Description

The invention relates to a method of driving a display device comprising an electro-optical display medium disposed between first and second supporting plates and including row and column electrodes said device including a plurality of pairs of opposing first and second pixel electrodes arranged on the supporting plates, each of said pairs of electrodes defining a respective pixel in the medium and being electrically connected to respective ones of the row and column electrodes for applying selection and data voltages to select and drive, respectively, selected ones of the pixels to predefined states of transmissivity.

The invention further relates to a device for using such a method.

In this respect it is to be noted that the term row electrode and column electrode in this application may be interchanged if desired, so that a column electrode may be meant where reference is made to a row electrode while simultaneously changing column electrode to row electrode.

In EP-A-0 299 546, which document constitutes a prior art publication in accordance with Article 54(3), EPC, a method of the type mentioned in the opening paragraph is described, which provides a wide choice of freedom in the colour filters to be used. This is possible by giving the pixels a given adjustment per row by charging or discharging the capacitance associated with these pixels after first having discharged or charged them too far (either or not accurately).

In the said Application this is realised by applying, prior to selection, an auxiliary voltage across the pixels beyond or on the limit of the voltage range to be used for picture display, for example an auxiliary voltage (reference voltage) or reset voltage.

In a preferred embodiment of the device described in this Application a zener diode is arranged between a pixel and a row or column electrode.

Such a zener diode has a strong asymmetrical current-voltage behaviour (IV-curve).

A method according to the invention is characterized in that for each of said selected pixels the application of said selection and drive voltages effect, in sequence:

a. charging of the pixel to an auxiliary voltage beyond or on the limit of the voltage range to be used for picture display through a respective, substantially nonlinear symmetrical switching element which is electrically connected to one of the electrodes of the pixel and

b. charging of the pixel through the substantially non-linear symmetrical switching element from the auxiliary voltage to a voltage of the same sign but of lesser magnitude at which said predefined state of transmissivity is effected, the charging to the auxiliary voltage and the charging to the voltage of lesser magnitude both being effected via the same current path.

Using a symmetrical switching element for selection of a picture element is known per se from FR-A-2 533 730 in which a diode ring is used as a switching element. A way of driving is shown in which during non-selection voltages are presented to the row electrodes which have such a value that voltage variations at the column electrodes do not effect the voltage at non-selected pixels.

In this way of driving selection occurs immediately after non-selection; since the charge on a pixel is not exactly known (it is determined in a previous selection period and may have been influenced by e.g. leakage currents) a certain inaccuracy of the voltage on the pixel may occur during selection.

By charging the pixel to an auxiliary voltage beyond or on the limit of the voltage range to be used for picture display prior to selection however the pixel is always charged during selection from said auxiliary voltage to a value determined by the data signals.

In this connection at least approximately symmetrical switching elements are understood to mean switching elements which have the same or approximately the same current-voltage variation (with opposite sign for current and voltage) such as, for example a MIM (metal-isolator-metal) when reversing the voltage. A completely identical variation in both directions is often made substantially impossible by the manufacturing method, in practice the current-voltage curves of such symmetrical switching elements have substantially the same shape in both directions and the on and off voltages in the positive part of the characteristic curve (with the exception of the sign) differ only little from those in the negative part, in contrast to, for example zener diodes. Other examples of symmetrical switching elements are, for example back-to-back diodes and certain semiconductor switching elements such as a nin switching element or a pip switching element (alternately comprising an n (p)-doped semiconductor region, a substantially intrinsic semiconductor region and an n (p)-doped semiconductor region).

The symmetrical behaviour can also be obtained by means of a combination of sub-switching elements such as one or more diode rings, a combination of the above-mentioned switching elements, or otherwise.

Viewed over the entire display device, the said switching elements may have a considerable spread in, for example the forward voltage so that unwanted voltage components may be introduced resulting in non-uniformities (grey variations) occurring in the display device when conventionally driving the rows with periodical inversion of the polarity of both the selection signals and the non-selection signals (simultaneously with that of the data signals). It is found that these voltage components can be compensated for when used in the method according to the invention in such a way that they hardly influence or do not in-

fluence the voltage determining the transmission of the liquid crystal.

Embodiments of the invention will now be described in greater detail with reference to the accompanying drawings in which:

Fig. 1 shows diagrammatically a picture display device for use of a method according to the invention,

Fig. 2 shows the transmission/voltage characteristic of an electro-optical medium, for example a liquid crystal,

Fig. 3 shows diagrammatically the current-voltage characteristic of an approximately symmetrically non-linear switching element (for example a MIM),

Fig. 4 shows the drive signals associated with a method according to the invention and

Fig. 5 shows a modification.

Fig. 1a shows diagrammatically a device for use of a method according to the invention. Pixels 12 arranged in a matrix are located at the area of crossings of row electrodes 11 and column electrodes 8, while the pixels are connected to the column electrodes 8 via symmetrical non-linear switching elements 15, in this example MIMs.

