

SPECIFICATION

Controlling Method and Apparatus for Induction Motor

Technical Field

This invention relates to a controlling method and apparatus for an induction motor (hereinafter referred to simply as "motor") which is applied for running the motor at its highest efficiency at all times irrespective of load variation or frequency variation.

Background Art

Running of a motor by an inverter is normally performed with a fixed ratio of V (voltage)/ f (frequency). If a running frequency has been determined, the motor has a slip at which it exhibits the highest efficiency, and this value does not depend on the magnitude of the load.

In running wherein ratio V/f is fixed, if frequency f is determined, then since the voltage is fixed, as the load varies, the slip varies and accordingly the efficiency varies, and therefore, it is impossible to always run the motor at the highest efficiency. Here, several conventional techniques of the type described are slightly examined.

As conventional technique 1, Japanese Patent Laid-Open No. 41029/1991 discloses a method for reducing the drop in efficiency caused by load variation. According to

conventional technique 1, a running method for an induction motor driven by a voltage type inverter is characterized in that, when the motor enters a steady running condition, an output voltage and an output current of the inverter are multiplied to determine output power P_N and the output voltage of the inverter is reduced to voltage V_1 which is lower by ΔV than output voltage V_0 at the beginning of the steady running. The difference ΔP_1 between output power P_1 of the inverter calculated at that time and output power P_0 of the inverter in the beginning of the steady running is compared with constant ΔP_S and then, if $\Delta P_1 < \Delta P_S$, the inverter output voltage is fixed to the voltage V_1 . But if $\Delta P_1 \geq \Delta P_S$, and after a fixed period the inverter output voltage is further reduced by ΔV from voltage V_1 , the value of the reduction (difference) of the output power at that time and the constant ΔP_S are compared with each other and the inverter output voltage is fixed or further reduced depending upon the relationship in magnitude between them, the steps being repeated to run the motor with the output voltage of the inverter.

Further, as conventional technique 2, Japanese Patent Laid-Open No. 203700/1995 discloses a method wherein a slip to obtain the highest efficiency is calculated from the motor constant and the frequency. The application of voltage to the motor is controlled to obtain the calculated slip. By this method, when the motor enters a steady

running condition, the output voltage and the output current of the inverter are multiplied to determine output power P_N , and the voltage which will achieve the highest efficiency with this power and frequency is calculated and supplied to the motor to control the motor indirectly for optimum slip.

Disclosure of Invention

However, with the method of conventional technique 1, since the minimum power running condition is sought while varying the voltage by small amounts, much time is required until the point of the minimum power is reached, and accordingly, there is a problem that the responsiveness to load variation is low. Meanwhile, with the method of conventional technique 2, since the motor is controlled indirectly to obtain the slip with the highest efficiency, there is a problem with accuracy. Further, as application of vector control to general purpose inverters proceeds, this controlling method has a problem in that it cannot be applied to a vector control inverter in which an excitation current and a current for torque are controlled independently of each other.

The present invention has been made in view of such problems of conventional techniques as described above, and it is an object of the present invention to provide a controlling method and apparatus for an induction motor in

which the slip of the induction motor is controlled rapidly so that the efficiency of the induction motor may always be highest with respect to load variation and frequency variation. In order to solve the problems described above, according to the present invention, a controlling method for an inverter controlled induction motor wherein the slip of the motor is detected and the optimum slip dependent on the motor constant and the frequency supplied to the motor providing the highest efficiency is calculated, the optimum slip and the slip of the motor are compared to determine a voltage manipulated variable which is used to form an output voltage instruction or a magnetic flux instruction for running the motor at the slip which provides the highest efficiency.

Also, according to the present invention, a controlling apparatus for an inverter controlled induction motor comprising a means for detecting the slip of the motor, a means for calculating the optimum slip dependent on the constant of the motor and the frequency supplied to the motor, providing the highest efficiency, a comparator for comparing the optimum slip and the slip of the motor, and a calculation means for calculating a voltage manipulated variable from the result of the comparison, wherein the voltage manipulated variable is used to form an output voltage instruction or a magnetic flux instruction for running the motor at the slip which provides the





highest efficiency.

