtained.

[0498]

As described above, by appropriately extending or reducing the coefficient table, which is recorded in an encoder and a decoder, to correspond to the number of sub-bands of a decoded high band signal which is to be produced, the coefficient table of a predetermined number of sub-bands can be shared for use. As a result, the size of a recording area of coefficient tables can be reduced.

[0499]

[Functional Configuration Example of Encoder]

10 When a coefficient table is extended or reduced as necessary, an encoder is configured as illustrated in, for example, Fig. 33. In Fig. 33, the same reference numbers are given to parts corresponding to those of the case illustrated in Fig. 18 and the description thereof will be appropriately omitted.

15 [0500]

 An encoder 111 of Fig. 33 is different from the encoder 30 of Fig. 18 in that the pseudo high band sub-band power calculation circuit 35 of the encoder 111 is provided with an extension/reduction unit 121, and the other configurations are the same.

[0501]

 The extension/reduction unit 121 extends or reduces a coefficient table which is recorded by the pseudo high band sub-band power calculation circuit 35 to correspond to the number of sub-bands into which high band components of an input signal are divided. As necessary, the pseudo high band sub-band power calculation circuit 35 calculates pseudo high band sub-band powers using the coefficient table extended or reduced by the extension or reduction unit 121.

30 [0502]

[Description of Encoding Processes]

Next, encoding processes which are performed by the encoder 111 will be described with reference to the flowchart of Fig. 34. Here, since processes of step S471 to step S474
5 are the same as those of step S181 to S184 of Fig. 19, the description thereof will be omitted.

[0503]

In step S475, the extension/reduction unit 121 extends or reduces a coefficient table as the decoding high band
10 sub-band power estimation coefficients, which are recorded by the pseudo high band sub-band power calculation circuit 35, to correspond to the number of the high band sub-bands of the input signal, that is, the number of the high band sub-band signals.

15 [0504]

For example, it is assumed that the high band components of the input signal are divided into high band sub-band signals of q sub-bands of the sub-bands $sb+1$ to $sb+q$. That is, it is assumed that pseudo high band sub-band powers of q sub-bands
20 are calculated based on the low band sub-band signals.

[0505]

In addition, it is assumed that a coefficient table having the coefficients $A_{ib}(kb)$ and B_{ib} of r sub-bands of the sub-bands $sb+1$ to $sb+r$ is recorded in the pseudo high band
25 sub-band power calculation circuit 35 as the decoding high band sub-band power estimation coefficients.

[0506]

In this case, when q is greater than r ($q > r$), the extension/reduction unit 121 extends the coefficient table
30 recorded in the pseudo high band sub-band power calculation circuit 35. That is, the extension/reduction unit 121

duplicates the coefficients $A_{sb+r}(kb)$ and B_{sb+r} of the sub-band $sb+r$ included in the coefficient table and sets the duplicated coefficients to coefficients of the respective sub-bands of the sub-bands $sb+r+1$ to $sb+q$ without any change. As a result,
 5 a coefficient table having the coefficients $A_{ib}(kb)$ and B_{ib} of q sub-bands is obtained.

[0507]

In this case, when q is less than r ($q < r$), the extension/reduction unit 121 reduces the coefficient table
 10 recorded in the pseudo high band sub-band power calculation circuit 35. That is, the extension/reduction unit 121 deletes the coefficients $A_{ib}(kb)$ and B_{ib} of the respective sub-bands of the sub-bands $sb+q+1$ to $sb+r$ included in the coefficient table. As a result, a coefficient table having the
 15 coefficients $A_{ib}(kb)$ and B_{ib} of the respective sub-bands of the sub-bands $sb+1$ to $sb+q$ is obtained.

[0508]

Furthermore, when q is equal to r ($q=r$), the extension/reduction unit 121 neither extends nor reduces the
 20 coefficient table recorded in the pseudo high band sub-band power calculation circuit 35.

[0509]

In step S476, the pseudo high band sub-band power calculation circuit 35 calculates pseudo high band sub-band
 25 power differences based on the characteristic amounts supplied from the characteristic amount calculation circuit 34 to be supplied to the pseudo high band sub-band power difference calculation circuit 36.

[0510]

30 For example, the pseudo high band sub-band power calculation circuit 35 performs the calculation according to

the above-described expression (2) using the coefficient table, which is recorded as the decoding high band sub-band power estimation coefficients and, as necessary, is extended or reduced by the extension/reduction unit 121, and the low band sub-band powers $\text{power}(\text{kb}, J)$ (wherein, $\text{sb}-3 \leq \text{kb} \leq \text{sb}$); and calculates the pseudo high band sub-band powers $\text{power}_{\text{est}}(\text{ib}, J)$.

[0511]

That is, the low band sub-band powers of the respective sub-bands on the low band side which are supplied as the characteristic amounts are multiplied by the coefficients $A_{\text{ib}}(\text{kb})$ for the respective sub-bands, the coefficients B_{ib} are further added to the sums of the low band sub-band powers which have been multiplied by the coefficients, and thus the pseudo high band sub-band powers $\text{power}_{\text{est}}(\text{ib}, J)$ are obtained.