If a data voltage Vd is presented to a column electrode 8, while a selection voltage Vs1 is presented to a selected row electrode 11, it holds for a selected pixel 12 that the voltage across this pixel, i.e. the pixel voltage Vp1 (see Fig. 2) is equal to:

$$Vp1 = Vd - Vs1 - Vm \quad (1)$$

in which Vm is the forward voltage of the MIM which supplies sufficient current to charge the pixel to the correct voltage within the desired period of time.

In a subsequent field the data voltage is presented in an inverted manner (-Vd) while the selection voltage is now Vs2. Since, as will be described hereinafter, the capacitance associated with the pixel 12 has first been negatively charged too far in a manner analogous to that described in the non-prepublished EP-A- 0 299 546, it is charged again while the current through the MIM has the same direction so that the pixel voltage Vp2 (see Fig. 2) is now equal to:

$$Vp2 = -Vd - Vs2 - Vm \quad (2)$$

It follows from (1) and (2) that:

$$Vp1 - Vp2 = 2Vd - Vs1 + Vs2 \equiv 2\,Vampl \quad (3)$$

$$Vp1 + Vp2 = -Vs1 - Vs2 - 2Vm \equiv 2\,V_{DC} \quad (4)$$

In the ideal case (no spread in the voltage Vm, completely symmetrical transmission/voltage characteristics) the pixel voltage in the case of equal but opposite data voltage Vd and -Vd is also equal but has an opposite sign, provided that it holds for the selection voltages Vs1, Vs2 that Vs2 = -Vs1-2Vm. Then it holds that Vp1 = -Vp2 = Vampl. Simultaneously V_{DC} = 0.

The same pixel voltages Vp1, Vp2 can also be written as:

$$Vp1 = 1/2(Vp1 + Vp2) + 1/2(Vp1 - Vp2) = V_{DC} + Vampl \quad (5)$$

$$Vp2 = 1/2(Vp1 + Vp2) - 1/2(Vp1 - Vp2) = V_{DC} - Vampl \quad (6)$$

The DC component V_{DC} which may be introduced by the fact that the MIM voltage Vm (or that of another approximately symmetrical switching element) is not identical throughout the surface of the display device and which results in a deviation of the voltage drop across an arbitrary MIM from the nominal value Vm appears to be compensated in practice by a movement of ions in the liquid crystal material so that after some time a direct voltage is only present over the insulating (orientation) layer covering the electrodes. The effective pixel voltage V*p is now determined by the (periodically changing) voltage Vmpl. For this it holds that:

$$V*p1 = -V*p2 = Vampl \quad \text{(see Fig. 2)}$$

Even if the said compensation does not occur, the effective voltage $Veff = V^2ampl + V^2_{DC}$ deviates to a small extent from the desired value Vampl across two fields at an average, provided that $|V_{DC}| \ll |Vampl|$, which can be quite well realised in practice. To inhibit possible disintegration of the liquid crystal due to this small remaining DC voltage, the sign of all operating voltages may be changed periodically.

The voltage Vampl is independent of the voltage drop across the MIM and possible variations therein. Variations due to non-uniform switching behaviour of the switching elements are therefore not found or are hardly found in the transmission behaviour of the device because possible DC components are compensated for. These DC components are independent of the data voltages (see (4)) so that no image retention or ghost pictures occur.

For writing information a first selection voltage Vs1 is presented on a selection line 11 during a selection period t_s while the information or data voltages V_d are simultaneously presented on the column electrodes 8; this leads to, for example a positive voltage across a pixel 12 which represents the information presented.

To prevent degradation of the liquid crystal and to be able to increase the so-called face flicker frequency, information having an alternating sign is preferably presented across the pixel 12. In the method according to the invention a negative voltage across the pixel 12, which represents the information presented, is achieved by presenting a second selection voltage Vs2 while simultaneously presenting inverted data voltages (-Vd) after having discharged the capacitance associated with the pixel 12 too far (or after having negatively charged it too far) via the MIM 15.

Fig. 4 shows how the drive signals are chosen for a plurality of rows of pixels 12 in order to write them with picture information which changes sign during each field (for example in TV applications).

From the instant t0 (see Fig. 4a) a selection voltage Vs1 is presented on a row electrode 11 during a

selection period ts (which in this example is chosen to be equal to a line period for TV applications, namely 64 µsec) while information voltages or data voltages Vd are simultaneously presented on the column electrodes 8. After the instant t1 the associated row of pixels 12 is no longer selected because the row electrode 11 receives a voltage Vns1. This voltage is maintained until just before the next selection of the row of pixels 12. In this example this is effected by presenting a reset voltage on the relevant row electrode 11 just before again selecting the first row of pixels 12, namely at an instant t3 = tf-ts, in which tf represents a field period. The reset voltage can then be chosen to be such that the pixels 12 are charged negatively to such an extent via the MIM 15 that the voltage across each associated pixel lies beyond the range to be used for picture display (up to a value of ≦-Vsat). In a subsequent selection period (from t4) they are then charged to the desired value determined by data voltages -Vd, via the MIM. To this end the row electrodes receive the voltage Vs2 and after the selection period (after t5) has elapsed, they receive a non-selection voltage Vns2. In this way the voltage across the pixels is inverted during each field period.