Further, the control apparatus for an induction motor is characterized in that the means for calculating an optimum slip, calculates an optimum slip S_f for a supplied frequency by calculating the product of the optimum slip $S \eta_{MAX}$, determined from the motor constant and a specific frequency f_0 , and a ratio $(f_0/f)^{1/2}$ when the supplied frequency is frequency f , and characterized in that the means for calculating a voltage manipulated variable by calculating the product of the result of the comparison and the gain coefficient K dependent on the load torque of the motor and optimum slip, and that the gain coefficient K increases in proportion to an exponential function of the motor torque and increases in inverse proportion to the optimum slip $S \eta_{MAX}$. Said means for calculating a voltage manipulated variable, calculates the voltage manipulated variable by calculating the product of the result of the comparison and gain coefficient K dependent on the load torque of said induction motor and the optimum slip.

Further, the control apparatus for an induction motor is characterized in that the optimum slip $S \eta_{MAX}$ is represented by

$$S \eta_{MAX} = r_2 / L_m [r_m / 2 / \pi / (r_1 + r_2)]^{1/2} / (f_0)^{1/2}$$

where r_1 is the primary resistance of the motor,

r_2 is the secondary resistance of the motor,

L_m is the excitation inductance of the motor,

r_m is the iron loss resistance of the motor, and

f is the supply frequency to the motor.

According to the present invention, The slip is not controlled indirectly to obtain the highest efficiency as in the conventional techniques and the point of the lowest power is not searched out while slightly varying the inverter output voltage. The slip of the induction motor is detected directly and the detected slip value and an optimum slip value dependent on the constant of the motor and the running frequency which provides the highest efficiency are compared. Then a voltage manipulated variable based on the result of the comparison is used to form a magnetic flux instruction, whereby the slip is controlled rapidly to obtain the highest efficiency even if load conditions change or a given frequency varies.

Brief Description of the Drawings

Fig. 1 is an L-type equivalent circuit of an induction motor;

Fig. 2 is a block diagram showing a circuit construction in an embodiment of the present invention; and

Fig. 3 is a graph illustrating a relationship between optimum slip S_f and gain coefficient K where a load factor is used as a parameter.

Best Mode for Carrying out the Invention

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In the following, the best mode for utilizing out the invention is described with reference to the drawings.

Fig. 1 shows an L-type equivalent circuit of an induction motor of an embodiment of the present invention.

Referring to Fig. 1, reference symbol V_1 represents a phase voltage, I_1 a primary current, r_1 a primary resistance, r_2 a secondary resistance, x a sum of primary and secondary reactances, g_0 an excitation conductance, b_0 an excitation susceptance, and S a motor slip. Where a complex symbol is represented by j , from the equivalent circuit shown, the following expressions are formed:

$$\text{Vector } I_1 = \{1/(r_2/S + r_1 + jx) + g_0 - jb_0\}V_1 \quad (1)$$

$$\text{Vector } I_2 = V_1/(r_2/S + r_1 + jx) \quad (2)$$

$$\text{Vector } I_N = (g_0 - jb_0)V_1 \quad (3)$$

Primary input complex power vector Pa_1 is given, where the number of phase of the power supply is m , by

$$\begin{aligned} \text{Vector } Pa_1 &= m \times \text{vector } I_1 \times V_1 \\ &= m[(r_2/S + r_1)/((r_2/S + r_1)^2 + x^2) + g_0 \\ &\quad - j\{x/((r_2/S + r_2)^2 + x^2) + b_0\}]V_1^2 \end{aligned} \quad (4)$$

Since the primary input power is given by the real number part of the expression (4),

$$\begin{aligned} \text{Primary input power } P_1 &= \{r_2/S + r_1\}/((r_2/S + r_1)^2 + x^2) \\ &\quad + g_0\}mV_1^2 \end{aligned} \quad (5)$$

and secondary complex power vector Pa_2 is given by

$$\begin{aligned} \text{Vector } Pa_2 &= m(\text{vector } I_2V_1 - I_2^2r_1) \\ &= m[(r_2/S)/((r_2/S + r_1)^2 + x^2) \end{aligned}$$

$$- j\{x/((r_2/S + r_1)^2 + x^2)\}V_1^2 \quad (6)$$

Since the secondary input power is given by the real number part of the expression (6),

$$\text{Secondary input power } P_2 = \{r_2/S/((r_2/S + r_1)^2 + x^2)\}mV_1^2 \quad (7)$$

$$\text{Secondary loss } P_{21} = mI_2^2 r_2 = r_2/((r_2/S + r_1)^2 + x^2)mV_1^2 \quad (8)$$

$$\begin{aligned} \text{Mechanical loss } P_0 &= P_2 - P_{21} \\ &= r_2(1 - S)/S/((r_2/S + r_1)^2 + x^2)mV_1^2 \end{aligned} \quad (9)$$

$$\begin{aligned} \text{Efficiency } \eta &= P_0/P_1 \\ &= r_2(1 - S)/S/\{r_2/S + r_1 + g_0((r_2/S + r_1)^2 + x^2)\} \end{aligned} \quad (10)$$