These pseudo high band sub-band powers are calculated for the respective sub-bands on the high band side.

[0512]

In addition, the pseudo high band sub-band power calculation circuit 35 performs the calculation of the pseudo high band sub-band powers for the respective decoding high band sub-band power estimation coefficients (coefficient table) which are recorded in advance. For example, it is assumed that K decoding high band sub-band power estimation coefficients in which the coefficient index is 1 to K (wherein, $2 \leq K$) are prepared in advance. In this case, for K decoding high band sub-band power estimation coefficients, as necessary, the coefficient table is extended or reduced and the pseudo high band sub-band powers of the respective sub-bands are calculated.

[0513]

In this way, when the coefficient tables is extended or reduced as necessary, the pseudo high band sub-band powers of the sub-bands sb+1 to eb can be appropriately calculated using the coefficient table which is recorded in advance, irrespective of the number of sub-bands on the high band side. Furthermore, the pseudo high band sub-band powers can be obtained with less decoding high band sub-band power estimation coefficients and higher efficiency.

[0514]

10 After the pseudo high band sub-band powers are calculated in step S476, processes of step S477 and S478 are performed and the square sums of the pseudo high band sub-band power differences are calculated. Here, since these processes are the same as those of step S186 and step S187 of Fig. 19, the description thereof will be omitted.

[0515]

In addition, in step S478, for K decoding high band sub-band power estimation coefficients, the sums of square differences $E(J, id)$ are calculated. The pseudo high band sub-band power difference calculation circuit 36 selects the smallest sum of square differences among the calculated K sums of square differences $E(J, id)$ and supplies the coefficient index, which indicates the decoding high band sub-band power estimation coefficients corresponding to the selected sum of square differences, to the high band encoding circuit 37.

[0516]

After the coefficient index capable of estimating high band signals with highest accuracy is selected and supplied to the high band encoding circuit 37, processes of step S479 and Step S480 are performed and the encoding processes end. Here, since these processes are the same as those of step S188

and step S189 of Fig. 19, the description thereof will be omitted.

[0517]

In this way, by outputting the low band encoded data
5 and the high band encoded data as an output code string, in
a decoder which receives the input of the output code string,
the decoding high band sub-band power estimation coefficients,
which are optimum for frequency band expansion process, can
be obtained. As a result, a signal with higher sound quality
10 can be obtained.

[0518]

Furthermore, it is not necessary for the encoder 111
to record coefficient tables for the number of sub-bands into
which high band components of an input signal are divided and
15 thus a sound can be encoded with less coefficient tables and
higher efficiency.

[0519]

In addition, information indicating the number of
sub-bands into which high band components of an input signal
20 are divided may be included in the high band encoded data or
information indicating the number of sub-bands may be
transmitted to a decoder as separate data from the output code
string.

[0520]

25 [Functional Configuration Example of Decoder]

In addition, a decoder which receives the output code
string, output from the encoder 111 of Fig. 33, as an input
code string to be decoded is configured as illustrated in,
for example, Fig. 35. In Fig. 35, the same reference numbers
30 are given to parts corresponding to those of the case
illustrated in Fig. 20 and the description thereof will be

appropriately omitted.

[0521]

A decoder 151 of Fig. 35 is the same as the decoder 40 of Fig. 20 in that the demultiplexing circuit 41 to the synthesis unit 48 are provided, but is different from the decoder 40 of Fig. 20 in that the decoding high band sub-band power calculation circuit 46 is provided with an extension and reduction unit 161.

[0522]

As necessary, the extension and reduction unit 161 extends or reduces a coefficient table as the decoding high band sub-band power estimation coefficients, which is supplied from the high band decoding circuit 45. The decoding high band sub-band power calculation circuit 46 calculates the decoded high band sub-band powers using the coefficient table extended or reduced as necessary.

[0523]

[Description of Decoding Process]

Next, decoding processes which are performed by the decoder 151 of Fig. 35 will be described with reference to the flowchart of Fig. 36. Since processes of step S511 to step S515 are the same as those of step S211 to step S215 of Fig. 21, the description thereof will be omitted.

[0524]

In step S516, as necessary, the extension and reduction unit 161 extends or reduces the coefficient table as the decoding high band sub-band power estimation coefficients supplied from the high band decoding circuit 45.

[0525]

Specifically, the decoding high band sub-band power calculation circuit 46 calculates decoded high band sub-band

powers of q sub-bands of the sub-bands $sb+1$ to $sb+q$ on the high band side. That is, it is assumed that the decoded high band signal includes components of q sub-bands.

[0526]

5 Here, the number of sub-bands " q " on the high band side may be specified in advance in the decoder 151 or may be specified by the user. In addition, the information indicating the number of sub-bands on the high band side may be included in the high band encoded data or the information indicating the
10 number of sub-bands on the high band side may be transmitted from the encoder 111 to the decoder 151 as separate data from the input code string.