Fig. 4b shows the same voltage variation as Fig. 4a but is then shifted over a field period plus a selection period (in this case a line period). This provides the possibility of writing two successive rows of pixels with inverse data voltages with respect to each other. Fig. 4c is identical to Fig. 4a, but is shifted over two selection periods.

For (television) pictures with half the vertical resolution in which the lines of the even and the odd field are written over each other, it is achieved that the picture information changes its sign and is refreshed once per field period. Although the line flicker frequency is 25 Hz (30 Hz) in this case, a face flicker frequency of 50 hz (60 Hz) is achieved between successive rows due to the phase difference of 180° introduced by changing the sign per row.

The selection voltages Vs1 and Vs2 may of course also be chosen to be shorter than one line period (64 µsec). In this case the reset voltage may alternatively be presented during a part of the line period in which selection takes place, provided there is sufficient time left to charge the pixels 12. The voltage variation on the electrode 11 is then effected, for example in the way as is shown diagrammatically in Fig. 4a by means of the broken line 14.

The device shown is very suitable for using a drive method in which

$$Vc = \frac{Vsat + Vth}{2}$$

and

$$-Vc = -\left(\frac{Vsat + Vth}{2}\right)$$

are chosen for the average voltage across a pixel (see Fig. 2) so that the absolute value of the voltage for the purpose of picture display across the pixels 12 is substantially limited to the range between Vth and Vsat.

A satisfactory operation as regards grey scales is obtained if, dependent on the data voltages Vd on the column electrodes 8, the voltage values across the pixels 12 are at most Vc+Vdmax = Vsat and at least Vc-Vdmax = Vth. Elimination of Vc yield:

$$|Vdmax| = 1/2(Vsat - Vth), \text{ i.e.:} \quad (1)$$

$$-1/2(Vsat - Vth) \leqq Vdmax \leqq 1/2(VSat - Vth). \quad (2)$$

In order to charge a row of pixels 12, for example, positively, the associated row electrode 11 is given a selection voltage Vs1 = -V1-1/2(Vsat+Vth) in which V1 (Fig. 3) is the forward voltage of the MIM 15. The voltage across the pixel 12 is therefore Vd-V1-Vs1; it ranges between

$$-1/2(Vsat + Vth) + 1/2(Vsat + Vth) = Vth \quad (3)$$

and

$$1/2(Vsat - Vth) + 1/2(Vsat + Vth) = V_{sat}, \quad (4)$$

dependent on Vd.

In order to negatively charge the same row of pixels 12 (in a subsequent field or frame period) at a subsequent selection with inverted data voltages, these are first charged negatively too far by means of a reset voltage Vreset on the row electrode 11. Subsequently the selected row electrode receives a selection voltage

Vs2 = -V1+1/2(Vsat+Vth) (in the same line period or in a subsequent period). The pixels 12 which are negatively charged too far are now charged via the MIM 15 to -Vd-V1-Vs2, that is to say, to values between:

$$1/2(Vsat - Vth) - 1/2(Vsat - Vth) = -Vth \quad (5)$$

and

$$-1/2(Vsat - Vth) - 1/2(Vsat - Vth) = -Vsat \quad (6)$$

so that information with the opposite sign is presented across the pixels 12.

In the case of non-selection the requirement must be satisfied that the MIMs 15 cannot conduct, or convey such a low current Ioff that discharge via the MIMs 15 is substantially negligible.

For a lowest non-selection voltage Vns1 it holds that the voltage V_A at the junction 16 ranges between the values:

$$V_{Amin} = Vns1 + Vth \quad (7)$$

and

$$V_{Amax} = Vns1 + Vsat \quad (8)$$

For the values V_{Amin} and V_{Amax} it holds that:

$$V_{Amax} \leqq -Vdmax + V3 \quad (9)$$

in which -V3 (Fig. 3) is the voltage at which charging via the MIM is substantially negligible.

To prevent discharge via the MIM 15, it also holds that:

$$V_{Amin} \geqq Vdmax - V2 \quad (10)$$

in which V2 (Fig. 3) is the voltage at which the current in the other direction is substantially negligible.