The slip that provides the highest efficiency is obtained by differentiating efficiency η of expression (10) with motor slip S until the value of S becomes zero. Accordingly, the slip $S \eta_{MAX}$ which provides the highest efficiency is given by

$$\begin{aligned} S \eta_{MAX} &= [r_2\{g_0^2\{(r_1 + r_2)^2 + x^2\} + g_0(r_1 + r_2)\}^{1/2} - \\ &\quad g_0r_2^2] / [g_0\{(r_1 + r_2)^2 + x^2 - r_2^2\} + r_1 + r_2] \end{aligned} \quad (11)$$

Since expression (11) is a function only of motor constants and frequency, it is not influenced by the load or the input voltage if only the frequency is determined. Accordingly, if the motor is controlled so that optimum slip $S \eta_{MAX}$ is maintained, it can be run at the highest

efficiency.

However, in order to calculate the optimum slip from expression (11), the constants of the motor must be known. Further, even if they are known, since a large amount of calculation is required, to perform the calculation for each disadvantageous scanning imposes an excessively high burden on a CPU and is not practical. Therefore, approximation as described below is used.

Since, in expression (11), $g_0\{(r_1 + r_2)^2 + x^2\}$ and $g_0\{(r_1 + r_2)^2 + x^2 - r_2^2\}$ are exceedingly small compared with $(r_1 + r_2)$ at a frequency that is higher than 1 Hz, they can be disregarded and expression (11) written as

$$S \eta_{MAX} = [r_2\{g_0(r_1 + r_2)\}^{1/2} - g_0 r_2^2] / (r_1 + r_2) \quad (12)$$

Further, since $r_2\{g_0(r_1 + r_2)\}^{1/2}$ in the numerator of expression (12) is exceedingly large compared with $g_0 r_2^2$, it can be disregarded and expression (12) written as

$$\begin{aligned} S \eta_{MAX} &= r_2\{g_0(r_1 + r_2)\}^{1/2} \\ &= \{r_m/2/\pi/(r_1 + r_2)\}^{1/2} r_2/L_m/f^{1/2} \\ &= K_0/f^{1/2} \end{aligned} \quad (13)$$

$$(\text{where } K_0 = \{r_m/2/\pi/(r_1 + r_2)\}^{1/2} r_2/L_m)$$

In particular, it can be seen that optimum slip $S \eta_{MAX}$ increases in proportion to the inverse number of the square root of the frequency. Accordingly, if optimum slip

$S \eta_{MAX}$ when the frequency is the rated frequency f_0 and is known, then optimum slip S_1 when the frequency is arbitrary

frequency f can be calculated in accordance with

$$S_1 = S \eta_{MAX} (f_0/f)^{1/2} \quad (14)$$

In the following, an embodiment of the present invention is described with reference to a block diagram showing a circuit construction of the same. Fig. 2 shows an embodiment in which the present invention is applied to vector control.

Referring to Fig. 2, reference numerals 1 and 5 each denote a divider, 2 denotes a square root calculator, 3 a multiplier, 4 a frequency-synchronous speed converter, 6 a gain coefficient multiplier, and 7 a filter. Further, reference symbol ϕ_{ref} denotes a magnetic flux instruction (prior to variation), and ϕ_{out} denotes a magnetic flux instruction after variation. Reference symbol $S \eta_{MAX}$ denotes a slip which provides the highest efficiency at a rated frequency. Recent vector control involves an automatic tuning function, and no significant difference occurs even if a rated slip specified by automatic tuning is used as slip $S \eta_{MAX}$ which provides the highest efficiency. Further, reference symbol f_n denotes a rated frequency, f a running frequency, and N_{fb} a speed feedback value.

Control in the present embodiment is described.

Based on expression (14), frequency f_n is divided by running frequency f at divider 1 and the value of the square root of the result of the division determined by

square root calculator 2 is multiplied at multiplier 3 by the slip $S_{\eta \text{ MAX}}$ which provides the highest efficiency for a rated frequency to determine optimum slip S_f for a running frequency.