[0527]

 In addition, it is assumed that a coefficient table
15 having the coefficients $A_{ib}(kb)$ and B_{ib} of r sub-bands of the sub-bands $sb+1$ to $sb+r$ is recorded in the high band decoding circuit 45 as the decoding high band sub-band power estimation coefficients.

[0528]

20 In this case, when q is greater than r ($q > r$), the extension and reduction unit 161 extends the coefficient table supplied from the high band decoding circuit 45. That is, the extension and reduction unit 161 duplicates the coefficients $A_{sb+r}(kb)$ and B_{sb+r} of the sub-band $sb+r$ included
25 in the coefficient table and sets the duplicated coefficients to coefficients of the respective sub-bands of the sub-bands $sb+r+1$ to $sb+q$ without any change. As a result, a coefficient table having the coefficients $A_{ib}(kb)$ and B_{ib} of q sub-bands is obtained.

30 [0529]

 In this case, when q is less than r ($q < r$), the extension

and reduction unit 161 reduces the coefficient table supplied from the high band decoding circuit 45. That is, the extension and reduction unit 161 deletes the coefficients $A_{ib}(kb)$ and B_{ib} of the respective sub-bands of the sub-bands $sb+q+1$ to $sb+r$ included in the coefficient table. As a result, a
5 coefficient table having the coefficients $A_{ib}(kb)$ and B_{ib} of the respective sub-bands of the sub-bands $sb+1$ to $sb+q$ is obtained.

[0530]

10 Furthermore, when q is equal to r ($q=r$), the extension and reduction unit 161 neither extends nor reduces the coefficient table supplied from the high band decoding circuit 45.

[0531]

15 After the coefficient table is extended or reduced as necessary, processes of step S517 to step S519 are performed and the decoding processes end. However, since these processes are the same as those of step S216 to step S218 in Fig. 21, the description thereof will be omitted.

20 [0532]

In this way, according to the decoder 151, the coefficient index is obtained from the high band encoded data obtained from the demultiplexing of the input code string; using the decoding high band sub-band power estimation
25 coefficients indicated by the coefficient index, the decoded high band sub-band powers are calculated; and thus the estimation accuracy of the high band sub-band powers can be improved. As a result, a sound signal with higher quality can be reproduced.

30 [0533]

Furthermore, in the decoder 151, it is not necessary

that coefficient tables are recorded for the number of sub-bands constituting a decoded high band signal; and as a result, a sound can be decoded with less coefficient tables and higher efficiency.

5 [0534]

<8. Eighth Embodiment>

[Regarding Blended Learning Method]

In the above-described cases, coefficient sets capable of dealing with the differences of the band-limited frequency, the sampling frequency, the coding, and the encoding algorithms are prepared, but there is a problem in that the size of tables increases. To deal with this problem, a method is contrived in which, using various band-limited frequencies, sampling frequencies, codings, and encoding algorithms as input, explanatory variables (sb-3 to sb) and explained variables (sb+1 to eb) are prepared and these are blended to perform learning. According to this method, for signals of various sampling frequencies, codings, and encoding algorithms, high band powers can be accurately estimated on average with one table.

10
15
20

[0535]

Specifically, for example, as illustrated in Fig. 37, for the respective conditions A to D, explanatory variables and explained variables are obtained from broadband instruction signals and decoding high band sub-band power estimation coefficients (coefficient table) are obtained by learning.

25

[0536]

In addition, in Fig. 37, the band-limited frequency represents the highest frequency among frequencies of components included in a low band signal or a decoding low

30

band signal, and the sampling frequency represents the sampling frequency of an input signal or an output signal. In addition, the coding represents a coding system of an input signal, and the encoding algorithm represents an encoding method of a sound.

5 For example, when encoding algorithms are different, decoding low band signals are different. As a result, for example, values of low band sub-band powers which are used as explained variables are different.

[0537]

10 In a case where coefficient tables are obtained for the respective conditions, when a sound is encoded or decoded, one coefficient table is selected according to the conditions such as the coding and the encoding algorithm from the coefficient tables obtained for the conditions.

15 [0538]

When the coefficient tables are obtained for the respective conditions as described above, in an encoder and a decoder, many coefficient tables should be recorded in advance for the respective conditions. Accordingly, there
20 is a case where the size of a recording area where the coefficient tables are recorded increases.

[0539]

Therefore, explanatory variables and explained variables, which are obtained from broadband instruction
25 signals for the respective conditions, may be blended and perform learning; and using a coefficient table thus obtained, high band powers may be accurately estimated on average, irrespective of the conditions.

[0540]

30 [Functional Configuration Example of Coefficient Learning Apparatus]

In such a case, a coefficient learning apparatus, which produces a coefficient table as the decoding high band sub-band power estimation coefficients by learning, is configured as illustrated in, for example, Fig. 38.

5 [0541]

A coefficient learning apparatus 191 includes a sub-band division circuit 201, a high band sub-band power calculation circuit 202, a characteristic amount calculation circuit 203, and a coefficient estimation circuit 204.