The equations (7) and (10) lead to:

$$+Vdmax - V2 \leqq Vns1 + Vth$$

or (with Vdmax = 1/2(Vsat-Vth))

$$Vns1 \geqq +1/2(Vsat - Vth) - V2 - Vth \quad (11)$$

The equations (8) and (9) similarly lead to:

$$-Vdmax + V3 \geqq Vns1 + Vsat$$

or

$$Vns1 \leqq -1/2(Vsat + Vth) - Vsat + V3 \quad (12)$$

Combination of (11) and (12) leads to:

$$-1/2(Vsat - Vth) - Vsat + V3 \geqq Vns\,1 > 1/2(Vsat - Vth) - V2 - Vth \quad (13)$$

or

$$V2 + V3 \geqq 2(Vsat - Vth) \quad (14)$$

The pixels 12 are subsequently discharged to a value ≦-Vsat (see Fig. 4a) by giving the row electrode 11 a sufficiently high reset voltage. For this it holds that:

$$Vreset \geqq Vdmax + Vsat + V4$$

in which -V4 (Fig. 3) is the voltage of the MIM 15 in the other direction at which sufficient conductance occurs, or

$$Vreset \geqq 1/2(Vsat - Vth) + Vsat + V4 \quad (5)$$

Subsequently the pixels are accurately charged via the MIMs 15. For this purpose a selection voltage $V_{S2} = 1/2(Vsat+Vth)-V1$ is presented on the row electrode 11 while simultaneously presenting data voltages on the column electrodes 8.

Subsequently the row electrode 11 receives a non-selection voltage Vns2. For the voltages at the junction 16 it now holds that:

$V_{Amax} = Vns2 - Vth$ and $V_{Amin} = Vns2 - Vsat$

With the equations (3) and (4) and Vdmax = 1/2(Vsat-Vth) this leads to:

$$Vns2 \leqq -Vdmax + V3 + Vth = -1/2(Vsat - Vth) + Vth + V3 \quad (16)$$

and

$$Vns2 \geqq Vdmax - V2 + Vsat = +1/2(Vsat - Vth) - V2 + Vsat \quad (17)$$

Combination of (16) and (17) leads to:

$$1/2(Vsat - Vth) - V2 + V_{sat} \leqq Vns2 \leqq -1/2(Vsat - Vth) + Vth + V3 \quad (18)$$

so that it holds again that:

$$V2 + V3 \geqq 2(Vsat - Vth) \quad (14')$$

To limit the number of drive levels, it may be desirable to use only one non-selection voltage Vns.

Combination of the equations (7) and (12) then yields

$$1/2(Vsat - Vth) + Vsat - V2 \leqq Vns \leqq -1/2(Vsat - Vth) - Vsat + V3 \quad (19)$$

In this case it holds that:

$$V2 + V3 \geqq 3Vsat - Vth \quad (20)$$

The invention is of course not limited to the example described hereinbefore, but it may alternatively be used in devices comprising different non-linear approximately symmetrical switching elements such as a back-to-back diode, a nin or a pip switching element.

The switching element and the display element may also exchange positions as is shown diagrammmatically in Fig. 1b.

A non-linear approximately symmetrical switching element may alternatively be assembled from different sub-switching elements as in the case of one or more diode rings or when providing redundancy, in a manner similar to that described in EP-A- 0 296 663.

As already stated in the opening paragraph, possible degradation of the liquid crystal due to a remaining DC voltage can be avoided by periodically changing the sign of the operating voltages. To reduce possible other detrimental effects of unilateral resetting, with the non-linear switching element always conducting in one direction (for example, electron migration in Schottky diodes), it may be advantageous to periodically change the sign of all operating voltages, for example after each frame or after a fixed number of frames. This is shown diagrammatically in Fig. 5 in which firstly the drive is effected with the drive voltages during n frame period t_f (n ≧ 1), as derived hereinbefore (with one non-selection voltage Vns) and subsequently with inverse drive voltages during m frame periods (m ≧ 1). The inversion need not take place at periodical instants. For practical considerations a periodical inversion will generally be preferred, where n and m are at least 10.

The row electrodes need not be connected directly to the picture electrodes but they may be capacitively coupled thereto, as has been described in greater detail in the non-prepublished EP-A- 0 357 147 in the name of the Applicant.

An extra (storage) capacitance may be arranged parallel to the pixel between the column electrode and the row electrode.

Moreover, prior to selection at t_o in Figure 4[a] a second reset voltage can be presented to the row electrode 11 which in a manner known per se charges the pixels to an auxiliary voltage of a sign opposite to the voltage sign obtained during the selecting period following immediately thereafter. To obtain uniform charging or discharging it may be advantageous to keep the data signals at zero Volt both during reset as described with reference to Figure 4 and during charging or discharging by means of said second reset voltage.