Meanwhile, running frequency f is multiplied by $120/P$ (the number of poles of the motor) at frequency-synchronous speed converter 4 to determine synchronous rotational frequency N_f , and the value obtained by subtracting speed feedback value N_{fb} from synchronous rotational frequency N_f is divided by synchronous frequency N_f at divider 5 to determine slip S_{det} of the motor. The value obtained by subtracting this slip S_{det} of the motor from optimum slip S_f is multiplied by gain coefficient K at multiplier 6, and the result of the multiplication is passed through a filter for stabilization of control, whereafter it is subtracted from magnetic flux instruction ϕ_{ref} to obtain magnetic flux instruction ϕ_{out} .

When slip S_{det} is smaller than optimum slip S_f , since the magnetic flux is small, control is performed by varying the output voltage to make slip S_{det} approach optimum slip S_f . In this manner, the motor is run at the highest efficiency.

Gain coefficient K is determined in the following manner.

When there is no energy saving control, with output voltage V_1 , slip S_1 , highest motor efficiency voltage V_2 ,

and the slip by S_2 , gain coefficient K can be calculated by

$$K = (V_1 - V_2)(S_2 - S_1)$$

Since V_1 , V_2 , S_1 and S_2 vary with the optimum slip and the load in rated running, the value of K must be changed accordingly.

When the value of coefficient K with respect to the optimum slip and the load factor was calculated with regard to motors of 0.4 to 55 kW and the relationship between them was investigated, the value of coefficient K increased in inverse proportion to the optimum slip, but had an exponential relationship to the load factor. When this is shown by graph, such as shown in Fig.3 coefficient K is obtained. In Fig. 3, the axis of abscissa indicates the optimum slip (%), the axis of ordinate the gain coefficient K , and parameters of the load factor (%).

In this instance, the value of gain coefficient K is

$$K = 15480 \exp(-0.696 T_L) / S \eta_{\text{MAX}} \quad T_L: \text{load torque}$$

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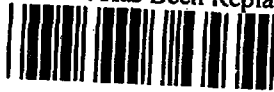
Claims

1. A controlling method for an inverter controlled induction motor wherein the slip of said induction motor is detected and the optimum slip dependent on the constant of said induction motor and the frequency supplied to said induction motor providing the highest efficiency is calculated, the optimum slip and the slip of said induction motor are compared to determine a voltage manipulated variable which is used to form an output voltage instruction or a magnetic flux instruction for running said induction motor at the slip which provides the highest efficiency.

2. A controlling apparatus for an inverter controlled induction motor comprising a means for detecting the slip of said induction motor, a means for calculating the optimum slip dependent on the constant of said induction motor and the frequency supplied to said induction motor providing the highest efficiency, a comparator for comparing the optimum slip and the slip of said induction motor, and a calculation means for calculating a voltage manipulated variable from the result of the comparison, wherein said voltage manipulated variable is used to form an output voltage instruction or a magnetic flux instruction for running said induction motor at the slip which provides the highest efficiency.

3. A control apparatus for the induction motor of claim 2,

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characterized in that said means for calculating an optimum slip calculates the optimum slip S_f for a supplied frequency by calculating the product of optimum slip $S \eta_{MAX}$, determined from the motor constant and a specific frequency f_0 , and a ratio $(f_0/f)^{1/2}$ when the supplied frequency is frequency f .

4. A controlling apparatus for the induction motor of claim 2, characterized in that the said means for calculating a voltage manipulated variable calculates a voltage manipulated variable by calculating the product of the result of the comparison and gain coefficient K dependent on the load torque of said induction motor and the optimum slip.

5. A controlling apparatus for the induction motor of claim 2, characterized in that gain coefficient K increases in proportion to the exponential function of the motor torque and increases in inverse proportion to the optimum slip $S \eta_{MAX}$, and said means for calculating a voltage manipulated variable calculates the voltage manipulated variable by calculating the product of the result of the comparison and gain coefficient K dependent on the load torque of said induction motor and the optimum slip.

6. A control apparatus for the induction motor of claim 2, characterized in that the optimum slip $S \eta_{MAX}$ is represented by

$$S \eta_{MAX} = r_2 / L_m [r_m / 2 / \pi / (r_1 + r_2)]^{1/2} / (f_0)^{1/2}$$

where r_1 is the primary resistance of said induction motor,

r_2 is the secondary resistance of said induction motor,

L_m is the excitation inductance of said induction motor,

r_m is the iron loss resistance of said induction motor, and

f is the supply frequency to said induction motor.

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