10 [0542]

To this coefficient learning apparatus 191, a plurality of musical data with plural conditions which have different conditions such as conditions A to D illustrated in Fig. 37 are supplied as the broadband instruction signals. The broadband instruction signal represents a signal including plural high band sub-band components and plural low band sub-band components.

[0543]

The sub-band division circuit 201 includes a band pass filter and divides a supplied broadband instruction signal into plural sub-band signals to be output to the high band sub-band power calculation circuit 202 and the characteristic amount calculation circuit 203. Specifically, high band sub-band signals of the respective sub-band on the high band side in which the index is sb+1 to eb are supplied to the high band sub-band power calculation circuit 202, and low band sub-band signals of the respective sub-band on the low band side in which the index is sb-3 to sb are supplied to the characteristic amount calculation circuit 203.

30 [0544]

The high band sub-band power calculation circuit 202

calculates high band sub-band powers of the respective high band sub-band signals supplied from the sub-band division circuit 201 to be output to the coefficient estimation circuit 204.

5 The characteristic amount calculation circuit 203 calculates low band sub-band powers as the characteristic amounts based on the low band sub-band signals supplied from the sub-band division circuit 201 to be output to the coefficient estimation circuit 204.

10 [0545]

 The coefficient estimation circuit 204 performs regression analysis using the high band sub-band powers supplied from the high band sub-band power calculation circuit 202 and the characteristic amounts supplied from the characteristic amount calculation circuit 203, thereby
15 generating and outputting decoding high band sub-band power estimation coefficients.

[0546]

[Description of Coefficient Learning Processes]

20 Next, coefficient learning processes which are performed by the coefficient learning apparatus 191 will be described with reference to the flowchart of Fig. 39.

[0547]

 In step S541, the sub-band division circuit 201 divides
25 plural supplied broadband instruction signals into plural sub-band signals, respectively. In addition, the sub-band division circuit 201 supplies high band signals of sub-bands, in which the index is sb+1 to eb, to the high band sub-band power calculation circuit 202 and supplies low band signals
30 of sub-bands, in which the index is sb-3 to sb, to the characteristic amount calculation circuit 203.

[0548]

The broadband instruction signal supplied to the sub-band division circuit 201 includes a plurality of musical data which have different conditions such as the sampling
5 frequency. In addition, the broadband instruction signal is divided according to the different conditions, for example, is divided into the low band sub-band signals and high band sub-band signals according to different band-limited frequencies.

10 [0549]

In step S542, the high band sub-band power calculation circuit 202 performs the same calculation as that of the above-described expression (1) with respect to the respective highband sub-band signals supplied from the sub-band division
15 circuit 201; and thus calculates high band sub-band powers to be output to the coefficient estimation circuit 204.

[0550]

In step S543, the characteristic amount calculation circuit 203 performs the same calculation as that of the
20 above-described expression (1) with respect to the respective low band sub-band signals supplied from the sub-band division circuit 201; and thus calculates low band sub-band powers as the characteristic amounts to be output to the coefficient estimation circuit 204.

25 [0551]

As a result, with respect to the respective frames of plural broadband instruction signals, the high band sub-band powers and the low band sub-band powers are supplied to the coefficient estimation circuit 204.

30 [0552]

In step S544, the coefficient estimation circuit 204

performs regression analysis using a least-squares method to calculate the coefficients $A_{ib}(kb)$ and B_{ib} for the respective sub-bands ib (wherein, $sb+1 \leq ib \leq eb$) on the high band side in which the index is $sb+1$ to eb .

5 [0553]

In the regression analysis, the low band sub-band powers supplied from the characteristic amount calculation circuit 203 are set to explanatory variables, and the high band sub-band powers supplied from the high band sub-band power calculation circuit 202 are set to explained variables. In addition, the regression analysis is performed using the low band sub-band powers and the high band sub-band powers of all the frames, which constitute all the broadband instruction signals supplied to the coefficient learning apparatus 191.

15 [0554]

In step S545, the coefficient estimation circuit 204 obtains residual vectors of the respective frames of the broadband instruction signals using the obtained coefficients $A_{ib}(kb)$ and B_{ib} of the respective sub-bands ib .

20 [0555]

For example, the coefficient estimation circuit 204 subtracts the sums between the sum total of the low band sub-band powers $power(kb, J)$ (wherein, $sb-3 \leq kb \leq sb$) which are multiplied by the coefficients $A_{ib}(kb)$; and the sum of the coefficients B_{ib} , from the high band sub-band powers $power(ib, J)$ for the respective sub-bands ib (wherein, $sb+1 \leq ib \leq eb$) of the frame J , thereby calculating residual errors. In addition, vectors including the residual errors of the respective sub-bands ib of the frame J are set to the residual vectors.