Claims

1. A method of driving a display device comprising an electro-optical display medium disposed between first and second supporting plates and including row and column electrodes said device including a plurality of pairs of opposing first and second pixel electrodes arranged on the supporting plates, each of said pairs of electrodes defining a respective pixel (12) in the medium and being electrically connected to respective ones of the row (11) and column (8) electrodes for applying selection and data voltages to select and drive, respectively, selected ones of the pixels to predefined states of transmissivity, character-

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The equations (8) and (9) similarly lead to:

$$-Vdmax + V3 \geqq Vns1 + Vsat$$

or

$$Vns1 \leqq -1/2(Vsat + Vth) - Vsat + V3 \quad (12)$$

Combination of (11) and (12) leads to:

$$-1/2(Vsat - Vth) - Vsat + V3 \geqq Vns\,1 > 1/2(Vsat - Vth) - V2 - Vth \quad (13)$$

or

$$V2 + V3 \geqq 2(Vsat - Vth) \quad (14)$$

The pixels 12 are subsequently discharged to a value ≦-Vsat (see Fig. 4a) by giving the row electrode 11 a sufficiently high reset voltage. For this it holds that:

$$Vreset \geqq Vdmax + Vsat + V4$$

in which -V4 (Fig. 3) is the voltage of the MIM 15 in the other direction at which sufficient conductance occurs, or

$$Vreset \geqq 1/2(Vsat - Vth) + Vsat + V4 \quad (5)$$

Subsequently the pixels are accurately charged via the MIMs 15. For this purpose a selection voltage $V_{S2} = 1/2(Vsat+Vth)-V1$ is presented on the row electrode 11 while simultaneously presenting data voltages on the column electrodes 8.

Subsequently the row electrode 11 receives a non-selection voltage Vns2. For the voltages at the junction 16 it now holds that:

$V_{Amax} = Vns2 - Vth$ and $V_{Amin} = Vns2 - Vsat$

With the equations (3) and (4) and Vdmax = 1/2(Vsat-Vth) this leads to:

$$Vns2 \leqq -Vdmax + V3 + Vth = -1/2(Vsat - Vth) + Vth + V3 \quad (16)$$

and

$$Vns2 \geqq Vdmax - V2 + Vsat = +1/2(Vsat - Vth) - V2 + Vsat \quad (17)$$

Combination of (16) and (17) leads to:

$$1/2(Vsat - Vth) - V2 + V_{sat} \leqq Vns2 \leqq -1/2(Vsat - Vth) + Vth + V3 \quad (18)$$

so that it holds again that:

$$V2 + V3 \geqq 2(Vsat - Vth) \quad (14')$$

To limit the number of drive levels, it may be desirable to use only one non-selection voltage Vns.

Combination of the equations (7) and (12) then yields

$$1/2(Vsat - Vth) + Vsat - V2 \leqq Vns \leqq -1/2(Vsat - Vth) - Vsat + V3 \quad (19)$$

In this case it holds that:

$$V2 + V3 \geqq 3Vsat - Vth \quad (20)$$

The invention is of course not limited to the example described hereinbefore, but it may alternatively be used in devices comprising different non-linear approximately symmetrical switching elements such as a back-to-back diode, a nin or a pip switching element.

The switching element and the display element may also exchange positions as is shown diagrammmatically in Fig. 1b.

A non-linear approximately symmetrical switching element may alternatively be assembled from different sub-switching elements as in the case of one or more diode rings or when providing redundancy, in a manner similar to that described in EP-A- 0 296 663.

As already stated in the opening paragraph, possible degradation of the liquid crystal due to a remaining DC voltage can be avoided by periodically changing the sign of the operating voltages. To reduce possible other detrimental effects of unilateral resetting, with the non-linear switching element always conducting in one direction (for example, electron migration in Schottky diodes), it may be advantageous to periodically change the sign of all operating voltages, for example after each frame or after a fixed number of frames. This is shown diagrammatically in Fig. 5 in which firstly the drive is effected with the drive voltages during n frame period t_f ($n \geqq 1$), as derived hereinbefore (with one non-selection voltage Vns) and subsequently with inverse drive voltages during m frame periods ($m \geqq 1$). The inversion need not take place at periodical instants. For practical considerations a periodical inversion will generally be preferred, where n and m are at least 10.

The row electrodes need not be connected directly to the picture electrodes but they may be capacitively coupled thereto, as has been described in greater detail in the non-prepublished EP-A- 0 357 147 in the name of the Applicant.

An extra (storage) capacitance may be arranged parallel to the pixel between the column electrode and the row electrode.

Moreover, prior to selection at t_o in Figure 4[a] a second reset voltage can be presented to the row electrode 11 which in a manner known per se charges the pixels to an auxiliary voltage of a sign opposite to the voltage sign obtained during the selecting period following immediately thereafter. To obtain uniform charging or discharging it may be advantageous to keep the data signals at zero Volt both during reset as described with reference to Figure 4 and during charging or discharging by means of said second reset voltage.