30 [0556]

In addition, the residual vectors are calculated for

all the frames, which constitute all the broadband instruction signals supplied to the coefficient learning apparatus 191.
[0557]

In step S546, the coefficient estimation circuit 204
5 clusters the residual vectors, obtained for the respective frames, into some clusters according to a k-means method or the like.
[0558]

In addition, the coefficient estimation circuit 204
10 calculates central vectors of the clusters for the respective clusters and calculates the distances between the central vectors and the residual vectors of the clusters with respect to the residual vectors of the respective frames. In addition, the coefficient estimation circuit 204 specifies the clusters
15 belonging to the respective frames, based on the calculated distances. That is, a cluster having a central vector, which has the shortest distance with a residual vector of a frame, is set to the cluster which belongs to the frame.
[0559]

20 In step S547, the coefficient estimation circuit 204 selects one of the plural clusters, obtained by clustering, as a process target cluster.
[0560]

In step S548, the coefficient estimation circuit 204
25 calculates the coefficients $A_{ib}(kb)$ and B_{ib} of the respective sub-bands ib (wherein, $sb+1 \leq ib \leq eb$) by regression analysis using a frame of a residual vector which belongs to the cluster selected as the process target cluster.
[0561]

30 That is, when the frame of the residual vector which belongs to the process target cluster is referred to as the

process target frame, the low band sub-band powers and the high band sub-band powers of all the process target frames are set to explanatory variables and explained variables, thereby performing the regression analysis using a

5 least-squares method. As a result, the coefficients $A_{ib}(kb)$ and B_{ib} are obtained for the respective sub-bands ib .

[0562]

A coefficient table having the coefficients $A_{ib}(kb)$ and B_{ib} of the respective sub-bands thus obtained are set to the
10 decoding high band sub-band power estimation coefficients and a coefficient index is given to this decoding high band sub-band power estimation coefficients.

[0563]

In step S549, the coefficient learning apparatus 191
15 determines whether or not all the clusters are processes as the process target cluster. In step S549, when it is determined that all the clusters has yet to be processed, the process returns to step S547 and the above-described processes are repeated. That is, the next cluster is selected as the process
20 target and the decoding high band sub-band power estimation coefficients are calculated.

[0564]

On the other hand, in step S549, when it is determined that all the clusters are processed, a predetermined number
25 of decoding high band sub-band power estimation coefficients, which have been desired to be obtained, are obtained. Therefore, the process processes to step S550.

[0565]

In step S550, the coefficient estimation circuit 204
30 outputs the obtained coefficient index and the obtained decoding high band sub-band power estimation coefficients to

an encoder or a decoder to be recorded and the coefficient learning processes end.

[0566]

In this way, the coefficient learning apparatus 191
5 produces the decoding high band sub-band power estimation
coefficients (coefficient table) of the respective
coefficient indices from the supplied broadband instruction
signals to be output. In this way, learning is performed using
plural broadband instruction signals which have different
10 conditions to produce a coefficient table; and as a result,
the size of a recording area of coefficient tables can be reduced
and high band sub-band powers can be accurately estimated on
average.

[0567]

15 The serial process described above is performed by a
hardware and a software. When a serial process is performed
by the software, a program constituted by the software is
installed to a computer incorporated into an indicated software
or a general-purpose personal computer capable of executing
20 various functions by installing various programs from a program
recording medium.

[0568]

Fig. 40 is block diagram illustrating a configuration
example of the hardware of a computer performing a series of
25 processes described above by the computer.

[0569]

In the computer, a CPU 501, a ROM (Read Only Memory)
502 and a RAM (Random Access Memory) 503 are connected each
other by a bus 504.

30 [0570]

In addition, an input/output interface 505 is connected

to the bus 504. An input unit 506 including a key board, an mouse a microphone and the like, an output unit 507 including a display, a speaker and the like, a storage unit 508 including a hard disk or non-volatile memory and the like, a communication
5 unit 509 including a network interface and the like, and a drive 510 that drives a removable medium 511 of a magnetic disc, an optical disc, a magneto-optical disc and semiconductor memory and the like are connected to the input/output interface 505.

10 [0571]

In the computer configured as described above, for example, the CPU 501 loads and executes the program stored in the storage unit 508 to the RAM 503 via the input/output interface 505 and the bus 504 to perform a series of processes
15 described above.

[0572]

The program to be executed by the computer (CPU 501), for example, is recorded in a removable medium 511 such as a package medium including a magnetic disk, (including a
20 flexible disc), an optical disc ((CD-ROM (Compact Disc-Read Only Memory)), DVD (Digital Versatile Disc) and the like), a magneto-optical disc or a semiconductor memory, or is provided via a wire or wireless transmission medium including a local area network, an internet and a digital satellite
25 broadcasting.

[0573]

In addition, the program can be installed to the storage unit 508 via the input/output interface 505 by mounting the removable medium 511 to the drive 510. In addition, the program
30 is received in the communication unit 509 via the wire or wireless transmission medium and can be installed to the

storage unit 508. In addition, the program can be installed in the ROM 502 or the storage unit 508 in advance.

[0574]

In addition, the program performed by the computer may
 5 be a program where the process is performed in time sequence according the sequence described in the specification and a program where the process is performed in parallel or in timing necessary when a call is made.

[0575]

10 In addition, the embodiment of the present invention is not limited the embodiment described above and various modifications is possible within a scope apart from a gist of the present invention.

15 REFERENCE SIGNS LIST

[0576]

- 10 Frequency Band Expansion Apparatus
- 11 Low-pass filter
- 12 Delay Circuit
- 20 13, 13-1 to 13-N Band Pass Filter
- 14 Characteristic Amount Calculation Circuit
- 15 High Band Sub-Band Power Estimation Circuit
- 16 High Band Signal Production Circuit
- 17 High-pass filter
- 25 18 Signal Adder
- 20 Coefficient Learning Apparatus
- 21, 21-1 to 21-(K+N) Band Pass Filter
- 22 High Band Sub-Band Power Calculation Circuit
- 23 Characteristic Amount Calculation Circuit
- 30 24 Coefficient Estimation Circuit
- 30 Encoder

	31	Low-pass filter
	32	Low Band Encoding Circuit
	33	Sub-Band Division Circuit
	34	Characteristic Amount Calculation Circuit
5	35	Pseudo High Band Sub-Band Power Calculation Circuit
	36	Pseudo High Band Sub-band Power Difference Calculation Circuit
	37	High Band Encoding Circuit
	38	Multiplexing Circuit
10	40	Decoder
	41	Demultiplexing Circuit
	42	Low Band Decoding Circuit
	43	Sub-Band Division Circuit
	44	Characteristic Amount Calculation Circuit
15	45	High Band Decoding Circuit
	46	Decoded High Band Sub-Band Power Calculation Circuit
	47	Decoded High Band Signal Production Circuit
	48	Synthesis circuit
	50	Coefficient Learning Apparatus
20	51	Low-pass filter
	52	Sub-Band Division Circuit
	53	Characteristic Amount Calculation Circuit
	54	Pseudo High Band Sub-Band Power Calculation Circuit
	55	Pseudo High Band Sub-Band Power Difference Calculation Circuit
25	56	Pseudo High Band Sub-Band Power Difference Clustering Circuit
	57	Coefficient Estimation Circuit
	101	CPU
30	102	ROM
	103	RAM

- 104 Bus
- 105 Input/Output Interface
- 106 Input Unit
- 107 Output Unit
- 5 108 Storage Unit
- 109 Communication Unit
- 110 Drive
- 111 Removable Medium

CLAIMS

1. A signal processing apparatus comprising:
 - a demultiplexing unit that demultiplexes input encoded
 - 5 data to at least low band encoded data and coefficient
 - information;
 - a low band decoding unit that decodes the low band encoded
 - data to produce low band signals;
 - a selection unit that selects a coefficient table which
 - 10 is obtained based on the coefficient information among a
 - plurality of coefficient tables used for the production of
 - high band signals and having coefficients for the respective
 - sub-bands on a high band side;
 - an extension and reduction unit that deletes the
 - 15 coefficients of some sub-bands to reduce the coefficient table
 - or produces the coefficients of predetermined sub-bands based
 - on the coefficients of some sub-bands to extend the coefficient
 - table;
 - a high band sub-band power calculation unit that
 - 20 calculates high band sub-band powers of high band sub-band
 - signals of the respective sub-bands constituting the high band
 - signals based on low band sub-band signals of the respective
 - sub-bands constituting the low band signals and the extended
 - or reduced coefficient table; and
 - 25 a high band signal production unit that produces the
 - high band signals based on the high band sub-band powers and
 - the low band sub-band signals.
2. The signal processing apparatus according to claim 1,
- 30 wherein the extension and reduction unit duplicates the
- coefficients of a sub-band having a highest frequency which

is included in the coefficient table and sets the duplicated coefficients to coefficients of a sub-band having a higher frequency than the highest frequency to extend the coefficient table.

5

3. The signal processing apparatus according to claim 1, wherein the extension and reduction unit deletes the coefficients of a sub-band, which has a higher frequency than that of a sub-band having a highest frequency among sub-bands of the high band sub-band signals, from the coefficient table to reduce the coefficient table.

4. A signal processing method of a signal processing apparatus which includes

15 a demultiplexing unit that demultiplexes input encoded data to at least low band encoded data and coefficient information,

a lowband decoding unit that decodes the lowband encoded data to produce low band signals,

20 a selection unit that selects a coefficient table which is obtained based on the coefficient information among a plurality of coefficient tables used for the production of high band signals and having coefficients for the respective sub-bands on a high band side,

25 an extension and reduction unit that deletes the coefficients of some sub-bands to reduce the coefficient table or produces the coefficients of predetermined sub-bands based on the coefficients of some sub-bands to extend the coefficient table,

30 a high band sub-band power calculation unit that calculates high band sub-band powers of high band sub-band

signals of the respective sub-bands constituting the high band signals based on low band sub-band signals of the respective sub-bands constituting the low band signals and the extended or reduced coefficient table, and

5 a high band signal production unit that produces the high band signals based on the high band sub-band powers and the low band sub-band signals,

 the method comprising the steps of causing:

 the demultiplexing unit to demultiplex the encoded data;

10 the low band decoding unit to produce the low band signals;

 the selection unit to select the coefficient table;

 the extension and reduction unit to reduce or extend the coefficient table;

15 the high band sub-band power calculation unit to calculate the high band sub-band powers; and

 the high band signal production unit to produce the high band signals.