Claims

1. A method of driving a display device comprising an electro-optical display medium disposed between first and second supporting plates and including row and column electrodes said device including a plurality of pairs of opposing first and second pixel electrodes arranged on the supporting plates, each of said pairs of electrodes defining a respective pixel (12) in the medium and being electrically connected to respective ones of the row (11) and column (8) electrodes for applying selection and data voltages to select and drive, respectively, selected ones of the pixels to predefined states of transmissivity, character-

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ized in that for each of said selected pixels the application of said selection and drive voltages effects, in sequence:

a. charging of the pixel (12) to an auxiliary voltage beyond or on the limit of the voltage range to be used for picture display through a respective, substantially non-linear symmetrical switching element (15) which is electrically connected to one of the electrodes of the pixel and

b. charging of the pixel (12) through the substantially symmetrical nonlinear switching element (15) from the auxiliary voltage to a voltage of the same sign but of lesser magnitude at which said predefined state of transmissivity is effected, the charging to the auxiliary voltage and the charging to the voltage of lesser magnitude both being effected via the same current path.

2. A method of driving a display device as claimed in claim 1, characterized in that the substantially non-linear symmetrical switching elements (15) belong to the group of MIM's, back to back diodes, nin and pip switching elements.

3. A method as in claim 1 or 2 where, during successive selections, each of said selected pixels is charged to predetermined voltages at which said predefined state of transmissivity is effected by passing current in the same direction through the substantially symmetrical switching element.

4. A method as claimed in Claim 1, 2 or 3 characterized in that during charging or discharging the pixels to an auxiliary voltage the data signals at the column lines are kept at zero Volt.

5. A method as claimed in Claim 1 to 4, characterized in that the auxiliary voltage is beyond or on the limit of the range of transition in the transmission/voltage characteristic of the electro-optical medium.

6. A method as claimed in any one of Claims 1 to 5, characterized in that the at least approximately symmetrical non-linear switching element comprises sub-switching elements.

7. A method as claimed in Claim 6, characterized in that the at least approximately symmetrical non-linear switching element comprises at least a diode ring.

8. A method as claimed in any one of the preceding Claims, characterized in that the polarities of the drive signals used are periodically or not periodically changed.

9. A method as claimed in Claim 8 characterized in that the polarities of the drive signals are periodically changed over a period of at least ten frame periods.

10. A method as claimed in any one of the preceding Claims, characterized in that the row electrodes are capacitively coupled to the associated picture electrodes at the area of pixels.

11. A display device comprising an electro-optical display medium between first and second supporting plates and including row and column electrodes said device including a plurality of pairs of opposing first and second pixel electrodes arranged on the supporting plates, each of said pairs of electrodes defining a respective pixel (12) in the medium and being electrically connected to respective ones of the row (11) and column (8) electrodes, the device comprising means for applying selection and data voltages to select and drive, respectively, selected ones of the pixels to predefined states of transmissivity, characterized in that for each of said selected pixels said means for applying said selection and drive voltages comprise:

means for charging of the pixel (12) to an auxiliary voltage beyond or on the limit of the voltage range to be used for picture display through a respective, substantially non-linear symmetrical switching element (15) which is electrically connected to one of the electrodes of the pixel and

means for charging of the pixel (12) through the substantially symmetrical non-linear switching element (15) from the auxiliary voltage to a voltage of the same sign but of lesser magnitude at which said predefined state of transmissivity is effected, one current path being present for both the charging to the auxiliary voltage and the charging to the voltage of lesser magnitude.

12. A display device as claimed in claim 11, characterized in that the substantially non-linear symmetrical switching elements (15) belong to the group of MIM's, back to back diodes, nin and pip switching elements.

Patentansprüche

1. Verfahren zum Steuern einer Anzeigeanordnung mit einem elektrooptischen Anzeigemittel zwischen ersten und zweiten Trägerplatten und mit Zeilen- und Spaltenelektroden, wobei die Anordnung eine Anzahl von Paaren einander gegenüberliegender erster und zweiter auf den Trägerplatten angeordneter Bildelementelektroden ent-

hält, wobei jedes Elektrodenpaar ein jeweiliges Bildelement (12) in dem Mittel definiert und elektrisch mit je einer der Zeilen- (11) und Spaltenelektroden (8) zum Zuführen von Wähl- und Datenspannungen zum Wählen bzw. Steuern ausgewählter einzelner Bildelemente in vordefinierte Reintransmissionsgradszustände verbunden sind, dadurch gekennzeichnet, daß für jedes der gewählten Bildelemente die Zufuhr der Wähl- und Steuerspannungen nacheinander folgendes verwirklicht:

a. das Aufladen des Bildelements (12)auf eine Hilfsspannung vorbei oder auf der Grenze des zu benutzenden Spannungsbereichs für Bildanzeige durch ein jeweiliges im wesentlichen nichtlineares symmetrisches Schaltelement (15), das mit einer der Elektroden des Bildelements verbunden ist, und

b. das Aufladen des Bildelements (12) über das im wesentlichen nichtlineare symmetrische Schaltelement (15) mit der Hilfsspannung auf eine Spannung mit demselben Vorzeichen, aber von geringerer Größe, bei der der vordefinierte Reintransmissionsgradszustand verwirklicht wird und das Aufladen auf die Hilfsspannung und das Aufladen auf die weniger größe Spannung beide über denselben Stromweg erfolgt.

2. Verfahren zum Steuern einer Anzeigeanordnung nach Anspruch 1, dadurch gekennzeichnet, daß die im wesentlichen nichtlinearen symmetrischen Schaltelemente (15) zur Gruppe von MIM, gegensinnig gepolten Dioden, nin- und pip-Schaltelementen gehören.

3. Verfahren nach Anspruch 1 oder 2, worin durch aufeinanderfolgende Wählvorgänge jedes der gewählten Bildelemente auf vorgegebene Spannungen aufgeladen wird, bei denen der vorgegebene Zustand des Reintransmissionsgrads durch Weiterleiten von Strom in derselben Richtung durch das im wesentlichen symmetrische Schaltelement verwirklicht wird.

4. Verfahren nach Anspruch 1, 2 oder 3, dadurch gekennzeichnet, daß beim Aufladen oder Entladen der Bildelemente auf eine Hilfsspannung die Datensignale auf den Spaltenleitungen auf Null Volt gehalten werden.

5. Verfahren nach Anspruch 1 bis 4, dadurch gekennzeichnet, daß die Hilfsspannung vorbei oder auf der Grenze des Übergangsbereichs in der Transmissions/Spannungskennlinie des elektrooptischen Mediums liegt.

6. Verfahren nach einem oder mehreren der Ansprüche 1 bis 5, dadurch gekennzeichnet, daß das wenigstens nahezu symmetrische nichtlineare Schaltelement Unterschaltelemente enthält.

7. Verfahren nach Anspruch 6, dadurch gekennzeichnet, daß das wenigstens nahezu symmetrische nichtlineare Schaltelement wenigstens einen Diodenring enthält.

8. Verfahren nach einem oder mehreren der vorangehenden Ansprüche, dadurch gekennzeichnet, daß die Polaritäten der benutzten Steuersignaleperiodisch oder nichtperiodisch geändert werden.

9. Verfahren nach Anspruch 8, dadurch gekennzeichnet, daß die Polaritäten der Steuersignale über eine Periode von wenigstens zehn Halbbildperioden periodisch geändert werden.

10. Verfahren nach einem oder mehreren der vorangehenden Ansprüche, dadurch gekennzeichnet, daß die Zeilenelektroden im Bereich der Bildelemente mit den zugeordneten Bildelektroden kapazitiv gekoppelt sind.

11. Anzeigeanordnung mit einem elektrooptischen Anzeigemittel zwischen ersten und zweiten Trägerplatten und mit Zeilen- und Spaltenelektroden, wobei die Anordnung eine Anzahl von Paaren einander gegenüberliegender erster und zweiter auf den Trägerplatten angeordneter Bildelementelektroden enthält, wobei jedes Elektrodenpaar ein jeweiliges Bildelement (12) in dem Mittel definiert und elektrisch mit je einer der Zeilen- (11) und Spaltenelektroden (8) zum Zuführen von Wähl- und Datenspannungen zum Wählen bzw. Steuern ausgewählter einzelner Bildelemente in vordefinierte Reintransmissionsgradszustände verbunden sind, dadurch gekennzeichnet, daß das Mittel für die Zufuhr der Wähl- und Steuerspannungen nacheinander für jedes der gewählten Bildelemente folgende Mittel enthältt:

ein Mittel zum Aufladen des Bildelements (12) auf eine Hilfsspannung vorbei oder auf der Grenze des zu benutzenden Spannungsbereichs für Bildanzeige durch ein jeweiliges im wesentlichen nichtlineares symmetrisches Schaltelement (15), das mit einer der Elektroden des Bildelements verbunden ist, und

ein Mittel zum Aufladen des Bildelements (12) über das im wesentlichen nichtlineare symmetrische Schaltelement (15) mit der Hilfsspannung auf eine Spannung mit demselben Vorzeichen, aber von geringerer Größe, bei der der vordefinierte Reintransmissionsgradszustand verwirklicht wird und das Aufladen auf die Hilfsspan-

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nung und das Aufladen auf die weniger größe Spannung beide über denselben Stromweg erfolgt.

12. Anzeigeanordnung nach Anspruch 11, dadurch gekennzeichnet, daß die im wesentlichen nichtlinearen symmetrischen Schaltelemente (15) zur Gruppe der MIM, der gegensinnig gepolten Dioden, und der nin- und pip-Schaltelemente gehören.