20 5. A program causing a computer to execute processes including the steps of:

 demultiplexing input encoded data to at least low band encoded data and coefficient information;

25 decoding the low band encoded data to produce low band signals;

 selecting a coefficient table which is obtained based on the coefficient information among a plurality of coefficient tables used for the production of high band signals and having coefficients for the respective sub-bands on a high band side;

30 deleting the coefficients of some sub-bands to reduce the coefficient table or generating the coefficients of

predetermined sub-bands based on the coefficients of some sub-bands to extend the coefficient table;

calculating high band sub-band powers of high band sub-band signals of the respective sub-bands constituting the high band signals based on low band sub-band signals of the
5 respective sub-bands constituting the low band signals and the extended or reduced coefficient table; and

generating the high band signals based on the high band sub-band powers and the low band sub-band signals.

10

6. A signal processing apparatus comprising:

a sub-band division unit that produces low band sub-band signals of a plurality of sub-bands on a low band side of an input signal and high band sub-band signals of a plurality
15 of sub-bands on a high band side of the input signal;

an extension and reduction unit that deletes the coefficients of some sub-bands to reduce a coefficient table or produces coefficients of predetermined sub-bands based on coefficients of some sub-bands to extend a coefficient table,
20 the coefficient table having the coefficients for the respective sub-bands on the high band side;

a pseudo high band sub-band power calculation unit that calculates pseudo high band sub-band powers, which are estimated values of powers of the high band sub-band signals,
25 for the respective sub-bands on the high band side based on the extended or reduced coefficient table and the low band sub-band signals;

a selection unit that compares high band sub-band powers of the high band sub-band signals and the pseudo high band sub-band powers to each other and selects one of a plurality
30 of the coefficient tables; and

a production unit that produces data containing coefficient information for obtaining the selected coefficient table.

5 7. The signal processing apparatus according to claim 6,
wherein the extension and reduction unit duplicates the
coefficients of a sub-band having a highest frequency which
is included in the coefficient table and sets the duplicated
coefficients to coefficients of a sub-band having a higher
10 frequency than the highest frequency to extend the coefficient
table.

8. The signal processing apparatus according to claim 6,
wherein the extension and reduction unit deletes the
15 coefficients of a sub-band, which has a higher frequency than
that of a sub-band having a highest frequency among sub-bands
of the high band sub-band signals, from the coefficient table
to reduce the coefficient table.

20 9. A signal processing method of a signal processing
apparatus which includes

a sub-band division unit that produces low band sub-band
signals of a plurality of sub-bands on a low band side of an
input signal and high band sub-band signals of a plurality
25 of sub-bands on a high band side of the input signal,

an extension and reduction unit that deletes the
coefficients of some sub-bands to reduce a coefficient table
or produces coefficients of predetermined sub-bands based on
coefficients of some sub-bands to extend a coefficient table,
30 the coefficient table having coefficients for the respective
sub-bands on the high band side,

a pseudo high band sub-band power calculation unit that calculates pseudo high band sub-band powers, which are estimated values of powers of the high band sub-band signals, for the respective sub-bands on the high band side based on
5 the extended or reduced coefficient table and the low band sub-band signals,

a selection unit that compares high band sub-band powers of the high band sub-band signals and the pseudo high band sub-band powers to each other and selects one of a plurality
10 of the coefficient tables, and

a production unit that produces data containing coefficient information for obtaining the selected coefficient table,

the method comprising the steps of causing:

15 the sub-band division unit to produce the low band sub-band signals and the high band sub-band signals;

the extension and reduction unit to reduce or extend the coefficient table;

the pseudo high band sub-band power calculation unit
20 to calculate the pseudo high band sub-band powers;

the selection unit to select the coefficient table; and

the production unit to produce the data containing coefficient information.

25 10. A program causing a computer to execute processes including the steps of:

generating low band sub-band signals of a plurality of sub-bands on a low band side of an input signal and high band sub-band signals of a plurality of sub-bands on a high band
30 side of the input signal;

deleting the coefficients of some sub-bands to reduce

a coefficient table or generating coefficients of predetermined sub-bands based on coefficients of some sub-bands to extend a coefficient table, the coefficient table having coefficients for the respective sub-bands on the high
5 band side;

calculating pseudo high band sub-band powers, which are estimated values of powers of the high band sub-band signals, for the respective sub-bands on the high band side based on the extended or reduced coefficient table and the low band
10 sub-band signals;

comparing high band sub-band powers of the high band sub-band signals and the pseudo high band sub-band powers to each other and selecting one of a plurality of the coefficient tables; and

15 generating data containing coefficient information for obtaining the selected coefficient table.