Revendications

1. Procédé de commande d'un dispositif d'affichage comprenant un milieu d'affichage électro-optique disposé entre une première plaque et une deuxième plaque de support, et comportant des électrodes de rangées et de colonnes, ledit dispositif comprenant une pluralité de paires de premières et de deuxièmes électrodes de pixels opposées agencées sur les plaques de support, chacune desdites paires d'électrodes définissant un pixel (12) respectif dans le milieu d'affichage et étant connectée électriquement à des électrodes respectives parmi les électrodes de rangées (11) et de colonnes (8) pour appliquer des tensions de sélection et de données afin de sélectionner et de commander, respectivement, des pixels sélectionnés pour les amener dans des états prédéfinis de transmissivité, caractérisé en ce que, pour chacun desdits pixels sélectionnés, l'application desdites tensions de sélection et de commande assure, dans l'ordre:
   a. la charge du pixel (12) à une tension auxiliaire au-delà ou à la limite de la plage de tensions susceptible d'être utilisée pour l'affichage des images via un élément de commutation (15) symétrique respectif sensiblement non linéaire, qui est connecté électriquement à l'une des électrodes du pixel, et
   b. la charge du pixel (12) via l'élément de commutation symétrique sensiblement non linéaire (15) de la tension auxiliaire à une tension de même signe, mais de moindre grandeur à laquelle ledit état prédéfini de transmissivité est effectué, la charge à la tension auxiliaire et la charge à la tension de moindre grandeur étant toutes deux effectuées par le même chemin de courant.

2. Procédé de commande d'un dispositif d'affichage selon la revendication 1, caractérisé en ce que les éléments de commutation symétriques sensiblement non linéaires (15) appartiennent au groupe des MIM, des diodes montées tête-bêche, des éléments de commutation nin et pip.

3. Procédé selon la revendication 1 ou 2, dans lequel, au cours de sélections successives, chacun desdits pixels sélectionnés est chargé à des tensions prédéterminées auxquelles ledit état prédéfini de transmissivité est effectué en faisant passer un courant dans le même sens à travers l'élément de commutation sensiblement symétrique.

4. Procédé selon la revendication 1, 2 ou 3, caractérisé en ce que, au cours de la charge ou de la décharge des pixels à une tension auxiliaire, les signaux de données sur les lignes de colonnes sont maintenus à zéro volt.

5. Procédé selon les revendications 1 à 4, caractérisé en ce que la tension auxiliaire est au-delà ou est à la limite de la plage de transition de la caractéristique de transmission/tension du milieu électro-optique.

6. Procédé selon l'une quelconque des revendications 1 à 5, caractérisé en ce que l'élément de commutation non linéaire au moins approximativement symétrique comprend des sous-éléments de commutation.

7. Procédé selon la revendication 6, caractérisé en ce qu'au moins l'élément de commutation non linéaire au moins approximativement symétrique comprend au moins un anneau de diodes.

8. Procédé selon l'une quelconque des revendications précédentes, caractérisé en ce que les polarités des signaux de commande utilisés sont modifiées périodiquement ou non périodiquement.

9. Procédé selon la revendication 8, caractérisé en ce que les polarités des signaux de commande sont modifiées périodiquement sur une période d'au moins dix périodes d'image.

10. Procédé selon l'une quelconque des revendications précédentes, caractérisé en ce que les électrodes de rangées sont couplées par voie capacitive aux électrodes d'images associées dans la zone des pixels.

11. Dispositif d'affichage comprenant un milieu d'affichage électro-optique entre une première plaque et une deuxième plaque de support et comprenant des électrodes de rangées et de colonnes, ledit dispositif comprenant une pluralité de paires de premières et de deuxièmes électrodes de pixels opposées agencées sur les plaques de support, chacune desdites paires d'électrodes définissant un pixel respectif (12) dans le milieu d'affichage et étant connectée électriquement à

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des électrodes respectives parmi les électrodes de rangées (11) et de colonnes (8), le dispositif comprenant des moyens pour appliquer des tensions de sélection et de données pour sélectionner et commander, respectivement, des pixels sélectionnés pour les amener dans des états prédéfinis de transmissivité, caractérisé en ce que, pour chacun desdits pixels sélectionnés, lesdits moyens pour appliquer lesdites tensions de sélection et de commande comprennent :

des moyens pour charger le pixel (12) à une tension auxiliaire au-delà ou à la limite de plage de tensions susceptible d'être utilisée pour l'affichage d'images via un élément de commutation symétrique respectif sensiblement non linéaire (15) qui est connecté électriquement à l'une des électrodes du pixel, et

des moyens pour charger le pixel (12) via l'élément de commutation non linéaire sensiblement symétrique (15) de la tension auxiliaire à une tension de même signe, mais de moindre grandeur, à laquelle ledit état prédéfini de transmissivité est effectué, un chemin de courant étant présent à la fois pour la charge à la tension auxiliaire et pour la charge à la tension de moindre grandeur.

12. Dispositif d'affichage selon la revendication 11, caractérisé en ce que les éléments de commutation symétriques sensiblement non linéaires (15) appartiennent au groupe des MIM, des diodes montées tête-bêche, et des éléments de commutation nin et pip.