11. A decoder comprising:

a demultiplexing unit that demultiplexes input encoded
20 data to at least low band encoded data and coefficient information;

a lowband decoding unit that decodes the lowband encoded data to produce low band signals;

a selection unit that selects a coefficient table which
25 is obtained based on the coefficient information among a plurality of coefficient tables used for the production of high band signals and having coefficients for the respective sub-bands on a high band side;

an extension and reduction unit that deletes the
30 coefficients of some sub-bands to reduce the coefficient table or produces the coefficients of predetermined sub-bands based

on the coefficients of some sub-bands to extend the coefficient table;

5 a high band sub-band power calculation unit that calculates high band sub-band powers of high band sub-band signals of the respective sub-bands constituting the high band signals based on low band sub-band signals of the respective sub-bands constituting the low band signals and the extended or reduced coefficient table;

10 a high band signal production unit that produces the high band signals based on the high band sub-band powers and the low band sub-band signals; and

a synthesis unit that synthesizes the low band signal and the high band signal with each other to produce an output signal.

15

12. A decoding method of a decoder which includes

a demultiplexing unit that demultiplexes input encoded data to at least low band encoded data and coefficient information,

20 a lowband decoding unit that decodes the low band encoded data to produce low band signals,

a selection unit that selects a coefficient table which is obtained based on the coefficient information among a plurality of coefficient tables used for the production of high band signals and having coefficients for the respective sub-bands on a high band side,

25 an extension and reduction unit that deletes the coefficients of some sub-bands to reduce the coefficient table or produces the coefficients of predetermined sub-bands based on the coefficients of some sub-bands to extend the coefficient table,

30

a high band sub-band power calculation unit that calculates high band sub-band powers of high band sub-band signals of the respective sub-bands constituting the high band signals based on low band sub-band signals of the respective sub-bands constituting the low band signals and the extended or reduced coefficient table,

a high band signal production unit that produces the high band signals based on the high band sub-band powers and the low band sub-band signals, and

10 a synthesis unit that synthesizes the low band signal and the high band signal with each other to produce an output signal,

the method comprising the steps of causing:

the demultiplexing unit to demultiplex the encoded data;

15 the low band decoding unit to produce the low band signals;

the selection unit to select the coefficient table;

the extension and reduction unit to reduce or extend the coefficient table;

20 the high band sub-band power calculation unit to calculate the high band sub-band powers;

the high band signal production unit to produce the high band signals; and

the synthesis unit to produce the output signal.

25

13. An encoder comprising:

a sub-band division unit that produces low band sub-band signals of a plurality of sub-bands on a low band side of an input signal and high band sub-band signals of a plurality of sub-bands on a high band side of the input signal;

30

an extension and reduction unit that deletes the

coefficients of some sub-bands to reduce a coefficient table
 or produces coefficients of predetermined sub-bands based on
 coefficients of some sub-bands to extend a coefficient table,
 the coefficient table having coefficients for the respective
 5 sub-bands on the high band side;

a pseudo high band sub-band power calculation unit that
 calculates pseudo high band sub-band powers, which are
 estimated values of powers of the high band sub-band signals,
 for the respective sub-bands on the high band side based on
 10 the extended or reduced coefficient table and the low band
 sub-band signals;

a selection unit that compares high band sub-band powers
 of the high band sub-band signals and the pseudo high band
 sub-band powers to each other and selects one of a plurality
 15 of the coefficient tables;

a high band encoding unit that encodes coefficient
 information for obtaining the selected coefficient table to
 produce high band encoded data;

a low band encoding unit that encodes low band signals
 20 of the input signal to produce low band encoded data; and

a multiplexing unit that multiplexes the low band encoded
 data and the high band encoded data to produce an output code
 string.

25 14. An encoding method of an encoder which includes
 a sub-band division unit that produces low band sub-band
 signals of a plurality of sub-bands on a low band side of an
 input signal and high band sub-band signals of a plurality
 of sub-bands on a high band side of the input signal,
 30 an extension and reduction unit that deletes the
 coefficients of some sub-bands to reduce a coefficient table

or produces coefficients of predetermined sub-bands based on coefficients of some sub-bands to extend a coefficient table, the coefficient table having coefficients for the respective sub-bands on the high band side,

5 a pseudo high band sub-band power calculation unit that calculates pseudo high band sub-band powers, which are estimated values of powers of the high band sub-band signals, for the respective sub-bands on the high band side based on the extended or reduced coefficient table and the low band
10 sub-band signals,

 a selection unit that compares high band sub-band powers of the high band sub-band signals and the pseudo high band sub-band powers to each other and selects one of a plurality of the coefficient tables,

15 a high band encoding unit that encodes coefficient information for obtaining the selected coefficient table to produce high band encoded data,

 a low band encoding unit that encodes low band signals of the input signal to produce low band encoded data, and

20 a multiplexing unit that multiplexes the low band encoded data and the high band encoded data to produce an output code string,

 the method comprising the steps of causing:

 the sub-band division unit to produce the low band
25 sub-band signals and the high band sub-band signals;

 the extension and reduction unit to reduce or extend the coefficient table;

 the pseudo high band sub-band power calculation unit to calculate the pseudo high band sub-band powers;

30 the selection unit to select the coefficient table;

 the high band encoding unit to produce the high band

encoded data;

the low band encoding unit to produce the low band encoded data; and

the multiplexing unit to produce the output code string.

5